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CCSDS

The Consultative Committee for Space Data Systems

Report Concerning Space Data System Standards

WIRELESS NETWORK COMMUNICATIONS OVERVIEW FOR SPACE MISSION OPERATIONS

INFORMATIONAL REPORT

CCSDS 880.0-G-2

GREEN BOOK

April 2015



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FOREWORD

This document is a CCSDS Informational Report, which contains background and explanatory material to support the CCSDS wireless network communications Best Practices for networked wireless communications in support of space missions.

Through the process of normal evolution, it is expected that expansion, deletion, or modification of this document may occur. This Report is therefore subject to CCSDS document management and change control procedures, which are defined in *Organization and Processes for the Consultative Committee for Space Data Systems* (CCSDS A02.1-Y-4). Current versions of CCSDS documents are maintained at the CCSDS Web site:

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1 INTRODUCTION

1.1 PURPOSE

This report examines the possibilities and advantages of the onboard application of *wireless communication network technology* to space missions. This Green Book describes a set of driving use cases in the space domain and evaluates the utilization of existing technologies and related terrestrial commercial standards to meet the resulting space-based use case requirements. Also included is relevant tutorial information intended to assist the reader in understanding basic concepts of wireless transmission and networking along with possible issues related to the deployment of wireless networks.

The information provided in this report will enable member agencies to select the best option(s) available for space communications and internetworking, based upon evaluation metrics such as network topology, power expenditure, data rates, noise immunity, and range of communication as well as on space systems metrics such as reliability, availability, maintenance and safety.

This document is a CCSDS Informational Report and is therefore not to be taken as a CCSDS Recommended Standard.

1.2 SCOPE

As demonstrated by the terrestrial marketplace, the potential uses of wireless technology are extremely broad. This ubiquity of use is also expected in the space domain and as a result wireless communications will cross the boundaries of existing areas of discipline where wireless transmission was typically limited to space-to-ground links. In an attempt to categorize its use, the CCSDS has identified the following application domains:

- a) **Intra-vehicle:** internal vehicle (or habitat) extremely short-range wireless links and networking as well as external vehicle-to-vehicle proximity communication wireless links and networking (up to a few 100 m range);
- b) **Inter-vehicle:** vehicle-to-vehicle short-range and medium range (up to tens of kilometers);
- c) **Planetary surface-to-surface:** wireless links and networking (up to several kilometers);
 - 1) Extra-Vehicular Activity (EVA) local links with planetary Rover Vehicles (RVs) and/or habitats;
 - 2) RV-habitat links when RV is close to habitat;
 - 3) links between independent local systems (e.g., habitats, robots, external assets);
- d) **Planetary Surface-to-Orbiter:** links and networking.

The Wireless Networking Communications document will be utilized as the basis for generating recommended practices for the application of wireless technology in the intra-vehicle, inter-vehicle, and planetary surface-to-surface domains.

1.3 RATIONALE

From an engineering standpoint, mission managers, along with engineers and developers, are faced with a plethora of wireless communication choices, both standards-based and proprietary. The provision of a CCSDS standard reference that summarizes wireless protocol capabilities, constraints, and typical deployment scenarios, will decrease the up-front engineering evaluation effort significantly, and provide a standards-based common reference to improve interoperability between disparate systems that need to cooperate in wireless data transmission and networking. Onboard systems standards are considered essential for fostering onboard interoperability (reference [1]).

1.4 DOCUMENT STRUCTURE

NOTE – This document is use-case oriented. As a result of this organizational paradigm, respective use cases follow rationale and benefits, with the detailed technical analyses and wireless standards review following as sections 4 and 5.

Section 2 provides an overview of the rationale and benefits of wireless network technologies for use in space operations.

Section 3 provides a set of high-priority canonical use cases as driving scenarios illustrative of selected wireless communications problem domains. Additional use cases are included as annexes.

Section 4 provides a detailed overview of wireless communications technologies and wireless communications standards.

Section 5 provides a comprehensive review of relevant standards-based wireless network communication technologies.

Section 6 overviews ElectroMagnetic Interference (EMI) and ElectroMagnetic Compatibility (EMC) issues for spacecraft in general and potential impacts of wireless networking transmissions.

Section 7 provides a report summary, conclusions, and recommendations regarding the most promising wireless technologies for identified application domains and use cases.

Annex A provides a list of commonly used acronyms associated with the field of wireless networking.

Annex B provides a glossary of terms commonly used in the field of wireless networking.

Annex C provides a number of *quick* reference tables including (1) a summary table of IEEE Wireless Personal Area Network (WPAN), Wireless Local Area Network (WLAN), and Wireless Metropolitan Area Network (WMAN) standards activities at the time of report publication; (2) detailed WPAN/WLAN specifications; (3) the International Telecommunication Union (ITU) Radio Frequency (RF) frequency designations for the Industrial, Scientific, and Medical (ISM) bands; and (4) commonly used RF band designations.

Annex D provides a compendium of additional use cases in the inventory management application area.

Annex E provides a compendium of additional use cases in the intra-spacecraft (intra-vehicle) application area.

1.5 DEFINITIONS

air interface. The term used within the field of wireless communication to refer to the Physical-Layer communication protocol for a generic wireless RF propagation medium. This same term is used to refer to many different propagation environments, including free-space, atmospheric, multi-path, line-of-sight (LOS), and Non-Line-Of-Sight (NLOS) environments.

data rate. The rate, measured in units such as kilobits per second (kb/s), Megabits per second (Mb/s), Gigabits per second (Gb/s), etc., at which data is transmitted across the wireless medium from the Physical Layer of a transmitting radio to the Physical Layer of a receiving radio.

low data rate. Data rates of 250 kb/s or less.

medium data rate. Data rates above 250 kb/s but less than 10 Mb/s.

high data rate. Data rates above 10 Mb/s but less than 100 Mb/s.

very high data rate. Data rates above 100 Mb/s.

frequency. The radio wave transmission rate of oscillation, measured in cycles per second (Hz).

interference. Unintended RF energy present in the operating frequency band of a system resulting in performance degradation to the intended communications link.

network. A connected, potentially routable and multi-hop, communication infrastructure for data transmission between multiple communication nodes.

optical. Communication networks that use light (visible, infrared or ultraviolet) as the transmission medium.

RF. The radio frequency segment of the electromagnetic spectrum, from 3 Hz to 300 GHz.

RF coexistence. The capability of a wireless network to operate properly in an environment in which noise and interference are present, e.g., a state in which two or more RF systems function within acceptable levels of mutual interference.

RFID. Radio Frequency Identification: refers to a system that automatically identifies various items and cargo by means of a simple radio transponder.

WLAN. Wireless Local Area Network: the linking of two or more devices into a data exchange network without wires. The dominant WLAN standard is IEEE 802.11, which from its inception was designed to be a wireless replacement of its wired IEEE 802.3 counterpart. IEEE 802.11 WLANs are commonly referred to as ‘Wi-Fi’ for wireless fidelity devices and networks. WLANs have a typical radio range of 150 meters and typical maximum theoretical data rates over 1 Gb/s.

WMAN. Wireless Metropolitan Area Network: geographically wide area wireless networks. The IEEE 802.16 standard, commonly known as Worldwide Interoperability for Microwave Access (WiMAX), has ranges from 5–20 km and (theoretical) data rates from 40–120 Mb/s.

WPAN. Wireless Personal Area Network: low power, low(er) data rate networks that typically involve little or no additional network infrastructure. WPANs have a typical range of 10 meters and data rates from a few kilobits per second up to 1 Mb/s, although IEEE 802.15.3 is a wideband protocol with data rates up to 400 Mb/s. WPAN standards are embodied in the IEEE 802.15 family of standards.

wireless. The transmission of data via electro-magnetic propagation, specifically via a digital packet communication network.

WSN. Wireless Sensor Network.

1.6 REFERENCES

The following documents are referenced in this Report. At the time of publication, the editions indicated were valid. All documents are subject to revision, and users of this Report are encouraged to investigate the possibility of applying the most recent editions of the documents indicated below. The CCSDS Secretariat maintains a register of currently valid CCSDS documents.

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2 OVERVIEW

2.1 RATIONALE AND BENEFITS

Wireless communication networks are an enabling technology for both manned and unmanned spacecraft; they enable un-tethered mobility of crew and instruments, increasing safety and science return, and decreasing mass and maintenance costs by eliminating expensive cabling. Wireless networks automatically enable communication between compliant devices that dynamically come into and out of range of the network. The focus of this document is on networked wireless communication rather than point-to-point communication. While point-to-point wireless communication is fundamental for communicating outside of a spacecraft (e.g., inter-spacecraft communications, planetary surface communications) the introduction of wireless networking enhances external communication in the vicinity of a spacecraft and also facilitates many aspects of communication within a spacecraft including mobile crew monitoring and communication, environmental monitoring and control, structural monitoring, and situational awareness. Added value for using wireless networks is also identified for ground mission support and Assembly, Integration, and Test (AIT) activities.

Several important advantages of wireless networks for space applications are summarized in table 2-1.

Table 2-1: Advantages of Wireless Networks for Space Applications

Benefit	Feature
Mobility of crew, sensors and instrumented systems	Enables operational communications capabilities that could not be accomplished otherwise.
Harness complexity reduction/elimination	Wireless communication enables the elimination of complex, expensive, cable harnesses.
Eases retro-fit activities	Wireless technologies facilitate add-on capabilities to existing vehicles without significant engineering (e.g., mechanical, electrical) effort.
Mass and volume reduction	Wireless communication enables the elimination of cables and supporting infrastructure (cable runs, cable ties, which can amount to 10 percent of total vehicle mass).
Lowers cost of distribution	Broadcast mechanism provides a relatively low cost of content distribution; can add users and systems in a cost-efficient manner (point-to-multipoint).
Reduced cost through flexible infrastructure	Elimination of infrastructure associated with wired systems.
Simplification of AIT activities	Wireless communications simplifies and eliminates any wired-biases associated with functional ground testing of the complex systems of modern spacecraft in addition to minimizing contamination issues and simplifying structural considerations.
Common network for onboard and off board communications	A single transceiver may be used for both onboard (intra-spacecraft) and off-board (inter-vehicle or surface) communications.
Rotating mechanisms and articulated structures	Wireless technologies are the easiest and sometimes the only way to implement contact-less data communications and acquisition systems.
Layout independence	Wireless techniques may bring additional flexibility when implementing fault tolerance and system reconfigurations.
Convenience	Allows access to network communications from anywhere within the range of the network, reduce complexity of operation and associated risk.

Benefit	Feature
Ease of deployment	Set-up of a infrastructure-based wireless network requires only an access point.
Flexibility	Within radio coverage the wireless nodes can communicate without restriction. RF radio waves can penetrate non-conductive walls so it is feasible that a sender or receiver could be hidden within or behind a physical wall.
Ad-hoc networking	Wireless ad hoc networks enable communication between compliant devices without the need of a planned system as would be required with a wired network.
Small form factor	Wireless devices are engineered to low mass, power and volume requirements, all three of which are fundamental constraints in spacecraft design.
Fault tolerance	Wireless devices can survive disasters, such as a catastrophic event of nature or even the common occurrence of a power loss (blackout). As long as the wireless devices are intact, all-important communications still exist.

Two important challenges associated with wireless networks for Space Applications include:

- a) **Quality/Reliability of Service:** Wireless networks typically offer greater challenges to providing quality and reliability of service than their wired counterparts, manifested as potentially lower data rates, higher bit error rates, and higher delay and delay variation (jitter). The underlying causes for these attributes include lower signal levels due to (typically) low directivity in coupling of energy between transmit and receive antennas, higher noise levels due to interference from multiple users and multiple systems operating in the same frequency band or spurious emissions from electronic equipment, signal fading due to multipath propagation, etc.
- b) **Safety/Security:** Using radio waves for data transmission might interfere with other critical equipment in the environment, e.g., spacecraft or test facilities. Additionally, the open-air interface makes eavesdropping much easier in wireless networks as compared to wired networks.

The issues of link quality and reliability-of-service lead effectively to less efficient link operation that must be offset against the benefits mentioned in table 2-1.

Space assets in close proximity or environmental factors are most likely to present challenges for wireless systems. Terrestrial environments are generally highly populated with wireless systems and therefore provide a useful context for the development and testing of wireless systems. If a space system is able to cope with the RF conditions found on Earth, it is likely that it will cope with situations it encounters in space, though there is no guarantee of this; hence caution and thoroughness of approach is necessary. In common with other space equipment, wireless system designs must also take account of the space environment in which they will spend their operational lives.

Wireless solutions should only be adopted if they do not compromise critical operations and allow adequate data throughput and timeliness. In some cases, wireless links may provide flexible, redundant (non-critical) communications or serve as complementary services to increase data volumes without the need for high levels of infrastructure. Such hybrid approaches can offer the best of both wired and wireless approaches, and can offer a dissimilar implementation for data transfer, thus increasing the overall data system reliability.

When designing space equipment and systems, the probability and impact (effect) of unintended events (e.g., malfunctions, misapplication, interference, failure, etc.) must be considered. For space systems such events can have much greater impact compared to terrestrial applications. This is due principally to the inaccessibility of space assets once launched and the difficulty and complexity of operating such systems at great distances. This must be borne in mind when designing and implementing wireless systems, thus ensuring not only safe and sustainable operation of critical assets, but also high levels of data return from such expensive assets and operations. When wireless systems are carefully designed and implemented, they can offer robust, flexible, highly adaptive solutions and many benefits for a whole range of missions, from design, integration, launch, and through sustained mission operations.

The Wireless Working Group adheres to the CCSDS guiding principal of a ‘3-Tier Prioritized Approach to Standards’:

- a) adopt proven standards where practical;
- b) adapt existing standards to meet defined requirements;
- c) develop new approaches only where absolutely necessary.

NOTE – Inclusion of any specific wireless technology does not constitute any endorsement, expressed or implied, by the authors of this Green Book or the agencies that supported the composition of this Green Book.

2.2 KEY APPLICATION AREAS

For the CCSDS categorization of functional wireless networking communication domains as (1) intra-vehicle, (2) inter-vehicle, (3) planetary surface, and (4) surface-to-orbiter, table 2-2 provides a summary of key application areas with associated network engineering characteristics. Table 2-3, on the following page, provides specific rationale and additional description of these important application areas.

Table 2-2: Key Application Areas for Functional Space Communication Domains

Functional Domain	Application Areas	Number of nodes	Data Rate	Range	Applicable Standards
Intra-vehicle	Inventory monitoring	100s	Very Low	< 10 m	ISO 18000-6C EPCglobal
	Environmental monitoring (e.g., temperature, pressure, humidity, radiation, water quality)	10s to 100s	Low to Medium	< 100 m	802.15.4 802.15.4e ISA100.11a
	Physiological monitoring (includes EVA suit biomedical monitoring)	1 to 10	Low to Medium	< 100 m	802.15.1 802.15.4 802.15.4e ISA100.11a
	Crew member location tracking	1 to 10	Medium to High	< 300 m	802.11 802.15.3 802.15.4 802.16 LTE
	Structural monitoring	10s	Medium to High	< 300 m	802.11 802.15.3
	Intra-spacecraft communications (voice and video)	10s	Medium to High	< 300 m	802.15.1 802.11 802.16 LTE
	Process monitoring and automated control and Scientific monitoring and control	10s to 100s	Low to High	< 300 m	802.15.3 802.15.4 802.15.4e ISA100.11a 802.11 802.16 LTE
	Retro-fit of existing vehicle with new capabilities	10s to 100s	Low to High	10 m – 100 km	802.15.3 802.15.4 802.11 802.16 LTE
AIT activities	Spacecraft assembly, integration and test	10s to 100s	Medium	< 100 m	802.15.3 802.15.4 802.15.4e ISA100.11a 802.11
Inter-vehicle	Inter-spacecraft communications (voice, video and data)	10	High to extremely high	1 m – 100 km	802.16 LTE Prox-1 AOS
Planetary Surface	IVA-EVA, EVA-EVA, Habitat-to-LRV, LRV-crew communications (voice, video and data)	10	Medium to High	1 m – 50 km	802.11 802.16 LTE
	Robotic Operations	10s	Low to High	1 m – 50 km	802.15.3 802.15.4 802.11 802.16 LTE
Orbiter relay to Surface*	Surface-to-orbit communications (voice, video and data)	10	High to extremely high	> 200 km	LTE Prox-1 AOS
* Application areas not addressed in this Green Book					

Table 2-3: Important Applications with Corresponding Rationale

Application	Rationale	Description	Subcategories
Inventory management	Provide automated inventory management and inventory location for improved efficiency	Wireless sensors (RFID tags) affixed to all inventory critical resources	
Environmental monitoring	Safeguard the crew and the vehicle from hazardous environmental contaminants and off-nominal physical conditions	Wireless sensors measuring ambient environmental phenomena to ensure within specified range for long term habitation	Atmospheric monitoring, leak detection assessment; in-situ water quality monitoring; EVA suit monitoring; temperature, pressure, relative humidity monitoring; light level monitoring, acoustic level monitoring
Radiation dosimetry monitoring	Safeguard the crew and vehicle electronic subsystems from radiation storms and cumulative radiation effects	Crew-worn monitors and deployable monitors that provide local and remote alarming of off-nominal radiation conditions	
Physiological (crew health) monitoring	Ensure the physical health of the crew members for manned missions	Wireless sensors and integrated devices to measure standard biomedical parameters of the crew	Heart rate; EEG and ECG; respiration rate, blood pressure, pulse rate, pulse oximetry, temperature, glucose levels, caloric expenditure
Crew member location tracking	Optimize crew member activities; detect potential crew member psyche problems	Use a high-precision 3D wireless localization system to provide precise crew member location tracking	
Structural monitoring	Provide wireless sensors to measure structural dynamics of space vehicles	Structural monitoring, leak detection, spacecraft avionics monitoring, propulsion system monitoring	
General spacecraft communications systems	Eliminate cabling and provide for user or system mobility for voice, video and data systems	Wireless communications systems for space vehicle inter- and extra-vehicular activities	PDA's and laptop communications; internal and external (EVA) communications; planetary base communications infrastructure
Spacecraft assembly, integration, and test (AIT)	Provide mobile wireless systems to improve efficiency of the AIT process	Advanced computer diagnostic systems that have wireless communications	
Robotic operations	Provide communications to EVA systems and instruments (such as roving cameras for external inspection activities)	Uses include roving cameras for external inspection, specialized EVA vehicle instruments, drone command and control, drone formation flying	
Retro-fit existing vehicle with new capabilities	Eliminate expense of running cabling for new electronics by using wireless communications	Structural vibrational monitoring, external collision monitoring	
Intra-spacecraft wireless low power sensor networks	Provide onboard short range low power communication with potential mass and power reduction and for increased functionalities and flexibility in spacecraft design, construction and testing	Wireless sensors (temperature transducers, radiation monitoring sensors, accelerometers, etc.)	

2.3 RF SPECTRUM PLANNING CONSIDERATIONS

2.3.1 GENERAL

Spectrum is a limited natural resource and shared commodity. The International Telecommunication Union, ITU, is the United Nations (UN) lead agency for information and communications technology. It is founded on a set of treaties that date back to 1865 and have binding force in international law, i.e., the ITU Constitution and Convention, the Radio Regulations, and the International Telecommunication Regulations, as well as resolutions, recommendations and other non-binding instruments adopted by its conferences. Individual administrations may further impose national regulations and rules for spectrum use within their sovereign territories and possessions; therefore consideration of deployment locations must be included for terrestrial and space-to-Earth application/link design and standards. Spectrum management regulations and rules enable and assure compatible and most efficient use of spectrum for a multitude of applications, both terrestrially and in space.

Internationally, the RF spectrum is allocated by the ITU to various classes of radio service according to different regions of the world (see figure 2-1). Radio service classes include satellite service, science service, broadcasting service, and terrestrial (fixed, mobile, radio determination, amateur, and amateur-satellite) services. Wireless networking communication is considered an application rather than a class of services; therefore use of wireless technologies discussed in the sections above is determined by the purposes (science vs. commerce) and physical location (space or terrestrial) and is governed under existing regulations and rules of the ITU and applicable national regulations and rules.

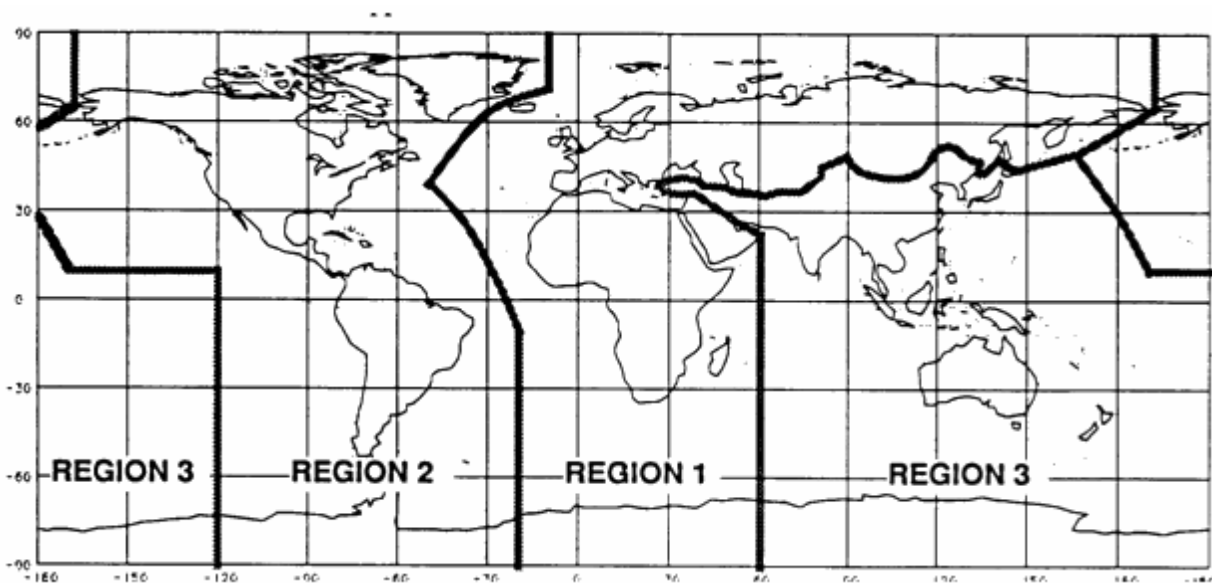


Figure 2-1: Geographic Regions for Frequency Allocation of the Spectrum

In addition to ITU regulations and rules, terrestrial use of wireless networking communications equipment must comply with local/national regulations and rules. For example, in the U.S., FCC part 15 certified devices, such as 802.11 devices, operating in the 2.4 GHz Industrial, Scientific and Medical (ISM) band and the 5 GHz Unlicensed National Information Infrastructure (UNII) band do not require individual license for each device but must operate on a non-interference basis and not cause harmful interference to licensed users in the band. While these devices are permitted to operate in the bands, they are not considered ISM equipment per ITU Radio Regulations definition; therefore they are operating in non-compliance to the Radio Regulations and cannot claim interference protection from any other users in the band nor create harmful interferences to other users.

Because of the unlicensed status of today's commercial wireless networking products that operate in the ISM bands, performance degradation due to in-band interferences may lead to the conclusion that unlicensed operational status is not acceptable for links carrying critical command/control data.

It is important, however, to recognize that modern advanced wireless communication standards are taking an increasingly more sophisticated approach to spectrum management. For instance, because of mobility requirements, and dense wireless communication systems deployment needs, modern mobile communication systems standards are emerging in which spectrum allocation is performed on a fully dynamic basis (see 4.3.5.3). These modern capabilities significantly modify forward-looking discussions of RF spectrum planning requirements.

2.3.2 SPACE SYSTEMS SPECTRUM REGULATION

2.3.2.1 General

For systems intended for operation in space where emitted RF energy is detectable by a large number of systems in low Earth orbit and on Earth, suitable spectrum for a terrestrial or an airborne application may not directly be usable in a space-borne application because of both limitations on the frequency allocations (regulatory, e.g., an aeronautical mobile service allocation will not be usable in space) and incompatible sharing with existing allocated services.

While this document highlights spectrum-planning considerations, it makes no recommendations for the actual allocation of frequencies for space use. This is solely under the responsibility of the relevant space agency RF spectrum managers in accordance with reference [2].

2.3.2.2 ITU Radio Regulations on Radio Astronomy in the Shielded Zone of the Moon

Regulatory issues have to be taken into consideration when evaluating RF technologies for planetary surface communications, for example, section V of Article 22 of the ITU Radio Regulations.

3 USE CASES

3.1 GENERAL

To properly scope the utilization of wireless technologies that are applicable to the space domain, this section presents several use cases for the two focused application areas of (1) inventory management and asset localization and (2) wireless communications for spacecraft. The use cases given are high-level operational scenarios that could directly benefit from the availability of wireless networking technologies. Illustrative diagrams are included where appropriate and specifications, as available at the time of report publication, are provided when available.

Subsections 3.2 and 3.3 each contain a set of design-driving, canonical use cases associated with inventory management and intra-vehicle wireless utilization, respectively. The set of reference use cases was selected as a means of focusing on a high-Technology Readiness Level (TRL) wireless communications system that can be expected to benefit space operations readily in the short term. Use-case scenarios in addition to those provided in this section are available in the annexes of this report, and it is expected that as technology matures, additional use cases, to be classified as canonical representatives, will be included in the subsections below.

Detailed technical analyses and wireless standards review follow in sections 4 and 5.

3.2 INVENTORY MANAGEMENT PROBLEM DOMAIN AND USE CASES

3.2.1 INVENTORY MANAGEMENT

3.2.1.1 General

RFID for automated logistics/inventory is an important international issue for future Exploration mission concepts. For human sustainability and supportability automated inventory logistics is identified as a primary technology to develop in support of human exploration systems in the 2012 NRC NASA Roadmaps report (reference [3]). Development and deployment on the International Space Station with corresponding international cooperation is specifically mentioned in reference [3]. Autonomous Logistics Management (ALM) is considered a high-priority adjacent technology to invest and develop in the *NASA Strategic Space Technology Investment Plan* (reference [4]). For these reasons RFID technologies (RFID RF transmission—reference [5], data communications, RFID tag-encoding and RFID sensing) are envisioned to be of primary importance to produce internationally recognized data standards for space agency utilization.

Inventory management is a critical function in many aspects of space operations, in both flight and ground segments. On the ground, thousands of controlled components and assemblies are stored in bond rooms across multiple centers and space agencies. These inventories are tightly controlled, typically using manual processes such as paper tags on individual items or small collections of identical items, such as small bags with screws. Bag

inventory is tracked by inking out the previous count and replacing with a revised count. In some instances, the process is aided with optical barcode technology.

Other ground operations also require complex inventories, including tracking all laboratory and office equipment with significant value. For example, at Johnson Space Center, a database containing approximately 38,000 items is maintained. Inventory audits of such equipment are currently very labor intensive and involve periodic room-by-room examinations and scanning of optical barcodes for each tagged item. Many inventory items require careful monitoring to assure, for example, that expiration dates are not exceeded. Replacement of consumables can also be highly critical; monitoring delivery and restocking of compressed gases and chemicals requires careful attention to assure, for example, that identical or compatible replacements are made.

Inventory management for flight applications entails an even greater degree of control, as improperly substituted items and early depletion of certain items can be catastrophic. Most short duration missions do not involve restocking, so resupply logistics are nonexistent, but initial stocking and tracking of inventories is nonetheless quite important. For most long-duration missions, resupply efforts are inherently complex, expensive, and infrequent. To date, the most extensive space-based inventory management operation has been the International Space Station (ISS). More detail on ISS inventory management, as well as a brief history of inventory management in human spaceflight, is provided below.

On the International Space Station, approximately 20,000 items are tracked with the Inventory Management System (IMS) software application. Both flight and ground crews update the database daily to reflect utilization of consumable items as well as delivery of new cargo and removal of some existing cargo via the regular resupply missions. A handheld optical barcode reader is used to update the onboard database, and the IMS application performs complex updates. The ground and flight segment databases are synchronized by uplinking and downlinking 'delta files'. The common transport apparatus for smaller items is the Crew or Cargo Transfer Bag (CTB) (see figure 3-1). The cargo ranges from crew clothing to office supplies, pantry (food) items, and personal effects. The CTBs are packed on the ground, and like items within a CTB are usually stored in Ziploc bags. For some cargos, items are tracked both at the Ziploc bag level and at the individual item level. For other cargo types, tracking resolution extends only to the Ziploc bag level. In addition, optical barcode tags are also affixed directly to the CTBs.

In the 2008 timeframe, approximately 500 CTBs were onboard the ISS at any given time. The CTBs are typically stacked several deep and are often restrained by webbing or lines. Inventory audits required approximately 20 minutes per day for each crewmember. The time required to inventory a single CTB is also about 20 minutes. The process requires removal of each Ziploc bag and each tagged item, orienting the barcode to enable line-of-sight reading, and re-bagging the items. The process is greatly complicated by the zero-g environment, which requires extra care to prevent items from floating out of reach.



Figure 3-1: Cargo Transfer Bags on the International Space Station

In addition to the tracking of smaller items packed in CTBs, localization of larger pieces of equipment has, at times, also proven to be difficult. Such difficulties might arise, for example, when the sought item is stored behind other cargo or closeout panels. Although this situation does not occur often, crew time can be significantly impacted when it does. Moreover, inability to locate critical equipment in a timely manner can entail obvious safety implications.

In 2005, RFID was investigated by NASA as a possible solution to inventory management problems. Studies of the technology were commissioned, including tests of the EPCglobal Class 1 Generation 1 standard. Although the read accuracy of the standard was believed too low to warrant immediate pursuit, later tests in 2006 of Surface Acoustic Wave (SAW) RFID showed greater promise (see reference [6]). In 2008, the first spaceflight RFID tests were conducted as a Station Detailed Test Objective. The test involved rotating a CTB in front of a fixed SAW RFID interrogator. In addition, the interrogator was used to locate a ‘hidden’ piece of equipment. Even though the read accuracy was less than the target 95 percent, the ease of audit, when compared with the optical barcode process, was found to be sufficiently improved to render a future operational RFID system highly desirable.

In 2008, NASA conducted tests of the EPCglobal Class 1 Generation 2 standard for interrogation of CTB cargos. The second generation showed considerable improvement over the first and over SAW RFID for the interrogation of tags in the CTBs. An additional study commissioned for the Crew Exploration Vehicle (CEV) Orion (see reference [7]) also found the Generation 2 implementation to be greatly superior to Generation 1. Although the CEV is not considered for long duration missions requiring resupply, it does constitute a supply ship for the ISS. As such, RFID is being considered for inventory management, including the transfer of items from the vehicle to the ISS.

3.2.1.2 RFID Return on Investment for Space Applications

Quantifying the potential savings that could be attributed to RFID for space operations is difficult, largely because of the complexities in attributing a cost to the crew's time. Nonetheless, a few attempts have been made, particularly in the context of the International Space Station. An abbreviated benefit analysis for RFID (see reference [6]) estimates potential savings of approximately 36 million USD per year.

A more in-depth cost-benefit analysis for RFID on ISS is provided in reference [8], although this analysis assumes the cost associated with a specific RFID implementation involving retrofitting or replacing the existing CTBs with an RFID 'wired' CTB. The wired CTB would have the capability to interrogate and report the contents of each CTB without crew involvement. Two different implementation scenarios are addressed: a gradual 'phase-in' in which new 'wired' CTBs would replace older ones as new supplies were transferred to the ISS; and a more abrupt transition in which existing CTBs would be enhanced via modification kits. The cost-benefit effects of many other variables are also studied. It is found that the more rapid transition is associated with a more favorable cost-benefit outcome, in large because of the limited planned life expectancy of the ISS. In some trials, the computed net value is found to be slightly negative; i.e., for the selected set of variables and implementation scenario, the incorporated 'wired-CTB' capability resulted in a mean net loss. The loss is greater for the gradual 'phase-in' scenario. For other variable combinations, the net value is significantly positive, and, in all cases, the standard deviation appears quite large.

The forward plan for ISS inventory management, as it relates to RFID, has not been determined as of the publication date of this document. Even if fully integrated and automated (i.e., audits and item localization involving little or no crew time) RFID is not realized on the ISS, it is likely that RFID will be incorporated to reduce the crew time expended in audits. The integration costs associated with a small number of onboard handheld RFID readers is expected to be much less than the cost of a larger number of RFID-wired CTBs.

For longer-term excursions in space, such as a lunar or Martian outpost, the complexities associated with inventory management are likely to greatly exceed those of the ISS. Indeed, the present day value attributed to RFID in reference [8] appeared to be largely restricted by the operational lifespan of the system on ISS. For longer-term outposts, the return on investment is expected to be quite large. Researchers in the Haughton-Mars Project estimated a time savings factor of 2–3, compared to optical barcode scanning, for inventory management based on an RFID gate, or portal experiment within the context of a remote outpost (see reference [9]). Larger comparative savings are attributed to larger quantities of tagged items, since the time required for RFID interrogation increases little with the number of items, in contrast to optical barcode scanning. It was noted in reference [9] that technology limitations at that time (2005) resulted in an accuracy of recording transactions between 70 and 85 percent. Several current and recent studies by, or for, NASA are examining recent improvements in RFID technology and integration of those technologies in a lunar habitat mockup test bed. These improvements will further increase the return on investment for RFID in space applications.

Several other factors will likely greatly decrease the cost of a fully automated RFID system for extended outpost scenarios. First, the technology will almost certainly improve over the next decade. This is especially significant since reader accuracy was found to be a critical cost variable in reference [8]. Second, integration is likely to be less costly when addressed at the outset of a new vehicle, as opposed to retrofitting an existing one. The routing of prime power for interrogators in necessary locations and the implementation of application software and middleware designed for integration of RFID technology are examples for which the associated cost should be much less when addressed in the early design stages of a vehicle. In addition, crew time, and hence cost, associated with retrofitting a vehicle (for example, see reference [10]) will not be applicable if RFID is integrated at the outset. It should be noted that the safety value associated with situational awareness and with the capability to rapidly find critical items lies outside the scope of the space-related cost-benefit analyses conducted to date.

Three design-driving high-priority inventory management use cases illustrate the potential benefits of a wireless IMS. Annex D contains additional inventory management use case scenarios for additional context.

3.2.2 GROUND-TO-LINE REPLACEMENT UNIT



Figure 3-2: RFID Ground-to-Line Replacement Unit Concept

Objective: Accurate and automated tracking of parts and Line Replacement Units (LRUs).

Description: RFID technology facilitates part tracking and inventory management. Use of RFID in commercial and U.S. Department of Defense (DoD) sectors supply logistics continues to increase. Space center bond rooms could replace existing paper tags with RFID tags. Tags are typically verified during or after tag attachment. Standards-based interrogators and tags permit read of vendor tag information. Part heritage material data, calibration data, and other information can be rapidly obtained in the context of an enterprise class network and broad interoperability with the supply chain. Advanced concepts, such as part environmental exposure history (e.g., shock or thermal extremes) are also possible.

Specifications:

Items tagged	Material
Components: bag level, LRUs	Conductive and non-conductive
Range:	1–3 meters
Reader type:	Portal, portable
Readability:	100 percent

3.2.3 INTRA-HABITAT EQUIPMENT/INVENTORY AUDITS

Cargo Transfer Bags (CTBs)



Figure 3-3: Cargo Transfer Bags Onboard the ISS

Objective: Inventory management and localization of assets.

Description: Provide audit capability of supplies, consumables, and equipment leading to a significant decrease in crew labor. This capability needs to be in place at the outset of planetary surface operations and exploration.

RFID technology can currently facilitate manual audits with portable reader (e.g., Personal Digital Assistant (PDA)-based).

Both ground- and flight-based assessment of crew-assisted RFID for item-level interrogation indicated 30–60 seconds per CTB, compared to over 20 minutes per CTB using an optical barcode scanner when reading all items in the CTB.

Special Considerations: Technology issues exist for full automation. Reliable item-level interrogation is currently an industry-wide issue for densely populated tagged items. Tag antennas can be obscured by other tag antennas, conductive or lossy items, and conductive storage containers. Combinations of existing technology, including ‘smart containers’, ‘smart shelves’ and ‘wired CTBs’ (see reference [8]) are likely to enable fully automated inventory audits.

3.3 SPACECRAFT WIRELESS COMMUNICATION PROBLEM DOMAIN AND USE CASES

3.3.1 GENERAL

To ensure that spacecraft vehicles and/or instruments are operating within defined nominal ranges, the relevant properties are monitored, assessed, and fed into a monitoring and control loop. The current solution is to route wired sensors throughout the spacecraft (or vehicle or habitat) to monitor critical and less critical areas; thermistors are used to monitor the temperature on space system surfaces, instruments, electronics and propulsion items; accelerometers are used to monitor launch vibration loads and spacecraft attitude; radiation sensors gather data on the direct particles environment for comparison with models. Other sensors are not meant to fly but are used on ground to provide more data points and to verify that the system meets (or exceeds) the requirements. These sensors (e.g., thermistors, thermocouples, three-axis accelerometers, etc.) are integrated onto the platform for verification testing and removed afterwards with a lifetime ranging from days to months.

Sensors are often directly linked to the onboard data handling system with harness that generally provides a data link and a power line. In a medium-class satellite where more than 400 of these sensors can be found, the related harness becomes a concern in terms of design, integration complexity, flexibility, and mass. For example, a considerable effort is required in planning the harness routes for each of the sensors, a process which is done early in the design phase. Each time a change is introduced in the design, the location of hundreds of cables dedicated to health monitoring sensors must be reviewed. The integration, testing, and debugging time is also a direct function of the amount of harness involved and generally leads to several days of work for the single integration process. It is worth noting that much time is lost during testing and integration because of errors or faults in the auxiliary equipment and related test harness. In the verification phase, technicians must route extra sensors and harness within the space system and test every connector, which introduces a considerable risk factor. These extra sensors and connectors have harnesses that protrude from the space system currently in test to connect to the Electrical Ground Support Equipment (EGSE), increasing the complexity of the test environment (e.g., clean chambers, thermal vacuum chambers, etc.). Some of these weaknesses are overcome by highly detailed and extensive procedures for technicians, to reduce human-caused risks, at the price of extra AIT time and cost. Moreover, the current wired solution does not provide much flexibility; at a stage where harness modifications are no longer possible, the late integration of opportunity payloads (e.g., micro-cameras for the deployment of appendices or separation maneuvers) on a spacecraft cannot be allowed. Another weakness of wired sensors is linked to launcher health data acquisition. Providing health data from launchers requires linking the sensors to long harness branches in order to reach the health data processing unit; the electrical signals being small, the harness needs to be protected against electromagnetic interferences in the form of shielding and bounding. Shielding further increases the mass of the upper stages, reducing the payload capacity.

Replacing the wires and connectors by wireless channels drives a series of consequences related to monitoring activities during test, launch, and flight phases. Numerous potential paybacks have been identified from using wireless technologies to reduce the complexity, AIT time, and cost of health monitoring applications in space systems:

- AIT technicians will spend less time in the assembly and integration processes;
- AIT procedures will be simplified, and the risk of mechanically damaging interfaces during tests and integration will be reduced;
- launchers might see a reduction of the harness mass and allow more payload capacity;
- late integration of opportunity payloads will have a better chance to be accepted;
- adding, removing, or replacing any remote sensor very late in the project is allowed;
- the test environment has fewer cables running out of the space system.

Wireless systems also introduce new functionalities that were just not possible with the current solutions:

- New redundancy concept: wireless techniques bring additional flexibility when implementing fault tolerance and system reconfiguration. In current systems, the cross-strapping of onboard equipment often introduces new potential fault mechanisms.
- Different users communicating at different speeds can share the same wireless channel. This is not possible with standard wired solutions since high speed signals require specific cables (shielding, coaxial).
- Off-board applications such as robotic surface elements may be interesting scenarios for wireless technologies.

Simulations have shown that replacing 70 percent of the *replaceable* data harness (not only health monitoring cables but also other data link types; see reference [10]) of a medium-class satellite, for example, the Mars Express, with wireless technologies results in about 20-percent reductions of Flight Model integration time and relevant associated integration phase cost (for Mars Express, it represents 25 days saving out of 130 for a team of about 15 people). There are many more studies discussing the benefits of reducing the amount of harness within the space industry.

The following subsection describes what is considered to be the highest priority applications that could benefit the most from wireless technologies.

3.3.2 SPACECRAFT HEALTH MONITORING

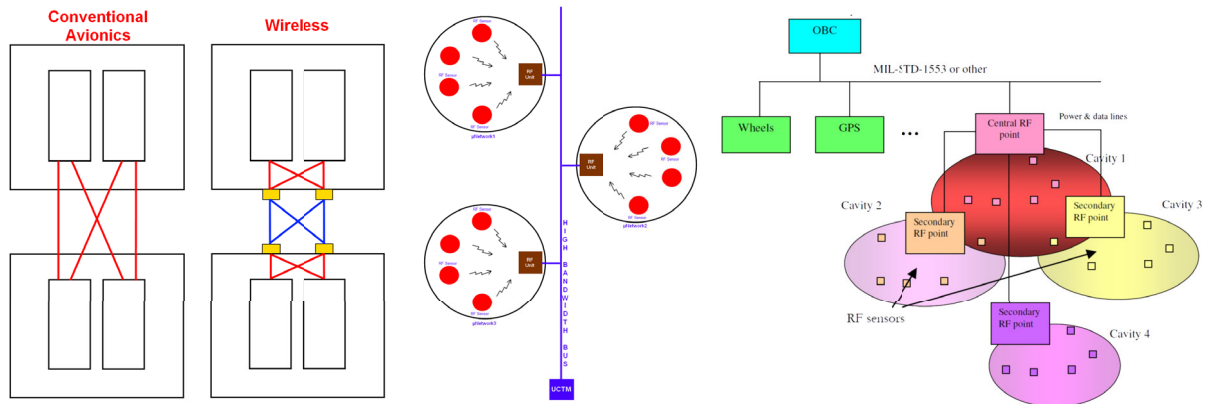


Figure 3-4: Wireless Health Monitoring (Redundancy, Launchers, and Intra-S/C)

Objective: Reduce harness related to health monitoring applications.

Description: With regard to robustness, power management, and flexibility, wireless sensor networking has made tremendous progress, which has led space agencies to study the possibility of using the technology within spacecraft, especially for non-critical health monitoring applications. In most cases, the required data rate is low and allows great receiver sensitivity and therefore a low transmitted power. Thermistors, thermocouples, accelerometers, and radiation detectors are the typical sensors to be integrated with the wireless interfaces. This use-case targets three similar application types: developmental flight instrumentation for spacecraft health monitoring during the test/verification phase, operational flight instrumentation for monitoring during the operational phase, and monitoring of the launcher during the launch phase. Launchers are between 30 and 60 meters tall, which results in long data cables. The short mission time of a launcher makes the wireless alternative advantageous in regard to the low-capacity, low-weight batteries that can be used to power the wireless interfaces and sensors. Studies have shown that it is possible to use technologies that will comply with the EMC constraints of spacecraft.

Special Considerations: Targeted unmanned launcher applications (non-critical) generally do not require real-time data transfers but do require accurate time tagging of the data for later analysis on the ground. For some types of sensor networks used by launchers, the reliability is not stringent (10^{-4}) but the availability is very important for the telemetry system.

The approximate size of the Wireless Sensor Network (WSN) provides a sense of the potential complexity of the network topology and the resulting complexity faced by routing protocols. The presence of *several cavities* within a spacecraft may require different network topologies to ensure the link budget in each one of the cavities. Because a low-power proximity sensor network would need to transport only one class of traffic, e.g., sensor data, greater traffic diversity may increase the need for the network to provide Quality of Service (QoS) assurance to the different classes of traffic.

Self-powered sensors allow the wireless sensors to be free from any power cables by embedding their own power source to supply the sensor, the internal electronic, and the radio device. The main constraint is the lifetime of the battery, which is directly dependent on the average consumption of the unit. Roughly, high data rate sensors will be usable only on short missions (launchers, vibration or shock monitoring, manned station with maintenance, etc.) while use on long missions of several years will be possible only with ultra low consumption units needing a very limited number of transferred bits.

Highly efficient air message formats should be used to minimize the power consumed while transmitting data over an RF link. Where possible, compute cycles should be traded-off against bits transmitting on the medium, even though developing general rules for making these trades is very difficult. It could nevertheless be useful in some cases for the Network Layer protocol to provide a facility to compress application data (e.g., sensors transmitting a high amount of data).

The EMC compatibility between the low-power sensors and the spacecraft is a potential design constraint. Limited emission power is needed in order not to disturb any unit located inside the spacecraft. The frequency band of the emitting sensors needs to meet the EMC requirements of the spacecraft.

Many Commercial Off-The-Shelf (COTS) wireless standards and technologies are able to provide a technical answer to the wireless sensor bus concept for space. However, their enhancement is likely to be needed, if only to withstand the harsh space environment.

Currently available technologies could reduce the risk of lengthy and expensive development programs. Several criteria can be considered when evaluating the current state of the technologies required for low power proximity sensor networks: applicability, reliability, scalability (can support large networks with few significant changes to the technologies), longevity, and technology readiness level. The compliance to international standards insures interoperability of different sensor devices and the long-term availability of wireless technology. The conformance to space requirements or the upgradeability to space qualified components is an asset for space use.

Specifications:

Network Attributes	Values
Range	10s of meters
Data rate	Typically low. Exceptions are found with accelerometers and other fast acquisition devices.
Data generation	Typically low.
Number of nodes	Typically high.

3.3.3 TEST AND AIT SUPPORT TOOLS



Figure 3-5: Technicians in the AIT Process

Objective: Reduce the complexity of test harness within clean rooms and test chambers.

Description: Testing a space system, its subsystems, or one of its instruments requires the integration of extra, temporary sensors for vibration tests or for a thermal vacuum session. Harnesses for these sensors can get very messy if the procedures are not accurately followed. Data and power links protrude from the satellite to link with the electrical ground support equipment making the data acquisition. Cable bundles are complex, delicate, and most of the time in the way of the technicians. Replacing the data wires with a wireless equivalent is thought to offer significant technician-time savings as well as simpler test procedures. There are several types of health characteristics that are monitored: health monitoring test applications using low data rate wireless interfaces between the individual nodes and the EGSE and spacecraft/instruments data bus traffic that is using a high-bandwidth channel to receive a copy of the bus content (wireless interfaces connected to the bus and to the EGSE, the system being used as a bridge). This use case therefore also targets wireless bridges for instruments using high-speed data links like SpaceWire between spacecraft and EGSE.

Specifications:

Network Attributes	Values
Range	10s of meters
Data rate	Typically low for health sensors and medium for data bus bridge
Data generation	Typically low for health sensors and medium for data bus bridge
Number of nodes	Typically high for health sensors and low for data bus bridge

3.3.4 PLANETARY EXPLORATION SENSORS

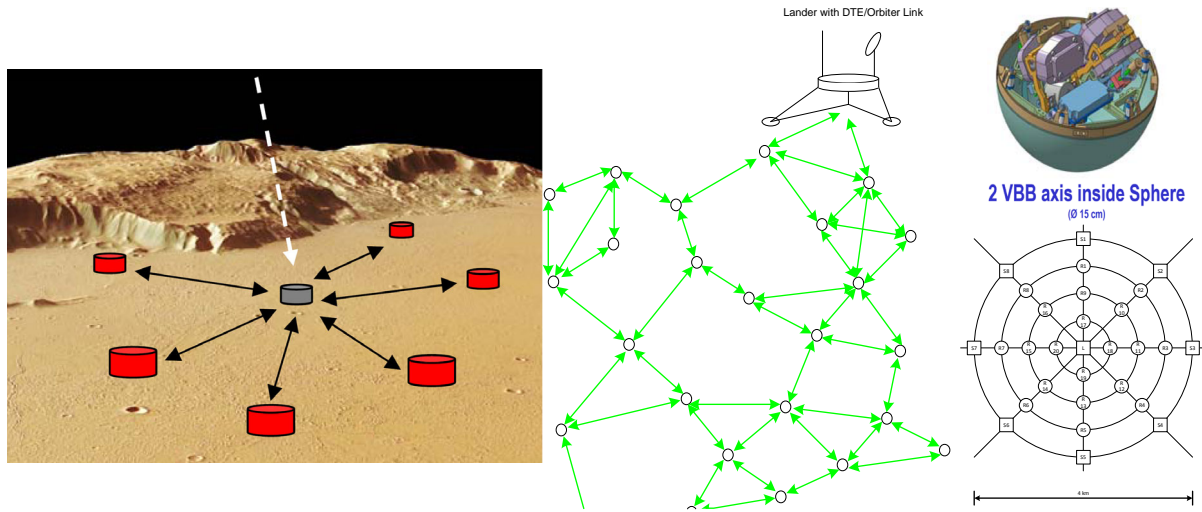


Figure 3-6: Planetary Exploration Applications Using Wireless Sensor Networks

Objective: Obtain extra science data during planetary exploration missions.

Description: Planetary surface exploration is a key goal for several Agencies and offers a great deal of science return. For a short or medium range (hundreds to thousands of meters), self-powered wireless payloads are considered as an extension of the master spacecraft (e.g., a lander), therefore justifying their pertinence in the intra-spacecraft class of wireless use-cases. Most of the use-cases are based on a lander-payload scheme, where the payload is made of one or several science instruments connected to the lander/rover through a wireless network of sensors. During the descent, probes are released and create a mesh network to relay the data to the lander/rover. Meteorological and geological units transmit, on a periodic basis, parameters such as atmospheric pressure, temperature, wind speed, humidity, light intensity, and soil constituents. Study of the seismological behavior of planetary bodies might generate very valuable science data and an understanding of the current activity of its core, where two important parameters are the accurate timing and the known position of the nodes.

Special Considerations: Similarly to launcher applications, planetary exploration applications generally do not require real-time data transfers but put more emphasis on accurate time tagging of the data. Time tagging, as well as synchronization, will determine the quality of the data (e.g., data obtained during atmospheric entry phase).

Specifications:

Network Attributes	Values
Range	10s to 100s of meters
Data rate	Typically low
Data generation	Typically low
Number of nodes	Typically medium to low

3.3.5 INTRA-SPACECRAFT WLAN

Objective: Provide wireless links for internal delivery of voice communications, video, and other data.

Description: WLANs are commonly used in terrestrial applications to access a variety of services from wireless devices. These can include peer-to-peer voice and video communication, on-demand distribution of video, and dissemination of data such as files (File Transfer Protocol [FTP]) and web pages (HyperText Transfer Protocol [HTTP]). It is to be expected that such services will be common in the spacecraft domain as well, with crewmembers accessing the WLAN through portable devices such as PDAs, laptop computers, and Voice-over-Internet-Protocol (VoIP) appliances.

Special Considerations: Analysis is needed of several wireless protocols utilized terrestrially with the capability to provided wireless LAN functionality internal to a vehicle. Of particular importance is security and quality of service provision, which is highlighted when transmitting crew health or ambulatory data.

Specifications:

Network Attributes	Values
Range	10s of meters
Data rate	Typically high
Data generation	Typically high
Number of nodes	Typically low

3.3.6 EVA SUPPORT

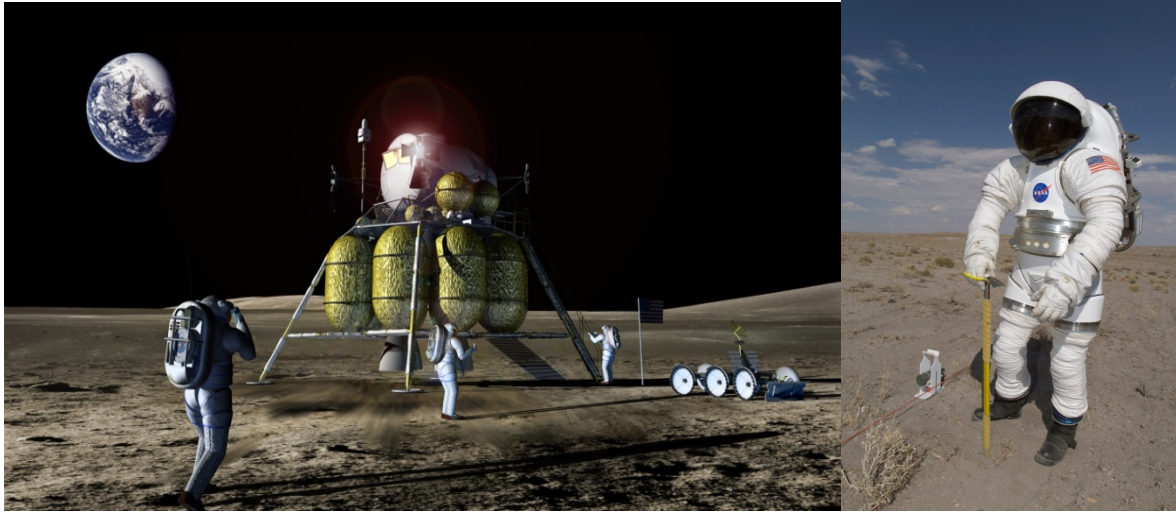


Figure 3-7: Wireless Networks for EVA Support

Objective: Provide wireless links for voice, telemetry, video, and other EVA data flows.

Description: Wireless communication is required for supporting extra-vehicular activities (EVAs) in proximity to a spacecraft, surface habitat, or surface vehicle without the encumbrance of a tether. EVA data flows may include voice, suit and crew health telemetry, video, and possibly navigation. Spacecraft proximity EVAs generally involve one or two crew members operating within 10s of meters of the spacecraft wireless access point for up to about 8 hours. Current surface mission concepts (i.e., proposed Moon and Mars missions) allow for up to 10 km between the crew members and an access point, which may be a surface habitat, relay communications terminal, or a surface vehicle. The most advanced surface mission concepts include several pairs of crewmembers performing simultaneous EVAs while relaying data through each other's suits and wirelessly interfacing with various instruments and experiments.

Special Considerations: EVA wireless networks must carry mission-critical data such as voice and crew/suit health telemetry, as well as non-critical data that include video, file transfers, and instrument telemetry. The mission-critical data is generally streaming data with latency and jitter constraints, and any interruption of these data flows will halt or possibly even terminate an EVA. This data must take precedence over non-critical data, and contentious channel access may not be acceptable.

Since crewmembers are mobile and EVAs often occur in dynamic environments, shadowing and multipath can be problematic. The diffraction characteristics of the UHF system currently used to support EVAs on ISS provide favorable external wireless coverage under a wide range of operating conditions. However, it is unlikely that enough UHF bandwidth will be available to support all of the EVA data flows.

A final consideration concerns a NASA requirement that EVA crewmembers must be able to communicate directly with each other for relay purposes (e.g., exploring craters or caves) or during a contingency scenario if the access point fails. This requirement has considerable implications on the design of the radio network.

Specifications:

Network Attributes	Values
Range	10s to 1000s of meters
Data rate	Typically medium (< 1 Mb/s) for most flows but high for video flows
Data generation	Typically high
Number of nodes	Typically low

3.3.7 FRACTIONATED SPACECRAFT

A fractionated satellite is a set of independent platforms hosting different sub-systems and capabilities of a single spacecraft, which autonomous modules are free-flying around each other within a small cluster. Several American (DARPA, MITÖ) and European entities have been investigating the concept and its benefits for industrials and customers. Globally, it is argued that fractionating a satellite may offer a greater overall value of a space system, where value is no longer solely related to cost, mass and revenue but is influenced by other non-traditional characteristics.

Spacecraft designers seem to react to the very long spacecraft development time and high costs by tending to increase the capability and lifetime of the system, which consequently increases its size and complexity, which again increase the costs and risks associated with a single mission. The robustness of a spacecraft is currently implemented through higher margins and parallel redundancy, which also contributes to this scheme. Another characteristic of current space system design is the selection of the capability of a spacecraft that is based on the predicted demand for the upcoming years. As the time between the capability selection and the actual commissioning of the spacecraft can be rather long, the uncertainty of an over/underestimation of the demand is higher and may result in a payload with too little capability, losing market shares to a competitor, or too much of it, having paid too much for the actual revenue. One must remember that the current over-capacity in transponders is about 30 percent. This is inherent from the lack of flexibility and scalability of monolithic spacecraft. One must also recall how difficult and sometimes disastrous a change in the requirements at design time can be, and how it can literally hurt the rest of the spacecraft. All these characteristics of space system design make it very difficult to rapidly respond to uncertainties. This responsiveness is an important parameter in any other type of businesses, as being vital to companies in reaction to planned or unplanned events that may affect their position in their market. Telecommunication, navigation or defense industries have no reasons to differ from this line of thinking, as they nowadays also evolve in a competitive environment. Launchers have an important and very early impact on a satellite design and their limitations are additional constraints that translate into design requirements. Its capability and availability dictates the initial size and limitation of a spacecraft.

Data transmission on fractionated architectures is meant to replace wired buses with a series of wireless links between the nodes. It is indeed well known that wireless networking has made very important progresses in the last decade and has already its influences on the aerospace industry. Wireless technology is a prerequisite enabler for fractionated spacecraft where data exchange is required by each node, with different data rates, behavior and security characteristics. Data rate between, e.g., the telecommand/telemetry module and the onboard computer, is considered high, together with the external payload, while the data throughput of other modules is considered low. Several wireless technologies can be used within a single fractionated spacecraft, and software-defined radios would offer great flexibility and scalability when delivering supplementary spacecraft modules to the fractionated entity.

3.3.8 BIOMEDICAL SYSTEMS SUPPORT

Objective: Provide wireless links for voice, physiological signs, video, and other biomedical data flows.

Description: Wireless communication is required to support both nominal and off-nominal medical events within the spacecraft, surface habitat, or surface vehicle without the encumbrance of a tether. Biomedical data flows are likely to include voice and video, to relay procedural information, and environmental and physiological monitoring data, to provide situational awareness to aid in decision-making. It is anticipated that medical care will be provided in one dedicated area within the various types of habitats. The care provided by that habitat will be dictated by the anticipated medical conditions that are most likely to occur in that habitat. Generally speaking, operations within this area will involve one to three crewmembers separated by meters from each other and from the vehicle and can last anywhere from minutes for routine nominal services to multiple days for off-nominal medical emergencies. The majority of medical events encountered during an exploration mission are expected to be routine, ambulatory, and primary-care issues. Only a small percentage, less than 10 percent, would require advanced medical capabilities. Routine, nominal medical services can occur simultaneously in all of the habitats, but off-nominal medical emergencies are likely to only take place within one habitat.

Special Considerations:

Biomedical wireless networks must carry critical and non-critical data, but a complicating factor is that data criticality can change based on situation. For example, in one scenario video can be the most prioritized data, but in a different scenario, it may not even be a critical data flow. The biomedical wireless network requires dynamic data criticality assignment based on the specific scenario.

Biomedical wireless networks will carry extremely time sensitive information that require start up times on the order of milliseconds. Generally speaking, early diagnosis significantly improves outcome, but in certain conditions, seconds can literally mean the difference between crew survival and loss of life.

Biomedical wireless networks, while often will be used for nominal events, will also be used for off-nominal emergencies. In these situations, where stress levels are extremely elevated, the network must be self-forming and be as simple and streamlined as possible and not require any crew interaction/intervention.

Specifications:

Network Attributes	Values
Range	Meters
Data Rate	Typically medium (<1 Mb/s) for most physiological monitoring activities (ECG, EEG, HR, BP), but high (>1 Mb/s) for video and other image acquisition activities (ultrasound, MRI, etc.)
Data Generation	Typically High
Number of Nodes	Typically Low

4 WIRELESS NETWORKING TECHNOLOGIES

4.1 INTRODUCTION TO WIRELESS NETWORKING TECHNOLOGIES

This section provides a summary overview of wireless networking technologies and engineering issues associated with the deployment of wireless networks. Properties of wireless networks as compared to wired networks are summarized and basic concepts of optical and RF wireless networks are given. RF coexistence, RF and optical propagation, and multiple access schemes along with multiplexing are examined in sufficient detail in order to provide the reader with a basic knowledge of common issues that may afflict wireless networking technologies.

Annex C provides a number of quick reference tables regarding current IEEE WPAN, WLAN, and WMAN standards activities; detailed WPAN and WLAN specifications; along with commonly used RF band designations associated with wireless communications and networking for the interested reader.

4.2 PROPERTIES OF WIRELESS NETWORKS

Wireless communication networks have several differences from their wired counterparts. Wireless communications are key to enabling mobility, often have lower cost because of the elimination of infrastructure associated with wired systems, and are inherently a broadcast transmission medium. The fact that a single transmission is received by any sufficiently proximate number of receivers is often referred to as the ‘wireless advantage’. Ease of broadcast produces a relatively low cost of distribution (e.g., television and Wi-Fi hotspots) and enables the addition of users in a cost-effective manner since the communication is point-to-multipoint. Furthermore, cooperation among users who all share the same broadcast information can be exploited to dramatically improve the overall performance of a network.

Typical wireless data networks are exemplified by standards such as IEEE 802.11 (Wi-Fi), IEEE 802.15.1 (Bluetooth), IEEE 802.15.4 (wireless personal area networking), including both ZigBee and ISA100.11a, and Long-Term Evolution networks (3GPP LTE). The basic properties of wireless data networks are:

- a) there are many transmitters and receivers;
- b) communication is mainly over wireless links;
- c) users can be mobile; thus the network is dynamic in terms of membership;
- d) communication is network packet-based.

There are several characteristics of the wireless channel that must be mitigated to provide reliable communications:

- a) there is very high signal attenuation by the environment;

- b) antennas gather all of the spurious energy in the environment including base thermal noise floor, interference, along with the desired signal; transmission is very noisy and subject to a higher Bit Error Rate (BER) than wired communication;
- c) the wireless broadcast channel is inherently insecure; there is no physical security to prevent spoofing of data packets;
- d) the wireless channel is not necessarily symmetric and is not transitive (although the physical channel is symmetric, transmitters and receivers are not symmetric because of purpose, electronics, etc.):
 - 1) not symmetric: A talking to B does not imply B can talk to A,
 - 2) not transitive: A talking to B and B talking to C does not imply A can talk to C;
- e) nodes of a network are mobile, which causes the network topology to change and can cause intermittent link connectivity;
- f) mobile nodes are often power constrained because of reliance on batteries;
- g) the radio transmission spectrum is regulated.

4.3 BASIC CONCEPTS OF WIRELESS NETWORKS

4.3.1 RADIO AND OPTICAL COMMUNICATION

There are two basic technologies in use today for the deployment of wireless networks: RF waves and InfraRed (IR). Infrared transmission occurs at a wavelength of 850–900 nm. Both technologies can be used to set up an ad hoc network, e.g., for wireless nodes that dynamically join and leave a given wireless network.

Infrared technology uses diffuse light reflected at walls, furniture, etc., or directed light in a Line-Of-Sight (LOS) between the sender and the receiver. Senders can be simple Light Emitting Diodes (LEDs) or laser diodes, whereas photodiodes act as receivers.

Advantages of infrared technology:

- a) Senders and receivers, which are integrated into most mobile devices today, are simple and very cheap. PDAs, laptops, notebooks, mobile phones, etc., often have an Infrared Data Association (IrDA) interface. Version 1.0 of the IrDA standard specifies data rates of up to 115 Kbit/sec, while IrDA 1.1 defines higher data rates of 1.152 and 4.0 (and possibly up to 16.0) Mb/s.
- b) No licenses are needed for infrared transmission.
- c) Shielding is very simple with IR devices; because of their limited range, shielding is much less of an issue than with RF devices.
- d) Electrical devices do not interfere with infrared transmission.

- e) There are optical advantages in regards to security; it is possible to control direction of IR radiation.
- f) Laser communication technologies can reach several hundreds of Mb/s.

Disadvantages of infrared technology:

- a) Bandwidth utility is low compared to other LAN technologies.
- b) Infrared is quite easily shielded. Infrared transmission cannot penetrate walls or other obstacles.
- c) For good transmission quality and high data rates, direct LOS is typically required.
- d) There is much less flexibility for mobility as compared to RF.

Advantages of RF technology:

- a) There is long term experience with radio transmission for wide area networks (e.g., microwave links) and mobile cellular telephones.
- b) Radio transmission can cover larger areas and can penetrate (non-conductive) walls, furniture, plants, etc.
- c) RF does not require direct LOS for reliable communication transmission.
- d) Current RF-based products offer much higher transmission rates than infrared.

Disadvantages of RF technology:

- a) Shielding is not simple.
- b) RF transmission of sensitive and command/control data requires implementation of high level of data security and authentication, translating to complexity of system and higher overall cost in design/development/implementation/verification/integration and operation.
- c) RF transmission can interfere with other senders or sensitive electronics. Requirements must be in place for sensitive electronics to be shielded properly and appropriate signal suppression techniques or filtering should be required on RF systems in specific bands.
- d) Electrical devices can emit EMI, which can corrupt/destroy data transmitted via radio. EMI from unintentional emitters, i.e., non-antenna connected electronics, should be required to implement proper shielding/grounding/bonding to suppress unwanted/spurious emissions, to minimize interferences to intentional emitters/receivers.

The more popular WLAN technologies rely on radio instead of IR. The main reason for this is the shielding problems of infrared. WLANs should, for example, cover a whole spacecraft and not be confined to a single module where a LOS exists. Furthermore, many mobile

devices might need to communicate while in an IR-shielded enclosure (e.g., inside a crew member’s pocket) and thus cannot rely on infrared.

Being of lower frequency as compared to IR, the RF channel behaves significantly differently from that of IR. Radio transmission can typically penetrate walls and nonmetallic/nonconductive materials, providing both the advantage of greater coverage and the disadvantage of reduced security and increased co-channel interference. RF transmission is robust to fluorescent lights and outdoor operation, thus being highly advantageous for outdoor applications. Nevertheless, RF equipment is subject to increased co-channel interference, atmospheric, galactic and man-made noise. There are also other sources of noise that affect operation of RF devices, such as high current circuits and microwave ovens, making the RF bands a crowded part of the ElectroMagnetic (EM) spectrum. However, careful system design and use of technologies such as spread spectrum modulation can significantly reduce interference effects in most cases.

RF equipment is generally more expensive than IR. This can be attributed to the fact that sophisticated modulation and transmission technologies, such as spread spectrum techniques, are often employed. This means complex frequency or phase conversion circuits must be used, a fact that might make end products more expensive. However, the advances in fabrication of components promise even larger factors of integration and constantly lowering costs. Finally, as far as the WLAN area is concerned, RF technology has an additional advantage over IR because of the large installed base of RF-WLAN products and the adoption of RF technology in current WLAN standards.

4.3.2 RADIO FREQUENCY BANDS

As indicated in figure 4-1, radio waves occupy the lowest part of the electro-magnetic spectrum.

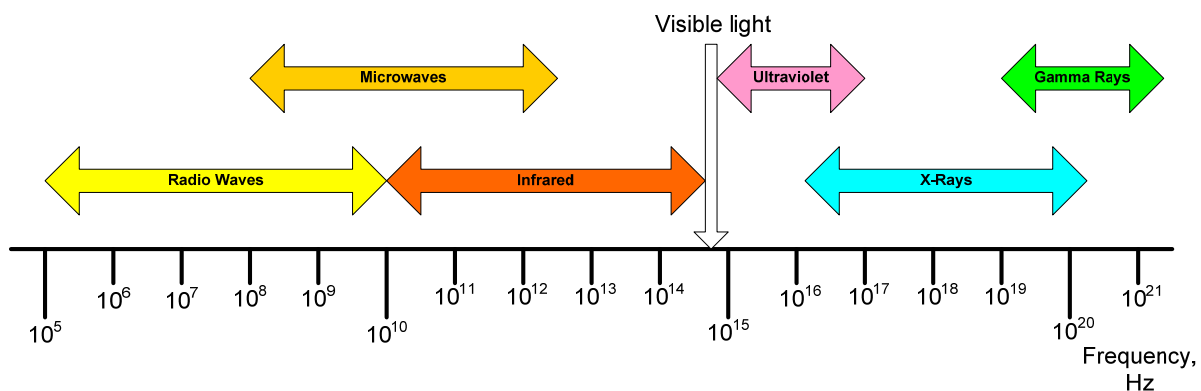


Figure 4-1: The Electromagnetic Spectrum

The EM spectrum is represented on a logarithmic scale so that frequency is increased by a factor of 10 at successive divisions across the horizontal scale. Bandwidth is the difference between the lower and upper cutoff frequencies of a communication band; thus higher

bandwidths can theoretically transport higher data rates (e.g., measured in bits per second, b/s). The bands above visible light are rarely used in wireless communication systems because the extremely high frequency waves are difficult to modulate (encode information). Table 4-1 summarizes common RF bands and typical applications.

Table 4-1: Common Radio Frequency Bands and Typical Applications

Frequency	Band Name	Applications
< 3 kHz	Extremely Low Frequency (ELF)	Submarine communications
3 kHz–30 kHz	Very Low Frequency (VLF)	Marine communications
30 kHz–300 kHz	Low Frequency (LF)	AM Radio
300 kHz–3 MHz	Medium Frequency (MF)	AM Radio
3 MHz–30 MHz	High Frequency (HF)	AM Radio
30 MHz–300 MHz	Very High Frequency (VHF)	FM Radio, TV
300 MHz–3 GHz	Ultra High Frequency (UHF)	TV, cellular, wireless systems
3 GHz–30 GHz	Super High Frequency (SHF)	Satellites
30 GHz–300 GHz	Extra High Frequency (EHF)	Satellites, radars

Different radio bands have different transmission properties. Attenuation is the reduction in amplitude of a signal; in the RF spectrum higher frequency waves typically have a shorter range of transmission because they are attenuated (blocked) more by obstacles than lower frequency waves. This is readily shown by the fact that any (non-transparent) wall will block light waves, while this is not necessarily true for RF waves. Since regulated frequency bands are assigned based on a percentage of their center frequency, lower frequency bands have less bandwidth than higher frequency bands; thus wireless networks typically operate in the higher RF frequency bands simply to enable faster data rates associated with higher bandwidth systems. The range of both low- and high-frequency RF transmission can be controlled via the radiated power of the signal; for wireless communications this is typically viewed as a benefit because it enables frequency reuse over large geographical areas (this frequency reuse is also known as Space Division Multiplexing [SDM]).

4.3.3 COEXISTENCE

RF coexistence mechanisms are used to optimize the spectral efficiency of different RF protocols operating in the same bandwidth and in the same general area. This issue has become particularly important with the widespread deployment of WLANs and WPANs operating in the same RF spectrum band. WLANs are used to access client and server devices typical of the Internet, whereas WPAN devices are used primarily in sensor networks or as a cable replacement technology. As such, both protocols are likely to be found in the same general area and could even be installed on the same computer. This scenario can be extended to space environments, where in a typical spacecraft or planetary habitat it will be commonplace for several wireless network protocols share bandwidth and be collocated in the same physical environment.

With the heightened awareness of co-existence between WLANs and WPANs, there is a significant effort by the IEEE wireless standards committee to consider the co-existence problem up front. This is true, for example, in current WLAN standards such as Wi-Fi (IEEE 802.11-2012) and WPAN standards such as Bluetooth (IEEE 802.15.1) and IEEE 802.15.4. The next generation of wireless networks and devices is expected to address this challenge to an even greater extent with advanced hardware for multipath mitigation technologies along with passive and active coexistence mechanisms.

4.3.4 TYPES AND TOPOLOGIES OF NETWORKS

Networks, both wired and wireless, can exhibit different physical topologies. For example, a wired LAN such as Ethernet will often be configured in a so-called bus topology, while a wireless LAN will often be configured in a star topology. Several different network topologies are illustrated in figure 4-2. In general, because of range limitations and mobility requirements, wireless networks are most often configured in star, mesh, or tree topologies.

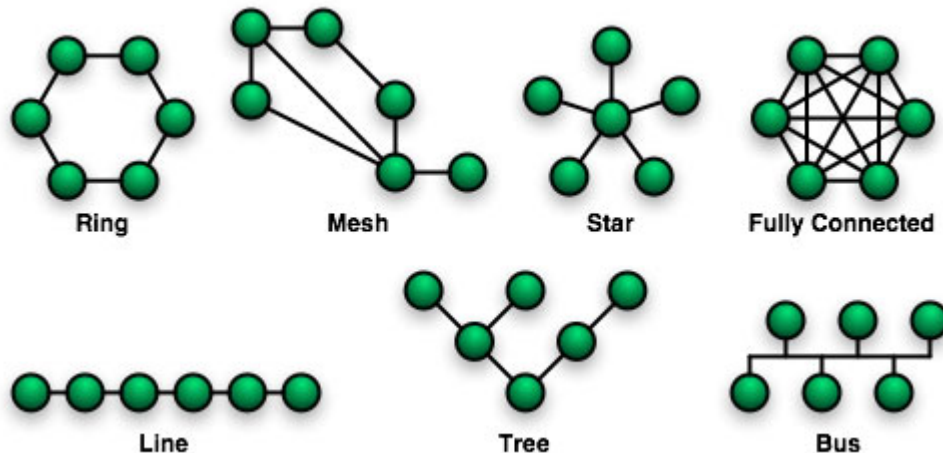


Figure 4-2: Different Network Topologies

When there are only two nodes in a network, the topology is referred to as a point-to-point network and is a simple example of a line topology. A point-to-multipoint network consists of a single wireless Base Station (BS) that communicates directly with one or more client Subscriber Stations (SS) in a star topology. The client subscriber stations are often free to roam within the radio range of the base station (sometimes referred to as an Access Point [AP]). The communication from the base station to the subscriber stations is termed downlink or forward link communications, while the communication in the reverse direction is termed uplink or reverse link communications.

Wireless point-to-point and point-to-multipoint topologies are single-hop, meaning that the data traverses only a single wireless transmission link. Mesh networks, on the other hand, can support data transport over multiple wireless links or hops in succession. Such networks are generically referred to as multi-hop networks. Mesh network protocols are necessarily

more complex than star topologies in order to enable the transmission of data across a potentially unknown number of hops from a source to a destination. The terrestrial Internet is the best example of a multi-hop mesh network, though typically only the last hop (the last mile in telecom vernacular) is wireless.

For situations in which the most appropriate wireless network topology cannot be determined a priori or where nodes are very mobile and network membership and connectivity can be expected to change in an unpredictable manner, so-called ad hoc networks are of interest. Ad hoc wireless networks are a special case of wireless networks that require no predetermined central administration. The wireless mobile nodes collaborate to form a mesh or fully connected topology. In the case of a mesh network, each node must be able to participate in the routing or forwarding of packets from a source to a destination. Ad hoc networks provide the capability for distributed (decentralized) operation, support dynamic topologies where roaming wireless nodes enter and leave the network in a random fashion, potentially make use of multi-hop packet routing, and may be power constrained if battery powered.

4.3.5 MOBILITY IN WIRELESS NETWORKS

4.3.5.1 General

Although it is possible to build simple wireless networks with a few nodes communicating one-to-one (a mesh), or via a star topology using a single central base station (BS), it is generally found that networks rapidly outgrow these topologies. In these more complex topologies, multiple network-connected BS nodes provide connectivity to wireless client nodes. In the case when the client nodes move around, and are thus mobile stations (MS), the resulting system is known as a mobile network.

With the advent of modern smart phones and other mobile devices as the primary form of user computing and communications, mobile network development is presently the single most advanced communications technology area, globally. Furthermore, mobile networks are rapidly growing to be the most deployed network technology. The need for modern mobility is driving fixed networks to now adopt mobile network technologies as mobile user requirements rapidly become dominant over fixed user requirements. These disruption-tolerant network approaches will be increasingly important for spaceflight system requirements. Furthermore, the evolution of standard computing and communications platforms operating within a mobile networking environment can be expected to generate new requirements on onboard data systems and corresponding interfaces.

4.3.5.2 Basic Mobility Requirements

Mobility creates significant technology requirements. The first is the need for an MS to dynamically hand-off (HO) between each BS in the network while preserving connectivity; in particular, without disrupting TCP/IP socket connections. This requires that the MS can identify which BS to HO to before a connection to the presently (camped on, in mobile network parlance) BS fails, and then can maintain packet flow to each BS as it transfers

connection, generally with the aid of each BS involved in the HO process. The BS network can be spread across a large geographic area, with potentially different network addresses in each BS service area (called a cell). Thus a mobile network generally needs to provide a central mobility anchor, a network service that forms a component of the Core Network (CN) providing mobile network-wide apparent fixed network addresses for each MS. The existence of a mobility anchor and a CN is a general differentiator between a true mobile network and a fixed network with multiple BS nodes; in the latter case, there is generally no expectation of continuity of network socket connections as a node transitions from using one BS to another, unlike the mobile network HO mechanism.

A second requirement is that there is a backhaul network connecting the BS nodes to each other and the CN. Generally, in addition to the mobility anchor service, there are other services providing management of the BS and MS nodes, in particular the details of the HO process, and to further nodes that provide for routing of network connections from the mobility anchor to other networks external to the CN. Modern backhaul networks operate at extremely high-speeds and, with the capabilities of the CN, provide for extremely high quality of service (QoS). Indeed, a classic component of a mobile network is a dedicated end-to-end QoS mechanism on a per-application / per-network socket (called a bearer) basis. This allows modern wireless mobile networks to provide for QoS levels not found in the majority of previous fixed shared wireless network technologies. The result is that these emerging mobile network technologies can provide for very high-levels of mission and life-critical QoS, and services based on these capabilities form an integral part of the modern networking world.

4.3.5.3 Dynamic Spectrum Management

Unlike a fixed node, the location of a MS and the requirements of each camped-on BS constantly change. Thus the geographic network throughput constantly changes, and thus the required RF spectrum bandwidths in each location are also dynamic. In addition, each MS and corresponding connection to and from the camped-on BS can RF interfere with any other MS and BS in the network, and the corresponding RF interference is highly dynamic. Finally, creation and destruction of service bearers (per-application network connections) increase and decrease MS, and hence BS, RF bandwidth requirements on timescales of a fractions of a second. Thus spectrum cannot be planned on a fixed basis, and mobile networks must dynamically manage spectrum allocation, and indeed time distribution of data packet transfers, in real time, often with many hundreds of spectrum management decisions being made per second in an average mobile network BS and corresponding CN infrastructure. Modern mobile networks self-manage spectrum in many dynamically allocated channels and time-slots in generally roughly assigned spectrum blocks.

4.3.5.4 The Mobile Tsunami and Densification

Mobile network capacity requirements are growing at approximately a factor of one thousand per decade. This so-called Mobile Tsunami leads to the 1000X problem, in which given deployed networks must rapidly grow to support new capacity and, in particular, the total data throughput available per given amount of service area. The primary determining factors

of the capacity per area are the amount of available RF spectrum, the capacity of the underlying radio technique to support the maximum number of bits per second out of a hertz of spectrum, and the number of times spectrum can be re-used, geographically, across the entire mobile network. It is the latter factor that supports the major capability for mobile networks to grow; by shrinking the physical network coverage of a BS, and hence increasing the density of the geographic distribution of BS nodes, spectrum is re-used. In modern cell-phone mobile networks, this process of densification has resulted in BS cells moving from having a diameter of several km to now less than 1 km in the last decade. In the future decade, the 1000X problem will require a density corresponding to separations of less than 100 meters and, in many cases, less than 10 meters. These latest mobile network types are called small cell and femtocell distributions, respectively, with the large-scale (km or above diameter) mobile network BS infrastructure being known as a macrocell deployment. This has resulted in the creation of extremely small, and low mass, power, and cost, femtocell BS technologies, often smaller than many consumer IEEE 802.11-type wireless access points. These emerging, but already available, technologies provide a full, extremely miniaturized, 'cell tower' in a low-cost box often significantly smaller than a paperback book, with all the corresponding spectrum and client management capabilities.

In addition to densification, the Mobile Tsunami is addressed via increasingly advanced radio technologies to boost capacity on a per-hertz basis, and to open up new areas of RF spectrum. Thus the Mobile Tsunami is driving mobile networks to provide the very latest and most advanced total network solutions for total data uplink and downlink capacity.

4.3.5.5 Heterogeneous Networking

The need for densification and the drive for increased spectrum also lead to the need to use many different radio access technology (RAT) solutions operating together for next-generation mobile networks. These heterogeneous networks, or HetNets, are expected to become a primary form of mobile network in the future. A HetNet allows the use of the most advanced RAT solutions inter-operating with advanced CN-controlled mobility control services, maximizing performance, in particular capacity via densification.

4.3.5.6 Self-Optimization and Self-Healing

HetNets and the emergence of small cell and femtocell solutions to address the Mobile Tsunami both produce a significant problem for optimizing a mobile network infrastructure; although spectrum allocation is dynamic, the actual rules for that allocation, plus other parameters such as radio power levels on a per-channel basis, must themselves change dynamically as BS nodes are added, removed, and moved in the network. Thus mobile network technologies are rapidly becoming Self-Optimized Networks (SONs), in which the entire network infrastructure can self-configure, down to the tiniest low-cost femtocell, for best performance. In addition to producing highly flexible networks, the emergence of large-scale SON deployment will result in networks that also increasingly self-heal in a fraction of a second of loss of a local BS node.

4.3.5.7 Functions, Virtualization, and Software-Defined Networking

It is important to note that a modern mobile network technology is as much defined by the network functions (providing key operational services) in the CN as the RATs used on the air interface of the network. In networks supporting many tens of millions of users, these functions are often hosted in many specialized, dedicated, servers in massive data centers. Evolution of the network generally is defined by evolution of the CN. However, there is now an emergence of the concept of Network Function Virtualization (NFV), in which these services are hosted in virtual machines running on a single computing platform. This provides for rapid implementation of improved and new functions. In addition to flexibility, NFV allows for easier scaling of solutions to support a comparatively low number of MS nodes, more appropriate to spaceflight missions, and, indeed, allows for a complete CN to be implemented in a single small laptop-scale server. Other non-CN functions, such as network routers and switches, can also be virtualized, in an example of Software-Defined Networking (SDN), to further reduce size and complexity of a mobile network implementation.

4.3.5.8 Fixed-Mobile Convergence and Next-Generation Networking

The growth of mobile networking beyond the deployment of fixed networking has resulted in a rapid evolution away from old Internet-style networking towards the high-criticality technologies and solutions found on advanced mobile networks; users will increasingly expect to have their mobile and fixed networks behave in a similar function, with no difference in access capabilities between the two. This has resulted in a transition of traditional fixed networking concepts to embrace mobile techniques, in what has become known as Fixed-Mobile Convergence (FMC). In the FMC approach, the fixed networks use the same form of CN as the mobile networks. In addition, the client systems use the same software applications and network protocols as are used on systems using MS nodes. Thus FMC is part of the next-generation networking model in which present-style internetworking approaches are being replaced by new approaches as networks rapidly evolve towards rapid dynamic changes in topology and node configuration during application sessions. Mobile network-driven FMC therefore produces an environment that is critical to understand for the spaceflight application area, because of changes in even conventional fixed networking solutions, at the application and protocol level, which can be expected to cause corresponding rapid evolution in support requirements for onboard systems.

4.3.6 RF PROPAGATION BASICS

4.3.6.1 Free Space Loss

Compared to wired channels, wireless channels are less directive in transmission of energy between two points. Radiated transmissions lose signal energy through multiple means, including absorption, spreading, and reflection. The Friis Transmission equation provides a commonly used relationship for the RF power transmitted and received between two antennas in an idealized free space environment; that is, an environment with no scattering objects or material losses outside of the antennas. Although it is idealized due to this

assumption, in some links, particularly some space-based links, this assumption can result in reasonable first-order performance estimates. In other cases, it provides an upper bound of sorts on the expected performance. One of the more common forms of the Friis Transmission equations is:

$$P_R = P_T G_T A_R / (4\pi d^2) ,$$

in which P_R and P_T are the received and transmitted power, respectively, G_T is the gain of the transmit antenna, d is the distance between the two antennas, and A_R is the effective aperture area of the receive antenna. Sometimes the Friis Transmission equation is expressed using gain for the receive antenna figure of merit. In this case, the equation appears as

$$P_R = P_T G_T / (4\pi d^2) [\lambda^2 G_R / (4\pi)] = P_T G_T G_R \lambda^2 / (4\pi d)^2 .$$

In this case, the term $(4\pi \lambda / d)^2$ is sometimes referred to as ‘free-space loss’. This term can be misleading, however, since the appearance of wavelength in the equation arises because of the assumption that the receive antenna gain, as opposed to receive antenna effective area, is held fixed. In lossless free-space propagation, as modeled in figure 4-3, the path loss is not frequency dependent.

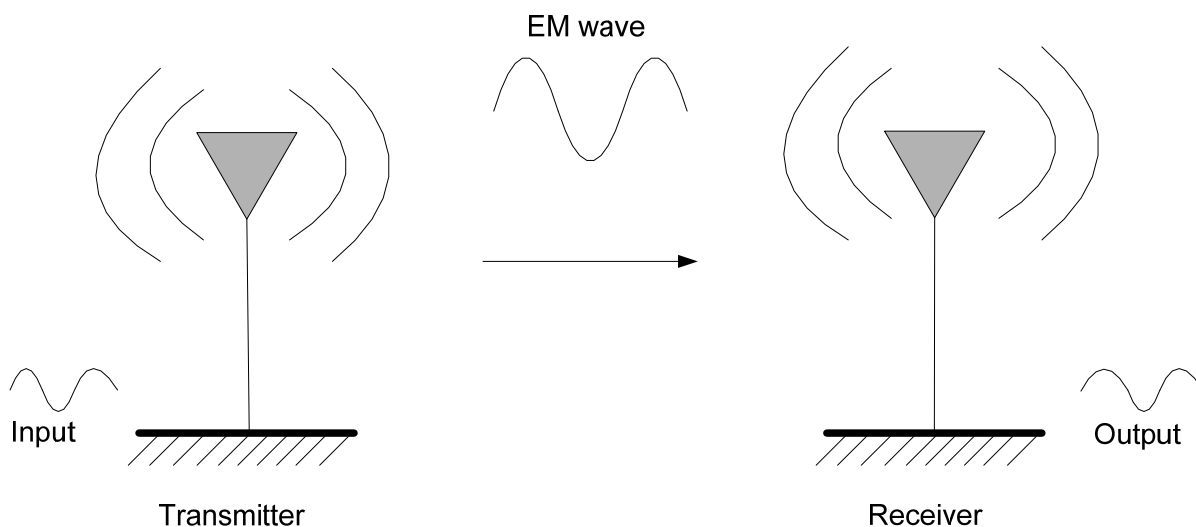


Figure 4-3: Free Space Path Loss (Attenuation) of a Signal

One key insight from the Friis Transmission equation is that the power at the receiver P_R decreases by the factor $1/d^2$ in a free space environment, for example:

$$P_R \propto 1/d^2$$

Examples where a free space loss model might be applied include transmission between two vehicles in orbit or between a satellite and a ground station on the moon, where, in both cases, it is assumed that none of the structures introduce reflections.

To account for path loss in more complicated environments, more sophisticated models are employed. For example, for transmission over an idealized flat ground plane, because of ground reflections, the receive power falls off more rapidly, and as d gets large, the receive power varies as:

$$P_R \propto \frac{h_t^2 h_r^2}{d^4}$$

where h_t and h_r are the transmit and receive antenna heights, respectively, above the ground as shown in figure 4-4.

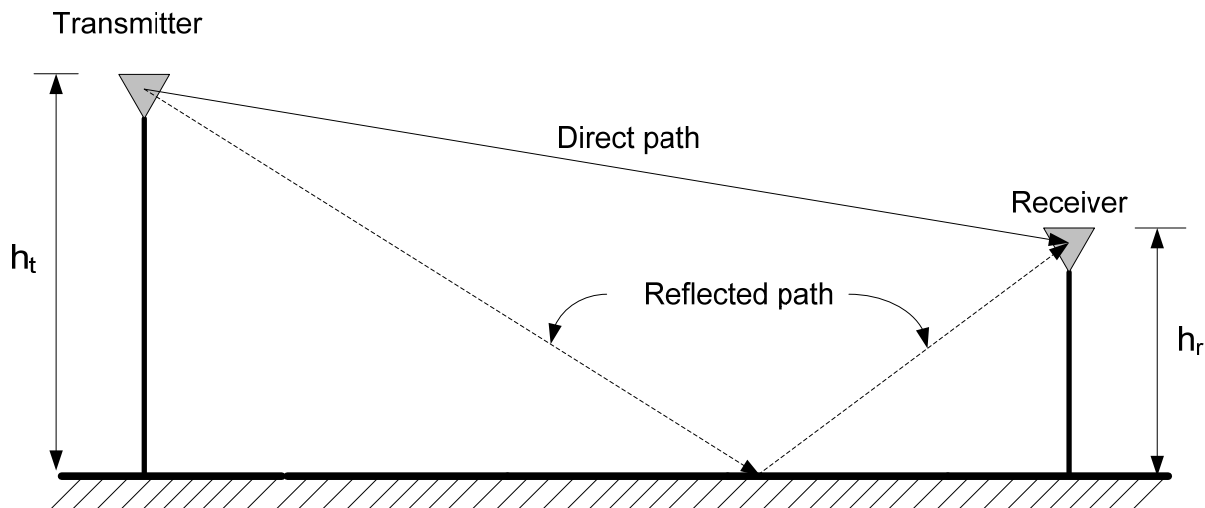


Figure 4-4: Two-Ray Ground Model (Attenuation) of a Signal

The results of the free space and ground models can be represented in a combined fashion as:

$$P_R = K/d^e,$$

where e is termed the path loss exponent (see reference [11]) and K is a proportionality constant. For free space, $e = 2$, and for the ground model, with large d , $e = 4$.

Additional environmental complexities often require still more sophisticated models. Such complexities might include curvature of the ground (e.g., a planet), atmospheric attenuation, and scattering obstacles. Sufficiently accurate propagation modeling might require the so-called asymptotic methods (e.g., the Geometric Theory of Diffraction), the so-called ‘full wave’ methods, or hybridizations between asymptotic and full-wave methods.

4.3.6.2 RF Propagation within a Cavity

Within a closed metallic cavity, free-space and surface propagation models are not applicable. Since most spacecraft resemble one or more conductive boundary cavities, this environment is of considerable importance for space applications of wireless technologies. The behaviors of the electromagnetic fields are dependent upon the dimensions of the structure, relative to the wavelengths of interest, the furnishings of the environment, and the material characteristics of the structure and furnishings. Typically, the structural dimensions presented by crewed spacecraft are sufficiently large relative to wavelengths commonly used in wireless applications (i.e., frequencies at UHF or higher) that the interior essentially constitutes a multi-moded, or overmoded, cavity. Smaller, uncrewed spacecraft might resemble either a single mode cavity or a cavity below cutoff frequency, even at UHF frequencies.

In overmoded cavities, the field structures can be quite complex, particularly if the quality factor, or 'Q' of the cavity, is high, implying that the constituent materials tend not to be considerably lossy. Moreover, the spacecraft environment can be considerably dynamic when crewed. In addition to the potential presence of human bodies (which are typically very lossy), furnishings in the environment can be rearranged. Thus designers cannot depend on a single particular field structure within the spacecraft. Because of the typically rich scattering environment in overmoded cavities, multipath can result in significant field nulls. Hence, multiple-antenna communication techniques, as discussed below, should be considered. To illustrate this, the insertion loss (i.e., S₂₁ scattering parameter measurement) between two antennas in a lunar habitat mockup was measured over a range of frequencies from 2.44 to 2.5 GHz. The results shown in figure 4-5 indicate very deep nulls arising from structurally induced multipath.

4.3.6.3 Noise and Interference

All wireless communication systems are subject to performance degradation caused by unknown signals superimposed on the signal of interest. Such intrusive additive RF signals are generally classified as either *noise* or *interference*. Although the distinction is somewhat arbitrary, the term noise usually refers to signals that are well characterized as random processes and do not originate from discrete, localized sources. Signals such as thermal background radiation and the thermal noise in electronic circuits fall into this category. The term interference, on the other hand, usually refers to signals with more deterministic structure that originate from discrete, localized, and often identifiable sources. Signals such as narrowband interference from electric appliances and both narrowband and broadband interference from other wireless communication systems fall into this category. Interference can sometimes be mitigated to a great extent by careful selection of frequency bands, shielding, or directive antennas, while the effects of noise are generally much more difficult to isolate and remove.

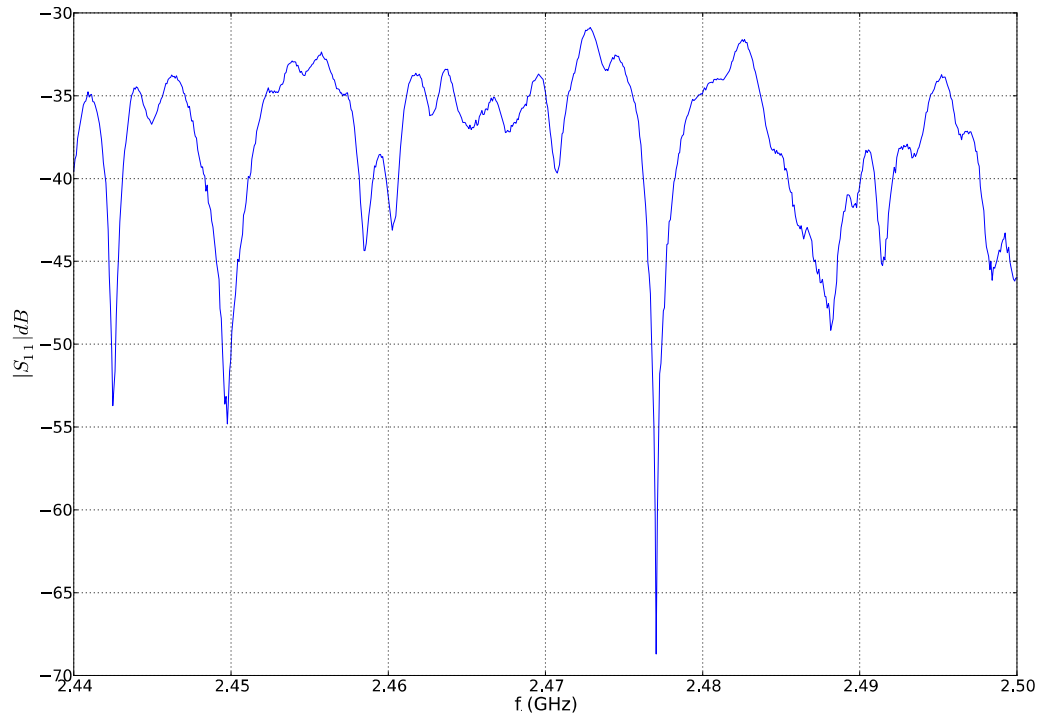


Figure 4-5: Transmission Loss Measurement in a Lunar Habitat Mockup

4.3.6.4 Brief Introduction to Antennas

An antenna is a structure that couples between guided and unguided electromagnetic waves. Performance factors include directivity, efficiency, and polarization. All of these are functions of frequency, and the directivity and polarization are also functionally dependent upon spatial angle. Together, the directivity and efficiency determine the gain, which is typically referenced with respect to an idealized isotropic radiator. Occasionally, gain is referenced to a particular standard antenna, such as a half-wave dipole. The size, shape, height, pattern, and material of the antenna provide degrees of freedom from which all of these performance factors can be affected.

As indicated in the Friis transmission equations in 4.3.6.1, antennas are a critical part of any link. The effective aperture of the antenna determines how directive the antenna is, or the degree to which the radiation is focused. Larger effective apertures provide greater directivity. Of course, more directive antenna patterns require pointing, either electrical or mechanical, when one or more nodes are not static.

Often, in wireless systems, small antennas are highly desirable from form or fit perspectives, assuming the effective aperture is at least sufficient to complete the link. It should be noted, however, that there are fundamental physical relationships that bound antenna efficiency as the antenna volume is reduced. These limitations are particularly relevant with antenna sizes on the order of $\lambda/8$ or $\lambda/16$, and smaller.

Recent technology advances have utilized multiple antennas on one or both sides of a communication link. Such multi-antenna technologies provide means for overcoming many issues associated with wireless communications. Such limitations included multipath fading, limited signal-to-noise ratio, multiplexing, jamming, and interference.

4.3.6.5 Multiple Antenna Communication Links

In general, wireless communication techniques can be divided into four different categories depending on the number of antenna nodes at the transmitter and receiver, as follows:¹

- a) *Single-Input, Single-Output (SISO)*. The simplest scenario, with one antenna at both the transmitter and receiver. SISO links generally have limited antenna gain and often suffer from signal attenuation due to multipath propagation, which is called multipath fading. Simple narrowband Additive White Gaussian Noise (AWGN) SISO links with transmitter power of P watts, bandwidth of B Hz, and noise Power Spectral Density (PSD) of N_0 watts per Hz at the receiver have an ergodic capacity of approximately $\log_2(1 + P/(BN_0))$ bits per second per Hz (b/s/Hz).
- b) *Single-Input, Multiple-Output (SIMO)*. SIMO is generally regarded as the next level of complexity, with one antenna at the transmitter and multiple antennas at the receiver. The multiple antenna nodes at the receiver amplify the signal by increasing the size of the antenna aperture (array gain) and decrease susceptibility to multipath fading by increasing the spatial diversity of the link (diversity gain). For narrowband SIMO links, the array gain and diversity gain are achieved simultaneously by coherently combining signals at the receiver, which requires knowledge of the channel (e.g., direction of arrival or multipath gains) only at the receiver. Such knowledge can be obtained adaptively with no cooperation from the transmitter. If the channel is a free-space channel, such coherent combining at the receiver is called receive beamforming. In a more general context, such as communication over multipath channels, this approach is called simply receiver combining. Narrowband AWGN SIMO links with M nodes at the receiver, transmitter power of P watts, bandwidth of B Hz, and noise PSD of N_0 watts per Hz at each receiver node have a capacity of approximately $\log_2(1 + MP/(BN_0))$ b/s/Hz.
- c) *Multiple-Input, Single-Output (MISO)*. Slightly more difficult to exploit than SIMO links, MISO links have multiple antennas at the transmitter and a single antenna at the receiver. The multiple nodes at the transmitter again provide both array gain to amplify the signal and diversity gain to combat multipath fading. For narrow-band MISO links, the array gain and diversity gain can be achieved simultaneously by precoding signals at the transmitter in order that they combine coherently at the receiver. On free-space channels, this is called transmit beamforming, and in the more general context it is called simply transmitter precoding. Alternatively, diversity gain alone (with no associated array gain) can be achieved by using space-time coding at the transmitter. Transmitter precoding requires knowledge of the channel (e.g., direction of receiver or multipath delays) at the transmitter while space-time coding requires no such

¹ Source: reference [54]

knowledge. Channel knowledge can generally only be obtained at the transmitter with some type of feedback from the receiver to the transmitter. Narrowband AWGN MISO links with N nodes at the transmitter, total transmitter power of P watts (from all nodes combined), bandwidth of B Hz, and noise PSD of N_0 watts per Hz at the receiver also have a capacity of approximately $\log_2(1 + NP/(BN_0))$ b/s/Hz when transmitter precoding is employed. If space-time coding is employed at the transmitter, the capacity drops to approximately $\log_2(1 + P/BN_0)$ b/s/Hz.

d) *Multiple-Input, Multiple-Output (MIMO)*. MIMO is the most complex scenario, with multiple antennas at both the transmitter and receiver, but it also offers the most potential performance gain. MIMO links not only provide both array gain and diversity gain, but also have the potential to provide multiplexing gain, which means that multiple independent data streams can be transmitted simultaneously across the link, as if the individual channels between different transmitter/receiver antenna pairs did not interfere with each other. The relationship between array gain, diversity gain, and multiplexing gain in a true MIMO context is discussed below.

1) *Array Gain and Diversity Gain*. For narrow-band MIMO links, array gain and diversity gain can be achieved simultaneously (with no associated multiplexing gain) by using receiver combining and transmitter precoding simultaneously. Alternatively, if no channel knowledge is available at the transmitter, space-time coding can be used at the transmitter together with receiver combining to provide less array gain maximum diversity gain. Narrowband AWGN MIMO links with N nodes at the transmitter and M nodes at the receiver, total transmitter power of P watts, bandwidth of B Hz, and noise PSD of N_0 watts per Hz at each receiver node have a capacity of approximately $\log_2(1 + NMP/BN_0)$ b/s/Hz when both transmitter precoding and receiver combining are employed. If space-time coding is employed at the transmitter, the capacity drops to approximately $\log_2(1 + MP/BN_0)$ b/s/Hz.

2) *Multiplexing Gain*. The availability of multiplexing gain on MIMO links depends on the geometry and/or statistical structure of the channel. In particular, the frequency response of the channels between different transmitter/receiver antenna pairs must be well modeled as statistically uncorrelated. On such channels, multiplexing gain can be achieved in several ways, but most efficiently by communicating across the eigenmodes of the channel. On free-space channels with widely separated receiver nodes, this is called spatial beamforming, and in the more general case, it is called simply spatial multiplexing. The special case of communicating across the eigenmodes of a channel with N antennas at the transmitter and N antennas at the receiver is called MIMO beamforming. Spatial multiplexing requires full channel knowledge at both transmitter and receiver. Under optimal conditions, spatial multiplexing on a narrowband AWGN link with N nodes at the transmitter and $M > N$ nodes at the receiver, total transmitter power of P watts, bandwidth of B Hz, and noise PSD of N_0 watts per Hz at each receiver node can achieve a capacity of approximately $N \log_2(1 + MP/(NB_0))$ b/s/Hz.

4.3.6.6 Fading: Multipath and Shadowing

In addition to path-loss effects, there are two other principal sources of signal attenuation during propagation. Both of these are generally classified as fading losses, with one being referred to as large-scale fading or shadowing and the other being referred to as small-scale or multipath fading. The distinction between path-loss effects, which can be caused by multipath, atmosphere, and/or blockage (shadowing) due to obstacles, and fading is that fading is modeled as random behavior that is not predictable in any deterministic sense while path loss follows some fairly simple rule, such as geometrical path loss or even exponential path loss. Small scale or multipath fading is the random behavior caused by rapidly varying carrier phase across multiple propagation paths, and large-scale or shadow fading is essentially a model for the errors between the predicted path-loss behavior and the actual average power loss over distance. For example, if the path-loss model is geometrical with some path-loss exponent, then the errors between a linear least-squares fit to the power loss (in dB) and the actual average power loss over distance are often approximately normally distributed, which leads to so-called log-normal shadow fading behavior. The cumulative effect of deterministic path loss together with both types of fading is generally modeled as the product of the random attenuation due to shadowing, in which the deterministic path loss is incorporated as a mean-value component, and the random attenuation due to multipath, which can have either a zero or non-zero mean component depending the existence of a LOS component in the signal path.

In other words, the propagation channel is modeled as the cascade of two random linear channels. The shadow-fading channel models amplitude only (so it is real-valued) and is dominated by path-loss and shadowing effects. It is characterized by a fairly large, non-zero mean (deterministic) behavior, a relatively small variance (random variations around the mean), and relatively slow variations over time and space. A common model is log-normal shadowing, but many models are in common usage (see references [12] and [13]). The multipath-fading channel models both amplitude and phase (so it is complex valued) and is dominated by the effects of carrier phase variation across multiple propagation paths. It is characterized by possibly large random fluctuations around a possibly zero mean behavior and relatively rapid variations over time and space. A common model is complex Gaussian, which for narrowband channels corresponds to either a Rayleigh envelope distribution if the mean is zero or a Ricean envelope distribution if the mean is non-zero. Many other models for multipath fading are in common usage (see references [12] and [13]).

On most wireless channels, by far the more problematic fading behavior, which frequently causes more performance degradation than noise, interference, and shadowing combined, is multipath fading. To better understand this phenomenon, consider figures 4-6 and 4-7. Figure 4-6 illustrates a fairly common and yet complex propagation environment and figure 4-7 illustrates the peaks and nulls in a standing wave pattern resulting from an RF transmission reflected off a flat surface. The distance between the signal peaks and nulls in figure 4-7 is $\lambda/4$ (where λ is the carrier wavelength) along a line segment from the transmitter to a point perpendicular to the reflecting surface. With the superposition of both direct-path arrivals and multiple such reflections the signal amplitude and phase become a complex function of space in the environment. When objects in the environment and or the transmitter/receiver are in motion, the signal amplitude and phase also become a function of *time*. Furthermore, for typical

wireless frequencies such as the 2.4 GHz band, the signal amplitude and phase fluctuations can occur very rapidly in both time and space because the carrier wavelength is very short. Hence, the overall effect is complex, very unpredictable, and sometimes quite dramatic.

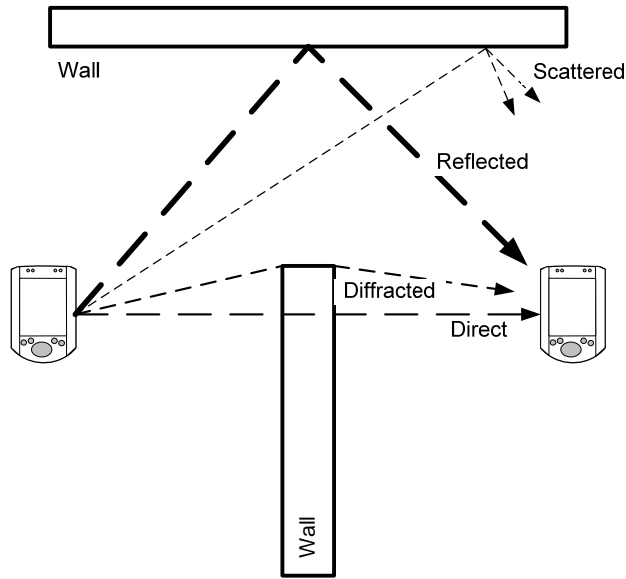


Figure 4-6: RF Transmission Wave Path Classes²

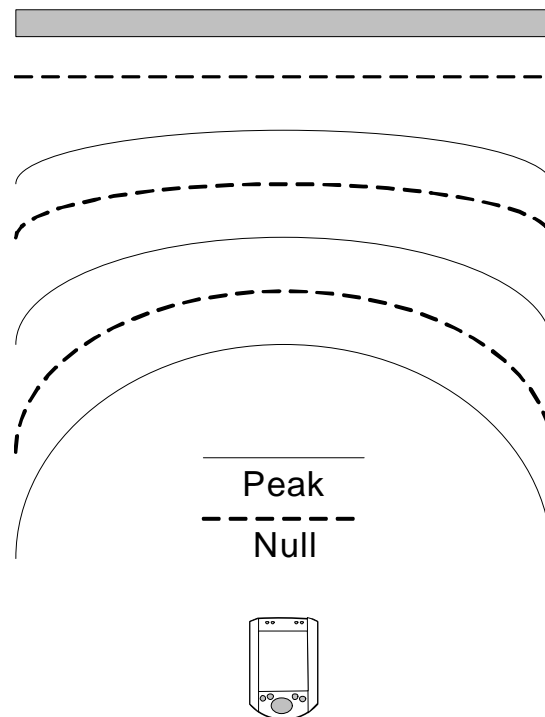


Figure 4-7: RF Standing Wave Pattern from a Reflecting Wall²

² Source: reference [55].

Real-world RF communication is generally more complicated than a single Line-of-Sight link or a purely Rayleigh Channel; generally a link is composed of a so-called averaged *fixed* component and a temporally and fine-scaled spatially varying component, formed from scattered and reflected signals, called the *variable* component. For destination exploration surface communications or internal / proximity spacecraft operations, an important concept from RF physics theory is the Ricean K -factor, defined by the following:

$$K = \text{Received Power in Direct Component} / \text{Received Power in Variable Component}$$

Hence $K = 0$ is the NLOS Rayleigh Channel case, and $K \rightarrow \infty$ corresponds to the pure, away-from-surface, LOS case or interplanetary / non-proximity link case. Generally, the lower the value of K , the more advanced and complex is the technology required to implement a high-speed communications link. It is also important to realize that K is a property of both the communication environment, the location of the endpoints of the link, and the corresponding antennas used.

A general Ricean Channel will have a (course-grained spatial) mean time-averaged fading power level that scales as $1/d^\gamma$ where γ is between 2 and 4. γ is generally close to 4 when K is close to zero, and close to 2 when K is large. In general, a full multipath channel is a sum of Ricean channels with differing delays (a multi-tap model), corresponding to reflections off major objects such as hills.

As K approaches unity from above, even though $K > 0$, the channel behavior moves to one that is very similar to Rayleigh channel physics. A so-called Ricean-Rayleigh Break usually accompanies this regime transition, in which the $1/d^2$ fade transitions rapidly to $1/d^4$ and spatial and temporal variations are similar in magnitude to the Rayleigh case.

In the high-multipath and scatter case, with low K , there are strong variations of total received signal strength with position and time, even in an LOS situation. Fades can be as high as 40 dB (reductions in signal power and possible data rates by a factor of 10,000) below that which would be expected from usual free path signal estimates.

It is important to realize that low- K environments are not only caused by reflections from large objects such as exploration-target hills or large spacecraft structure, but also by the general scattering off ground-clutter or irregular surfaces; the direct signal transmitted and received between two points corresponds to a small fraction of the radiation emitted from a transmitter, received from a small angle around the line of sight between transmitter and receiver. However, as the range between transmitter and receiver increases, the available surface area for surface scatter and multi-path increases, allowing increasingly more of the remaining indirect signal from the transmitter to reach the receiver. Thus, eventually, the indirect, variable, component of the signal comes to dominate over the fixed, direct, component. This will even occur for long-range communication over an open planar, but irregular, surface with perfect line of sight.

It is also important to realize that K , in many experiments by researchers in a wide range of environments on Earth, has been shown to have high variation, with a consistent 8 dB (over six-fold) standard deviation, as a function of position and time. Thus models and predictions involving multi-tapped Ricean channels are primarily statistical in nature.

The variation of propagation with frequency is a point of critical importance, as mentioned in Subsection 4.3.6.2 for the case of propagation inside a cavity; in a general high-multipath channel, received signal strength varies rapidly with frequency and, furthermore, this Frequency-Selective Fading (FSF) varies rapidly with time.

In particular, if the bandwidth of the transmitted signal is larger than the frequency scale (the coherence bandwidth) of the variations in fading behavior, frequency-selective fading (FSF) results, in which signals are highly distorted during propagation through the channel. Importantly, in this regime, the total power of the received signal will not suffer major fluctuations, because of FSF removing a fractionally small component of the spectrum of the signal. Indeed, if the structure of the waveform used for the transmitted signal is such that the spectral distortion can be tolerated, FSF may not disturb communication. However, time-varying fading also has an extreme effect on narrowband communications; that is, a low-bandwidth signal such as that used for simpler voice or low data-rate telemetry and telecommand systems, will rapidly vary in received signal strength, often becoming totally unusable for short periods of time. The effect, often called *picket fencing*, is a well-known problem in urban, high-multipath environments, and will also be seen in the exploration, internal, and proximity communications environment. FSF is the primary limiter to narrowband communications approaches to critical command and communications, especially in human missions and teleoperated robotic missions, in which unstable communications may have life-critical or mission-critical impacts.

4.3.6.7 Doppler Spread and Problems with Mobility

Another critical problem for wireless operations in spaceflight environments is the physics involved in communicating with mobile mission elements or in cases in which any component of the environment is in motion; In a Ricean environment, the received signal is comprised of many distinct signal components, each taking a different route from transmitter to receiver. When the transmitter, receiver, or any reflecting, or even atmospheric scattering, component is in motion, the corresponding signal components thus suffer different Doppler shifts, causing complex phase relationships, and hence constructive and destructive signal effects, to change rapidly. This *Doppler spread* results in strong variations of signal strength in time, in addition to that just caused by change in position. In modern high-speed communication systems, where wavelengths are of order 10 cm or less, even small motions cause extreme variations in signal strength in high-multipath environments.

Mobile high-speed wireless communications is thus generally considered a magnitude greater problem than static or nomadic communications. Indeed, Doppler spread, combined with the low K-factor in ground-level or other near-reflector environments, result in a factor of ten or more reduction in maximum possible data rate for mobile surface-based, internal, or proximity systems, compared to static and nomadic systems. However, even in a primarily static internal spacecraft environment, moving components such as operating fans and floating objects may cause considerable Doppler spread implications for high-rate communications, even when the transmitter and receiver are not in motion relative to each other or the gross environment. Similarly, high-speed wireless communications for relatively slow spacecraft and EVA crew motion external to a large spacecraft, such as ISS, are highly affected by Doppler spread.

4.3.6.8 Delay Spread and Inter-symbol Interference Effect

The rapid deep-fading behavior of low K -factor environments is an impediment to availability of communications systems, but a different effect in multipath environments is an even more serious impediment to the implementation and performance capabilities of high-speed communication systems.

The performance problem is caused by the delay spread of the communication channel; Signals for a given communication impulse period, a symbol, interfere with other symbols at later times, because of the overlapping range of arrival times of a symbols from the transmitter to the receiver. This effect is called Inter-symbol Interference (ISI) and corresponds to the data component of the *modulated* RF from the transmitter signal interfering with itself on the way to the receiver. This is different from multipath fade, which corresponds to the *carrier* component of the transmitter signal interfering with itself.

ISI is the largest commercial and industrial technical problem for high-speed RF communication, and is considered by the advanced communication industry to be the primary barrier to developing full and true fourth-generation (4G) communication technologies; The problem is a source of major R&D in communications technology development in recent years, and for communications sector R&D planned for future decades.

The physics of the problem is easy to understand; because of scatter and multipath, a concentrated (delta-function) impulse from a transmitter will arrive at the receiver spread over a period of time τ , the instantaneous delay spread, which generally varies in time and position with log-normal statistics. For a given transmit-receive path, there will be a time-averaged mean delay, which can be considered to be a measure of the ISI problem at that path. A communication symbol will have a given symbol time T_s , and generally considerably more than twice the delay spread time is required between symbols for them to be decoded by a receiver because of the smearing of the symbol caused by the delay spread. In other words, communication is only possible, on average, if $T_s \gg 2(\text{average } \tau)$. If R_s is defined as the symbol rate of the communication system (number of symbols per second), $R_s = 1/T_s$, the following important well-known result for that communication system is functional:

$$R_s \ll \frac{1}{2(\text{average } \tau)}$$

Thus ISI controls the maximum average symbol rate possible for a communication path. Correspondingly, instantaneous τ controls the instantaneous symbol rate possible. High delay spread, and hence high multipath, leads to low maximum symbol rates.

Generally a given communications technology will be able to produce a given mean data throughput rate D (actual true mean throughput, and not just raw data rate) in a given proportion to the symbol rate. The following can be defined:

$$N = D/R_s,$$

where N will generally be higher for more advanced wireless communication technologies. Indeed, the computational power, and thus general cost and complexity, to implement a given value of N are proportional to N . Thus the maximum ISI-limited throughput rate on a communication path is limited by

$$D \ll \frac{N}{2(\text{average } \tau)}$$

which connects the required performance D to the communications environment and the technology being used for the infrastructure. The average value of τ generally increases with the distance d between transmitter and receiver, but can be near-constant in sealed-cavity environments such as those internal to a spacecraft.

4.3.7 OPTICAL PROPAGATION BASICS

4.3.7.1 Basic Channel Structure

In telecommunications, Free Space Optics (FSO) is an optical communication technology that uses light propagating in free space to transmit data between two points. Most present-day optical channels are termed intensity modulated, direct detection channels.

Wireless optical links consist in modulating the instantaneous optical intensity, $I(t)$, in response to an electrical input signal, $x(t)$. Systems encode the signal as a sequence of light pulses in a binary form. This is called On-Off Keying (OOK) modulation. A Light-Emitting Diode (LED) or a Laser Diode (LD) is in charge of doing the electro-optical conversion process. These emitters usually operate in the 850–950 nm wavelength band.

An output electrical photocurrent, $y(t)$, proportional to the irradiance at the receiver, is produced by a silicon photodiode. The photodiode detector is said to perform direct-detection of the incident optical intensity signal.

4.3.7.2 Channel Topologies

It is important to differentiate a point-to-point link, with direct LOS, from a diffuse one, in which direct LOS may or may not exist. When there is a direct path between a transmitter and a receiver, the wireless optical link is called point-to-point. To reject ambient light and achieve high data rates and low path loss, all the optical power is confined in a narrow beam oriented to the receiver. Therefore these links require pointing. Moreover, they are sensitive to blocking and shadowing.

LOS links are suited for fixed positions of the emitter and the receiver. The optical path is a straight line, so there is no possibility for multi-path dispersion effects due to multiple reflections. This method lacks mobility and, depending on the distance, power budget and data rate, may require an accurate orientation of the optical heads.

LOS links can have a very long range and achieve very high data rates, but their use will be limited within the confines of a typical spacecraft where clear paths are likely to be short. Also, it is not easy to monitor the data traffic on LOS optical links, especially during or after integration of the spacecraft, and this make testing more difficult.

Diffuse links present a communication with no need of pointing between emitter and receiver. They rely on multiple reflections on walls and obstacles to diffuse the emitted optical beam. This scheme offers freedom for placing and orienting emitters and receivers and also allows mobility. The traffic can be monitored very effectively. The main disadvantages of these links are that they suffer optoelectronics bandwidth limitations, inefficient power budget, and low-pass multi-path distortion. This causes the widening of the emitted pulses in reception, thereby resulting in Inter-Symbol Interference (ISI) at high data rates. However, diffuse channels do not exhibit fading.

Quasi-diffuse communications generally consist of transmission between two terminals without LOS through a passive reflector, so these are a compromise solution between the above-mentioned methods. Such a configuration forces the receivers to face the illuminated area and consequently collect the scattered light. The Field Of View (FOV) of the receivers must be large enough to permit relaxation of the pointing requirements. The power budget and channel capacity is intermediate between LOS and Diffuse configurations.

In both diffuse and quasi-diffuse links, reflectors and repeaters may be used to distribute signals over longer distances that do not have an unobstructed path. This kind of interfacing technology is fundamentally point-to-multipoint and can be used to implement point-to-point, multicast, or broadcast type of communications. In particular, it can replace command/response type buses in spacecraft, and network type services could be implemented over it, just as they are envisioned to be provided over ESA OBDH, Mil. Std 1553B, or CAN Bus. Optical wireless interfaces, both LOS and diffuse, are relatively immune to electromagnetic interference and are unlikely to interfere with other onboard equipment.

4.3.7.3 Eyes and Skin Safety

One of the advantages of IR communications is that there is not a spectral regulation for them. However, since the energy is propagated in a free-space channel, the impact of this radiation on human safety must be considered.

There are a number of international standards bodies which provide guidelines on LED and laser emissions namely: the International Electrotechnical Commission (IEC) (IEC 60825-1), American National Standards Institute (ANSI) (ANSI Z136.1), European Committee for Electrotechnical Standardization (CENELEC) among others.

4.3.7.4 Brief Introduction to Optoelectronics

4.3.7.4.1 Basic Optical Properties of Semiconductors

As in other matter, the electrons in semiconductors can have energies only within certain bands. The energy bands correspond to a large number of discrete quantum states of the electrons, and most of the states with low energy are full, up to a particular band called the valence band. The conduction band contains more energetic electrons that are free to move throughout the material in response to applied electromagnetic energy.

Detectors and emitters are made of semiconductor materials. Their behavior is based on band-to-band photon transitions. Electron excitation from the valence to the conduction band may be induced by the absorption of a photon of appropriate energy ($E_g < h\nu$) so an electron-hole pair is created. This increases the conductivity of the material. This effect is used to detect light. Electron de-excitation from the conduction to the valence band (electron-hole recombination) may result in the emission of a photon of energy $h\nu > E_g$. Emitters use this effect.

4.3.7.4.2 Light Emitting Devices

The two most popular solid-state light emitting devices are LEDs and LDs.

Light Emitting Diodes: An LED is a light source that emits light when an electrical current is applied to it. As in other diodes, current flows easily from the p-side, or anode, to the n-side, or cathode, but not in the reverse direction. Charge-carriers—electrons and holes—flow into the junction from electrodes with different voltages. When an electron meets a hole, it falls into a lower energy level, and releases energy in the form of a photon (emission effect). The wavelength of the light emitted, and therefore its color, depends on the band gap energy of the materials forming the p-n junction.

LEDs are often used in low performance applications. Although their modulation rates are low, the fact that they emit over a larger solid angle is sometimes advantageous, particularly in cases where the link budget is solid and where beam alignment is an obstacle (for instance when the emitter and receiver are moving with respect to one another).

Laser Diodes: LEDs undergo spontaneous emission of photons when carriers traverse the band gap in a random manner. LDs exhibit a second form of photon generation process: stimulated emission. In this process, photons of energy are incident on the active region of the device. In the active region, an excess of electrons is maintained such that in this region the probability of an electron's being in the conduction band is greater than that of its being in the valence band. This state is called population inversion and is created by the confinement of carriers in the active region and the carrier pumping of the forward biased junction. The incident photon induces recombination processes to take place. The emitted photons in this process have the same energy, frequency, and phase as the incident photon. The output light from this reaction is said to be coherent. In short distance optical links, the emitters of choice are very often AlGaAs- or GaAs-based laser diodes.

4.3.7.4.3 Photodetectors

Photodetectors convert the incident radiant light into an electrical current. Since the fraction of photons producing detected photoelectrons is less than the unity (η), the electric current is $I = RP$, where P is the optical power and $R = \eta\lambda_0 (\mu\text{m})/1.24$ is the responsivity. In devices with gain, $R = G\eta\lambda_0 (\mu\text{m})/1.24$, where G is the gain.

Inexpensive photodetectors can be constructed of silicon (Si) for the 780–950 nm optical band. The photonic energy at the 880 nm emission peak of GaAs is approximately 1.43 eV. Since the band gap of silicon is approximately 1.15 eV, these photons have enough energy to promote electrons to the conduction band and hence are able to create free electron-hole pairs.

Two popular examples of photodiodes currently in use include p-i-n photodiodes (also called PIN photodiodes) and avalanche photodiodes.

PIN Photodiodes: As the name implies, PIN photodiodes are constructed by placing a relatively large region of intrinsic semiconducting material between p⁺ and n⁺ doped regions. When a photon of sufficient energy strikes the diode, it excites an electron, thereby creating a mobile electron and a positively charged electron hole. If the absorption occurs in the junction's depletion region, or one diffusion length away from it, these carriers are swept from the junction by the built-in field of the depletion region. Thus holes move toward the anode, and electrons toward the cathode, and a photocurrent is produced.

Avalanche Photodiodes: An Avalanche PhotoDiode (APD) operates by converting each detected photon into a cascade of moving carrier pairs. Weak light can then produce a current that is sufficient to be readily detected by the electronics following the ADP. The device is a strongly reverse-biased photodiode in which the junction electric field is large; the charge carriers therefore accelerate, acquiring enough energy to excite new carriers by the process of impact ionization.

4.3.8 MULTIPLE ACCESS AND MULTIPLEXING

4.3.8.1 General

Wireless communication systems are typically designed with the intention that many users will share the available bandwidth, thus requiring many separate communication links to be established. In order for a wireless system to share resources among users without interference, multiple access and multiplexing techniques are used. *Multiple access* is the ability of a wireless system to allow multiple users to share the same communication capacity with minimal interference from other users. Multiple access refers to multiple transmitters sending information to one or more receivers. *Multiplexing* refers to a single transmitter sending information to one or more receivers. Multiplexing is the process of a single user combining a number of signals into one signal, so that it can be transmitted to other users over a single radio channel. Multiplexing can be done at baseband or at radio frequency. Often multiplexing will involve combining different types of traffic, including voice, video, and data.

There are three basic multiple access techniques (see reference [14]). In Frequency Division Multiple Access (FDMA) all users share the available bandwidth at the same time, but each user transmits at a unique allocated frequency and within an allocated bandwidth. In Time Division Multiple Access (TDMA) each user is allocated a unique time slot for transmission, but all users transmit at the same frequency. In Code Division Multiple Access (CDMA) each user transmits on the same frequency and at the same time. Each user transmits pseudo-randomly coded spread spectrum signals that can be separated at the receiver by correlation with the known transmitted code. Similarly, there are three basic multiplexing techniques, including Frequency Division Multiplexing (FDM), Time Division Multiplexing (TDM) and Code Division Multiplexing (CDM). The fundamental properties of the basic multiplexing techniques are the same as the corresponding multiple access schemes.

4.3.8.2 Time Division Multiple Access (TDMA)

TDMA systems divide the entire transmission interval into time slots, and in each slot only one user is allowed to either transmit or receive a burst of data. All users transmit at the same frequency. Typically, each user is allowed to use a large part of the available bandwidth at one time, and thus TDMA systems are generally considered wideband communication systems. Guard times are provided between user bursts so that collisions are avoided. Longer guard times are beneficial to avoid collisions; however, more potential user time is wasted. Users must transmit their burst at precisely the correct time so that the burst is located in the correct position within the TDMA frame. This requires all users to have very precise timing synchronization for both entry into the TDMA network as well as maintaining correct burst timing after network entry.

4.3.8.3 Frequency Division Multiple Access (FDMA)

In FDMA systems each user is allocated a unique frequency band or channel for transmission. This allows all users to transmit at the same time. If a user is idle and has nothing to transmit, no other user can use the bandwidth and thus resources are wasted. FDMA is typically implemented in narrowband communication systems. Guard bands are provided between user channels and are essential in FDMA systems by allowing receive filters to select individual user channels without excessive interference from other users. A special case of FDMA that is highly bandwidth efficient is Orthogonal Frequency Division Multiple Access (OFDMA). In OFDMA the users are assigned orthogonal subcarriers. OFDMA is currently being used or considered for various standards including IEEE 802.16.

FDMA typically applies to radio carrier, which is more often described by frequency. However, an optical carrier is usually described by its wavelength. Therefore the term applied to an optical carrier is Wavelength Division Multiple Access (WDMA). Since wavelength and frequency are inversely proportional, the two terms are equivalent in this context.

4.3.8.4 Code Division Multiple Access (CDMA)

CDMA systems use spread spectrum techniques to allow users to occupy all of the available channel bandwidth at the same time and at the same frequency. CDMA is often referred to as spread spectrum. The most common form of CDMA is Direct Sequence CDMA (DS-CDMA). In DS-CDMA each user is allocated a unique CDMA code that is orthogonal to other user codes. The bits of a CDMA code are called chips, and the chip rate is always much greater than the data rate. The chip sequence modulates the data bits of the message to transmit and spreads the signal over a wide bandwidth. When the modulated message is received, the receiver correlates the sequence with the transmitted user CDMA code to retrieve the original data bits. The spreading and de-spreading of DS-CDMA cause transmissions to be very hard to detect as well as provides a resistance to jamming. Figure 4-8 shows an example of DS-CDMA modulation. Another form of CDMA that is commonly used is Frequency Hopping CDMA (FH-CDMA). FH-CDMA does not use a spreading code to spread the signal, but rather uses a pseudo-random pattern to hop to different frequencies at predetermined times. The frequency hopping helps to avoid narrowband interference by not spending very much time at any specific frequency. For FH-CDMA it is very important for all users to be precisely synchronized in both time and frequency. FH-CDMA is mostly used for shorter-range wireless systems and is currently used in the Bluetooth standard.

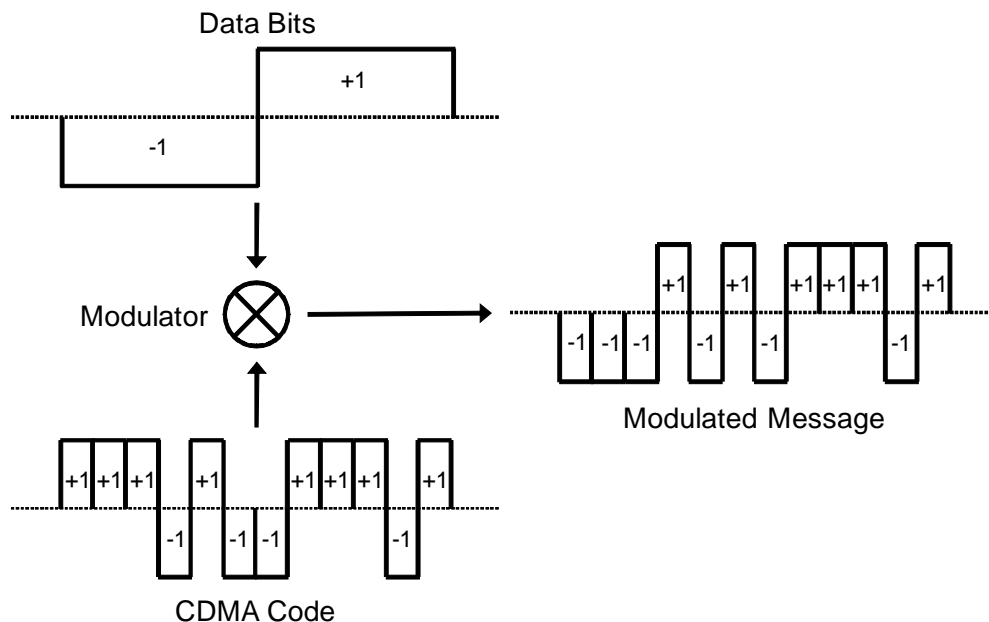


Figure 4-8: Example of DS-CDMA Modulation

4.3.8.5 Space Division Multiple Access

Space Division Multiple Access (SDMA) utilizes the spatial separation of users in order to optimize the use of the frequency spectrum. A common example of SDMA is when the same frequency is reused in different cells in a cellular wireless network. A more advanced application of SDMA uses smart antenna arrays backed by some intelligent signal processing

to steer the antenna pattern in the direction of the desired user, placing nulls in the direction of interfering signals. This enables frequency reuse within a single cell as long as the spatial separation between the users is sufficient. In typical cellular systems it is improbable to have just one user fall within the receiver beam width. Therefore it is necessary to use other multiple access techniques, such as TDMA, FDMA or CDMA, in conjunction with SDMA.

4.3.9 BAND-PASS COMMUNICATION CHANNELS

Wireless communication systems are generally restricted to operate on channels defined over a particular subinterval or *frequency band* in the frequency domain. These channels are generically referred to as *band-pass channels* and the signals transmitted over such channels are called *band-pass signals*. The treatment of band-pass channels and signals given in this subsection closely follows the development given in chapter 4 of reference [15].

It can be assumed that the signal $s(t)$ to be transmitted has a frequency content entirely contained in a narrow region in the vicinity of frequency f_c , which is called either the *center frequency* or *carrier frequency*, for reasons that will become clear. The signal $s(t)$ is assumed to be real-valued with *Fourier transform* or *frequency spectrum* $S(f)$. The frequency content of a typical band-pass spectrum is illustrated in figure 4-9.

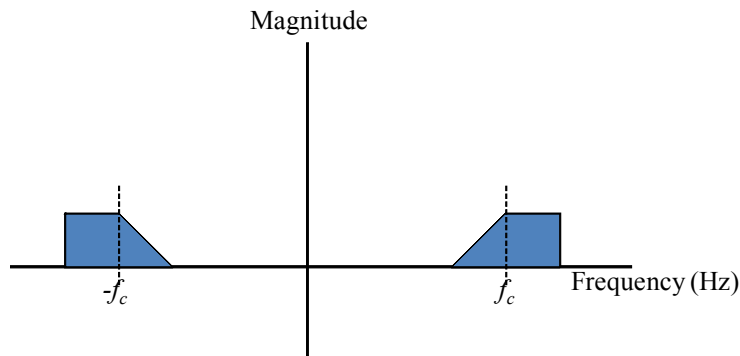


Figure 4-9: Amplitude Spectrum of a Band-Pass Signal

In particular, it is generally assumed that $S(f) = 0$ for all $|f \pm f_c| > B/2$, where B represents the *bandwidth* of the band-pass signal. The so-called *analytic signal* corresponding to $s(t)$ can be defined as the (generally complex-valued) time-domain signal $s_+(t)$ with Fourier transform given by

$$S_+(f) = \begin{cases} 2S(f), & f > 0, \\ S(f), & f = 0, \\ 0, & f < 0. \end{cases}$$

It follows that

$$\begin{aligned} s_+(t) &= \int_{f_c-B/2}^{f_c+B/2} S_+(f) e^{j2\pi ft} df \\ &= s(t) + j\hat{s}(t), \end{aligned}$$

where $\hat{s}(t)$ is the *Hilbert transform* of $s(t)$, which is given by

$$\hat{s}(t) = \frac{1}{\pi} \int_{-\infty}^{\infty} \frac{s(\tau)}{t-\tau} d\tau.$$

Finally, the *low-pass equivalent* or *baseband* signal corresponding to $s(t)$ is given by

$$s_l(t) = e^{-j2\pi f_c t} s_+(t) = x(t) + jy(t),$$

with Fourier transform $S_l(f) = S_+(f + f_c)$, where $x(t)$ and $y(t)$ are called the *in-phase* and *quadrature* components of $s(t)$, respectively.

With all of these definitions, the following can be written:

$$\begin{aligned} s(t) &= \text{Re} \left[s_l(t) e^{j2\pi f_c t} \right] \\ &= x(t) \cos 2\pi f_c t - y(t) \sin 2\pi f_c t. \end{aligned}$$

Alternatively, the following can be written:

$$\begin{aligned} s(t) &= \text{Re} \left[s_l(t) e^{j2\pi f_c t} \right] \\ &= a(t) \cos \left[2\pi f_c t + \theta(t) \right], \end{aligned}$$

where

$$\begin{aligned} a(t) &= \sqrt{x^2(t) + y^2(t)}, \\ \theta(t) &= \tan^{-1} \frac{y(t)}{x(t)}, \end{aligned}$$

are called the *envelope* (or *amplitude*) and *phase* of $s(t)$, respectively.

Hence, all band-pass signals can be viewed as either amplitude or phase modulation (or both) of an equivalent baseband signal onto a carrier. As a result, signals used in band-pass communication systems are generally defined and analyzed at baseband, and simply modulated (either in amplitude, frequency, or both) onto a carrier with the desired center frequency for transmission. At the receiver, the signals are first converted to baseband (or sometimes just to a low *intermediate frequency* or *IF*) before being sampled and demodulated in the digital domain.

4.3.10 ORTHOGONAL FREQUENCY DIVISION MULTIPLEXING (OFDM)

4.3.10.1 General

Many high data-rate wireless communication systems that operate in complex multipath environments utilize a multicarrier modulation technique called *orthogonal frequency division multiplexing* or OFDM. Since this type of modulation is referenced repeatedly throughout the remainder of this document, a brief introduction to multicarrier modulation in general and OFDM in particular is given in this subsection. The treatment of multicarrier modulation and OFDM given in this subsection closely follows the development given in chapter 12 of reference [16] and further details regarding this modulation technique can be found there.

In multicarrier modulation, the information bitstream to be transmitted over the channel is converted into several parallel streams, each of which is modulated onto a different subcarrier and transmitted over a different subchannel. The subchannel bandwidths are much less than the total channel bandwidth and under ideal conditions, the subchannels are generally mutually orthogonal. The bandwidth of each subchannel is chosen to be much smaller than the so-called *channel coherence bandwidth*, which implies that each subchannel is subjected to relatively *flat fading* that causes no substantial *multipath delay spread* on the subchannel and consequently very little *intersymbol interference* (ISI). Furthermore, multicarrier modulation can be implemented digitally very efficiently using the Fast Fourier Transform (FFT) and the Inverse Fast Fourier Transform (IFFT). This particular approach is called OFDM, and in such an implementation, ISI can be completely eliminated in a very convenient fashion using a *cyclic prefix*.

4.3.10.2 Mitigation of Subcarrier Fading

Although the ISI on each subchannel is significantly reduced when the subchannel bandwidth is less than the channel coherence bandwidth, each subchannel still experiences flat fading, and subchannels that experience deep fades will experience very high bit error rates during the fade. The common approach to combating the data loss on subchannels due to fading is to employ error-correcting coding with optional interleaving over both time and frequency. That is, the information bitstream is first encoded into codewords, the coded bits are optionally interleaved, and the resulting interleaved coded bitstream is converted into parallel streams of symbols that are transmitted on different subchannels. If this is done correctly, the bits of a particular codeword can all be transmitted over different, independently fading subchannels, which decreases the probability of a decoding error due to deep fades on a small number of subchannels.

4.3.10.3 Frequency Equalization

If the complex-valued flat-fading frequency-response coefficient corresponding to subchannel i is given by $H[i]$, then the symbols $X[i]$ transmitted on subchannel i are distorted multiplicatively in amplitude and phase to produce output symbols of the form $Y[i] = H[i]X[i]$ (ignoring additive

noise). To demodulate the distorted output symbols at the receiver, it is first necessary to equalize each subchannel by computing $\tilde{X}[i] = Y[i]/H[i]$ in order to recapture the original signal constellation geometry. Naturally, any additive noise in the received symbols is also multiplied by the same constant $1/H[i]$, so the SNR on each subchannel is unchanged by equalization. Hence, frequency equalization does nothing to improve overall detection performance on a noisy channel.

4.3.10.4 OFDM Implementation of Multicarrier Modulation

The OFDM implementation of multicarrier modulation is depicted in figure 4-10. At the transmitter, the input data stream is first converted into complex-valued Quadrature-Amplitude Modulation (QAM) symbols $X[0], X[1], \dots$. This new symbol stream is then serial-to-parallel converted to produce N parallel symbol streams to be modulated onto N different subcarriers and transmitted over N different subchannels. Each block of N symbols of the form $X[0], X[1], \dots, X[N-1]$ is then converted into a block of N time-domain samples of the form $x[0], x[1], \dots, x[N-1]$ using an IFFT. To produce the final transmitted baseband signal, a cyclic prefix is added to the output from the IFFT and then parallel-to-serial converted to produce an OFDM transmission symbol of the form $x[N-\mu], \dots, x[N-1], x[0], x[1], \dots, x[N-1]$, where μ is an integer greater than the maximum number of samples of ISI introduced by the channel. Finally, the time samples of the OFDM symbol are passed through a D/A converter to produce the desired baseband signal, which is then modulated onto in-phase and quadrature carriers for transmission as a band-pass signal.

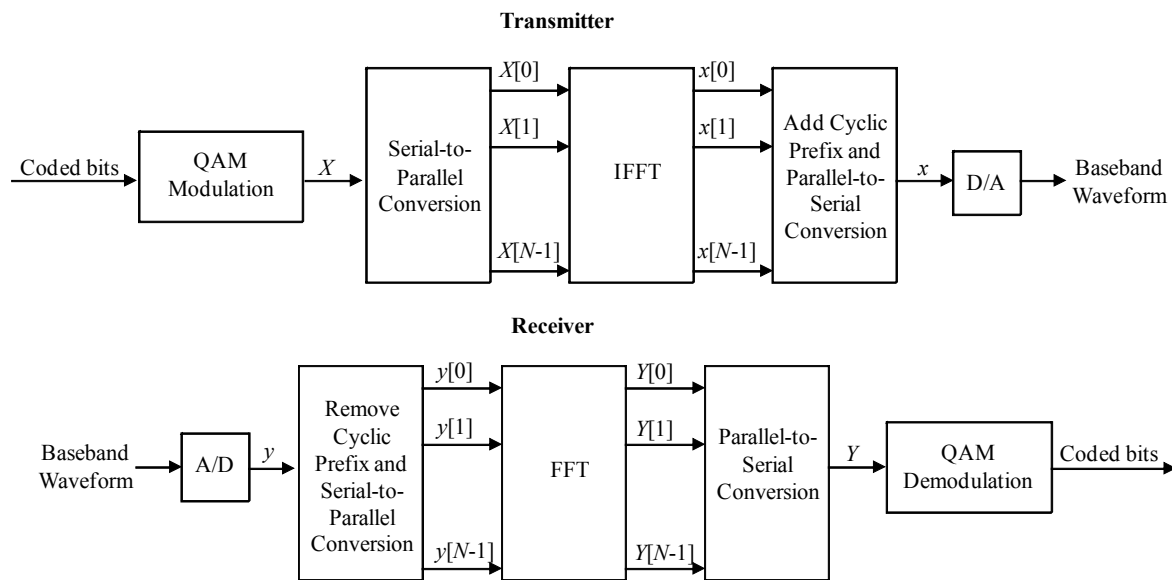


Figure 4-10: OFDM Implementation of Multicarrier Modulation

At the receiver, the received band-pass signal is first converted to baseband and then sampled to produce a received version of the OFDM symbol, from which a prefix of μ samples is discarded to eliminate ISI. The remaining received samples are then serial-to-parallel converted, transformed back into the frequency domain using an FFT, parallel-to-serial converted, frequency equalized, passed through a QAM demodulator, and output as a stream of received coded bits.

4.3.10.5 Doppler Shift and Doppler Spread

If there are no frequency and timing offsets and no relative motion between the transmitter and receiver in an OFDM system, and the channel frequency response is stable over the entire OFDM symbol interval, then the only source of error at the receiver will be additive noise. For space systems with stable oscillators and good symbol synchronization, frequency and timing offsets can generally be ignored, but Doppler shifts due to high relative velocity and Doppler spread due to time-varying multipath in the vicinity of transmitter and receiver antennas can both be a problem. Significant Doppler shifts will cause the subcarrier center frequencies and corresponding orthogonal subchannels to shift at baseband in the demodulator, effectively causing InterCarrier Interference (ICI) between the subcarriers. Similarly, significant Doppler spread will cause the bandwidth of each subchannel to increase, destroying the orthogonality of the subchannels and again causing ICI. Hence, when designing OFDM systems for space application, care must be taken to track and compensate for Doppler shifts at the receiver and to keep OFDM symbol intervals short enough that the frequency response of the channel remains nearly constant during individual symbol intervals.

5 STANDARDS BASED WIRELESS TECHNOLOGY REVIEW

5.1 WIRELESS NETWORKING STANDARDS INTRODUCTION

5.1.1 GENERAL

This section focuses on space-agency and space-exploration applicable standards for wireless networking, including emerging RFID standards (ISO 18000, EPCglobal), IEEE 802.11, IEEE 802.15, IEEE 802.16, and 3GPP LTE with the *goal of interoperable networked wireless communications*. Figure 5-1 depicts the typical maximum range or coverage area diameter of these wireless networks.

For any spacecraft or planetary wireless application there are several evaluative factors to be considered before deciding upon a specific wireless standard. The first two factors are typically the required network topology, such as an ad-hoc topology, a star topology, a point-to-point, or a point-to-multipoint topology, along with the maximum number of devices the network is expected to support at any one time. The next factors to evaluate are the required data rate and the required battery life (assuming the radio is not wall-powered). Because of the relatively small size of a spacecraft, transmit (Tx) power and transmit range typically are not design discriminating factors. Typically, for wireless spacecraft applications low power radio transmissions are desirable to reduce multipath reflections and to simply maximize battery lifetime.

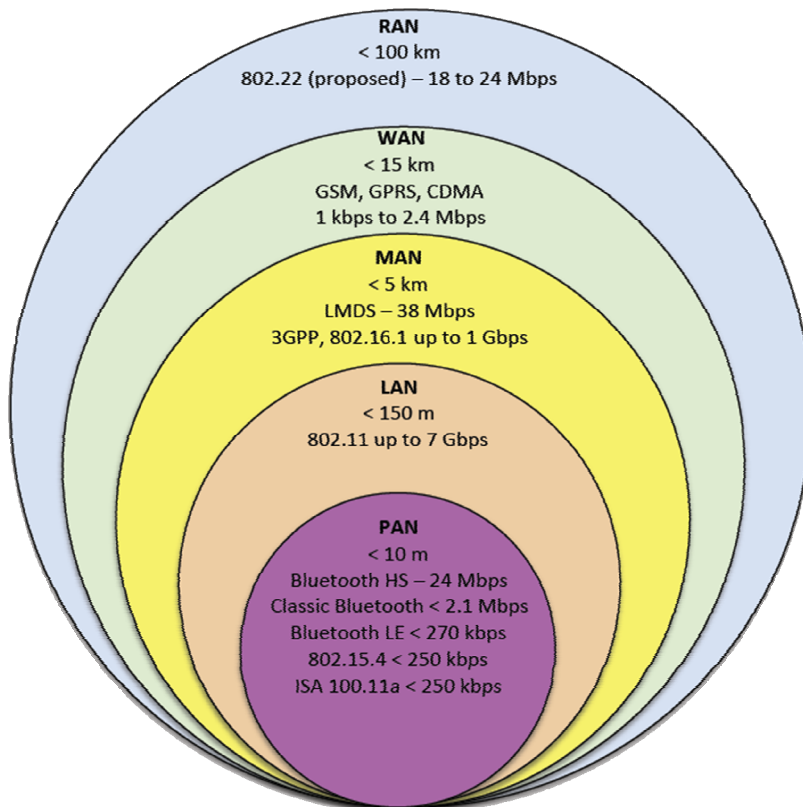


Figure 5-1: Wireless Area Network Classifications

5.1.2 RFID TECHNOLOGY OVERVIEW AND STANDARDS

5.1.2.1 Background

RFID is a method of identifying items using radio waves. The underlying concept for RFID has existed since the late 1940s when the British pioneered it to aid identification of their own aircraft (see reference [17]). However, three key hurdles were recently traversed that enabled and stimulated widespread adoption. The first of these hurdles, technological in nature, was the cost and size of the reader and tags, particularly the latter since, in an operational system, they would typically occur in much greater number and would often constitute a mobile aspect of the system. Standardization was a second significant catalyst for widespread RFID acceptance. It is important to note that standardization here pertains not just to the Physical Layer, but also to the Network and Application Layers. The third hurdle is represented by two key mandates for RFID use, one issued by the commercial sector and another by the government sector. Discussion of both the technologies involved and standardization efforts follow.

RFID technologies are used today in many applications, including security and access control, transportation and supply chain tracking, and inventory control (see reference [18]). Overall, the collective RFID technology works well for collecting multiple pieces of data on items for tracking and counting purposes in specific, cooperative environments. At the time of this publication it has not reached full potential because of technology limitations. In particular, the technology to date has been extremely effective in superseding optical barcode technology by obviating the need for LOS conditions between the reader and the tagged item. However, a number of environmental situations commonly occur that limit read success rate. For example, item-level interrogation of large groups of tagged items with metal or liquid content is often less than fully successful. Different specific RFID technologies are better suited than others in meeting particular challenges such as these. The following discussions provide some insight into these factors.

5.1.2.2 RFID Technology

Typical RFID systems are made up of two basic components: readers and tags. The reader, sometimes called the interrogator, sends and receives RF data to and from the tag via antennas. A reader may have multiple antennas that are responsible for sending and receiving the radio waves. There are many different types of tags to support a variety of applications. Tags can vary in terms of frequency at which they communicate, the protocol, how or if they are powered, and how they store data.

The tag comprises an antenna and a transponder, which can be categorized as one of three basic types: the strictly passive transponder, the transponder that scavenges power to drive an Integrated Circuit (IC) ('passive IC-based'), and the battery powered active transponder. In addition, there are hybrid versions of these three basic types. These types are discussed in more detail further below.

The power-scavenging transponder retransmits a stored ID and possibly a small amount of locally stored data. Of the three basic types addressed here, it is usually characterized by the

shortest range for specified levels of transmit power and antenna gain. The battery powered active transponder typically incorporates a battery and can transmit an ID and a fairly large amount of data. Of the three types addressed here, this type is characterized by the longest range. The strictly passive transponder re-radiates only a predetermined identification (ID) signal by reflecting energy back to the interrogator. The range of this type typically lies between the shorter range of the power-scavenged type and the longer range associated with battery-powered transponders. A hybrid semi-passive tag type contains onboard power for logic and control functions, but reflects RF energy from the interrogator in the same fashion as the first class that scavenges power; that is, this hybrid version does not use onboard resources to power RF sources. A summary of basic characteristics of the three basic tag types and additional details follow.

- a) **Strictly Passive Surface Acoustic Wave (SAW) RFID Tags** do not contain a battery and also do not contain an IC chip. Instead, the energy received from the reader is re-radiated back to the reader as a sequence of pulses using RF-acoustic conversion at the receive antenna for energy capture, acoustic propagation and attenuation/reflection along a piezoelectric substrate to create the pulses, and acoustic-RF conversion at the transmit antenna (possibly the same as the receive antenna) once again for transmission. SAW tags have no memory but have far greater read ranges than IC-based tags.
- b) **Passive IC-Based RFID Tags** do not contain a battery. Instead, they draw their power from the reader. The reader transmits a low power radio signal through its antenna to the tag, which in turn receives it through its own antenna to power the IC chip. The tag will briefly converse with the reader for verification and the exchange of data. As a result, passive tags can transmit information over shorter distances (typically 3 m or less) than active tags. They have a smaller memory capacity and are considerably lower in cost making them ideal for tracking lower cost items.
- c) **Active RFID Tags** are battery powered. They broadcast a signal to the reader and can transmit over the greatest distances (30+ meters). Shipping containers are a good example of an active RFID tag application.

In addition, both active and IC-based passive RFID tags are available in both read-only and read-write formats. Read-only tags are programmed with unique information stored on them during the chip manufacturing process. The information on read-only chips can never be changed. With read-write chips, the user can add information to the tag or write over existing information when the tag is within range of the reader. Read-write chips are more expensive than read-only chips. Another method used is called a Write-Once, Read-Many (WORM) chip. It can be written once and then becomes read-only afterwards. Chips can also vary widely in the data storage capacity of the chip. SAW tags are all read-only.

For many applications, self-powering or no-power tags are highly desirable. In the commercial sector at the time of this publication, IC-based passive RFID is far more prevalent. However, SAW-based RFID technology has some advantages that render it highly desirable for certain applications. A comparison of key attributes of IC-based and SAW-based passive RFID sensors is provided below in summary form.

Table 5-1: Summary Comparison of IC- and SAW-Based Passive RFID Technologies

Passive RFID Type	Attribute
IC-based	
General	Most common RFID form IC tags reflect or absorb incident wave to modulate the return signal
Pros	Large growth in capabilities and features anticipated Collision avoidance is easier to implement Easy to permanently disable Can assign the tag ID in the field Multiple standards exist for air interface
Cons	Tag rectifies field energy to power the IC Reduced range compared to SAW-based RFID
SAW-based	
General	Tag encoding is performed on an acoustical wave
Pros	Extremely robust Longer range than passive IC-based Typically operates with much lower transmit power Does not require <i>any</i> DC power Also has sensing capabilities (signal changes in predictable fashion in response to changes in tag temperature and/or stress) Some types of sensor telemetry are fairly mature Extremely rugged with respect to thermal and ionizing radiation environments
Cons	ID is factory programmed Collision avoidance is more difficult to implement Currently there are few providers Must account for signal distortions due to temperature/stress on tag in order to decode ID No existing standards for air interface

There are many different versions of RFID that operate at different radio frequencies. The choice of frequency is dependent on the requirements of the application. Three primary frequency bands have been allocated for RFID use:

- a) **Low Frequency (LF)** (125/134 kHz): most commonly used for access control and asset tracking;
- b) **High Frequency (HF)** (13.56 MHz): used where medium data rate and read ranges are required;
- c) **Ultra High Frequency (UHF)** (850 MHz to 950 MHz and 2.4 GHz to 2.5 GHz): offers the longest read ranges and high reading speeds.

The choice of operational frequency has important design impacts for practical RFID use. Engineering properties of higher frequency tags include:

- a) smaller tag antennas, typically the largest physical tag component;
- b) less diffraction / increased shadowing;
- c) shallower penetration of lossy and conductive media;
- d) higher implementation cost;
- e) potential for spatial diversity.

While lower frequency RFID system properties include:

- a) greater diffraction / decreased shadowing;
- b) larger antennas;
- c) lower implementation cost;
- d) broad interrogator patterns, which may limit spatial diversity.

Since Ultra High Frequency (UHF) can cover dock door portals up to three meters wide it has gained widespread industry support as the choice bandwidth for inventory tracking applications including pallets and cases. For item-level applications, the read range requirements are not as long. In addition, it becomes more difficult to place tags in positions to avoid liquids and metals for some item-level tagging applications such as pharmaceuticals.

Each RFID tag is designed to a specific protocol. The protocol defines how the tag will communicate to the outside world. If a reader is set to speak one protocol and the tag is designed to a different protocol, then the reader and the tag will not be able to communicate. Built within the protocol are features such as security (data encryption, lock abilities, etc.) and anti-collision algorithms. Technology providers are developing readers that work with multiple system protocols and frequencies so that users will be able to choose the RFID products that work best for their application area.

5.1.2.3 Surface Acoustic Wave Tags

SAW tags do not contain a battery or an IC chip. The tags are completely passive and transmit information simply by reflecting energy back to the reader. SAW tags have no memory but can be interrogated at far lower received power levels (hence far longer ranges) than IC-based tags. In addition, the tags have some inherent sensing capabilities.

The operation of one type of SAW tag, termed a reflecting tag, is illustrated in figure 5-2. As the figure indicates, a pulse transmitted by the reader is received at the tag antenna and converted into an acoustic signal by the InterDigital Transducer (IDT) connected to the antenna. The acoustic signal propagates as a compression wave along the surface of the piezoelectric tag substrate and is partially reflected back to the IDT at each of the reflectors

etched onto the substrate. When the reflected pulses reach the IDT, they are converted back into electrical signals and re-radiated from the antenna as a sequence of pulses that constitutes the impulse response of the tag. The relative amplitude, timing, and/or phase of the sequence of reflected pulses encode the ID of the tag and are determined by the position and reflection coefficient of each of the tag reflectors.

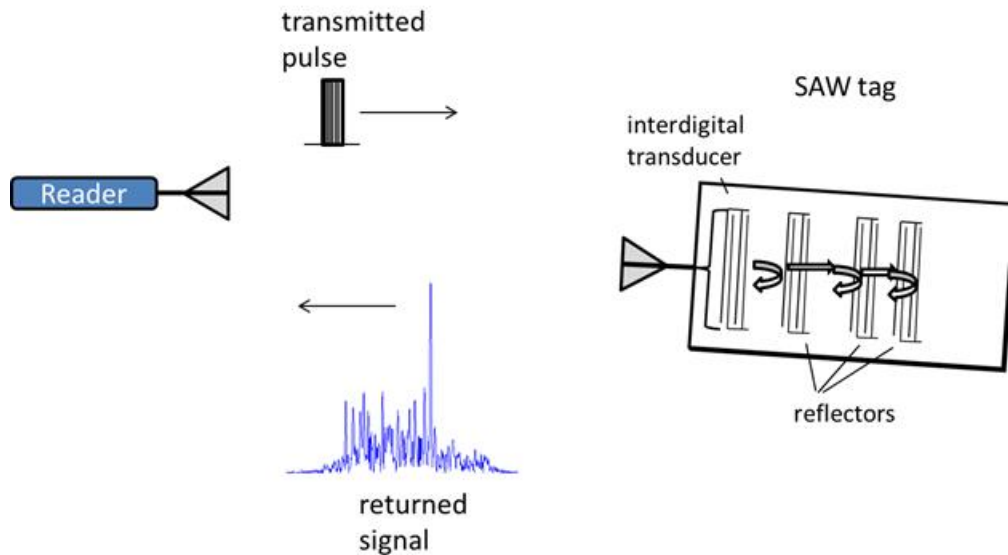


Figure 5-2: SAW-Based RFID Tag Operation

The impulse response of a SAW tag changes in response to both the temperature of the tag and the stress on the tag substrate. Hence, the tag can be used to sense both temperature and stress. The temperature sensing modality is by far the more common application and is described briefly below.

The temperature of a SAW RFID tag can be estimated by direct measurement of the time dilation (or contraction) of the tag impulse response. In particular, measurement of the time dilation of the impulse response at an arbitrary temperature relative to the response at a known reference temperature (usually 0° C) constitutes an observation of the Temperature Coefficient of Delay (TCD) for the tag at its current temperature. Here, the term TCD refers to the mathematical function of temperature that quantifies the relationship between the relative time dilation of the tag response and the temperature of the tag, with respect to a fixed reference temperature. Although the TCD can theoretically be determined from the piezoelectric properties of the crystalline material used to manufacture the tag, it is more common (and probably more accurate) to estimate it experimentally.

5.1.3 RFID STANDARDS

There are two primary competing RFID standardization efforts: ISO and EPCglobal.

The International Organization for Standardization (ISO) is the world’s largest developer and publisher of International Standards. It is a network of the national standards institutes of 157 countries, one member per country, with a Central Secretariat in Geneva, Switzerland, responsible for coordinating the system of standards development and related activities. ISO is a non-governmental organization that forms a bridge between the public and private sectors. The CCSDS is directly affiliated with ISO, and, similar to the CCSDS, ISO enables a consensus to be reached on solutions that meet both the requirements of business and the broader needs of society.

EPCglobal was formed in October 2003 as the successor organization to the MIT Auto-ID Center, the original creator of the EPC technology. EPCglobal manages the EPC network and standards, while its sister organization, Auto-ID Labs, manages and funds research on the EPC technology. EPCglobal has a very specific focus of developing standards for a system that would ultimately allow unique identification of manufactured goods along with an information system that could retrieve a lifetime history for such goods. Such historical information may include, for example, date and place of manufacture, lot number, and transportation history from the moment of manufacture.

From a pragmatic perspective both ISO and EPCglobal strive to produce an RFID communication and data exchange standard to enable interoperability of multi-vendor systems. Historically, communication protocol standards have almost exclusively been the domain of IEEE and ISO. The CCSDS is the space-communications standards committee for ISO. The Electronic Product Code (EPC) is not an international standard approved by ISO. However, EPC has significant traction because of the familiar UPC bar codes and member clout of the EPCglobal consortium. Most importantly, EPC deals with more than just how tags and readers communicate: EPCglobal has established and maintains *network standards* to govern *how EPC data is shared* among companies and other organizations.

Table 5-2: Summary of RFID Standards and Frequency Bands

Frequency Band	LF 125/134.2 kHz	HF 13.36 MHz	HF 433 MHz	UHF 860–960 MHz	UHF 2.45 GHz
ISO	ISO 11784 ISO 18000-2A ISO 18000-2B	ISO 14443 ISO 15963 ISO 18000-3	ISO 18000-7	ISO 18000-6A ISO 18000-6B ISO 18000-6C	ISO 18000-4
EPCglobal				Class 0 Class 1 Class 1 Gen 2	

The EPCglobal Class 1 Gen 2 is one of the most rapidly growing standards (see reference [19]). Interrogators operate somewhere within the 860–960 MHz band, whereas tags are required to operate over that full range. European readers typically operate in the lower part of that band, and U.S. readers operate in the upper part. EPC Class 1 Gen 2 utilizes passive, IC-based RFID tags. Range has been reported historically as less than three meters, although at the time of this publication, ranges in the vicinity of seven meters are not uncommon with moderate gain (e.g., 8 dBi) interrogator antennas and approximately 1 W transmit power. The EPC Class 1 Gen 2 specification forecasts future classes with advanced features such as sensor capabilities, tag-tag communications, and ad hoc networking. It is important to note that, in 2006, ISO approved the EPC Class 1 Gen 2 standard as an amendment to its 18000-6 standard (reference [20]).

For space-centric operations the following practical observations are identified: (1) CCSDS agency members are considered to be ‘high-end’ RFID users who will share some technical hurdles in common with terrestrial industrial users, e.g., the problem of tags obscured by metal or liquid; and (2) tag and portal costs can be appreciably higher than for terrestrial industrial users without impacting the return on investment for the use of the technology. RFID technologies are applicable to the application areas of:

- a) inventory management;
- b) localization;
- c) portal-based readers and longer-range tag interrogation;
- d) assurance of ready access to spares;
- e) enhanced situational awareness.

Best practices and recommendations regarding RFID considerations for space agency utilization are contained in *Spacecraft Onboard Interface Services—RFID-Based Inventory Management Systems* (reference [5]).

5.1.4 WPAN TECHNOLOGY OVERVIEW AND STANDARDS

5.1.4.1 General

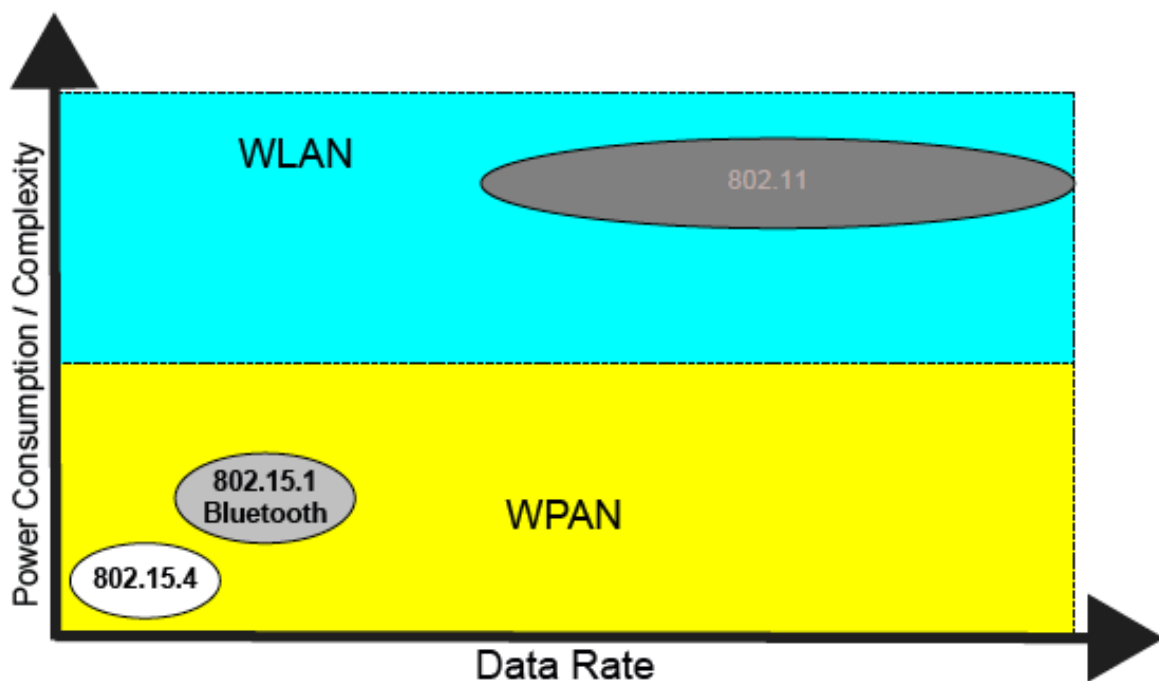


Figure 5-3: Operational Characteristics of Various WLAN and WPAN Standards

WPANs are used to convey information over relatively short distances among the participant receivers. Unlike WLANs, connections effected via WPANs involve little or no infrastructure. This allows small, power efficient, inexpensive solutions to be implemented for a wide range of devices.

The IEEE 802.15 Working Group has defined three classes of WPANs that are differentiated by data rate, battery drain, and QoS. The high-data rate WPAN (802.15.3) is suitable for multimedia applications that require very high QoS. Medium-rate WPANs (802.15.1/Bluetooth) are designed as cable replacements for consumer electronic devices centered on mobile phones and PDAs with a QoS suitable for voice (9.6–64 kb/s) applications. The last class of WPAN, LR-WPAN (802.15.4) is intended to serve applications enabled only by low power and cost requirements not targeted in the 15.1 or 15.3 WPANs. LR-WPAN applications have a relaxed need for data rate and QoS. Figure 5-3 (shown above) illustrates the operating space of the 802 WLAN and the WPAN standards. The IEEE 802.15.4 standard is not designed to overlap with higher end wireless networking standards. LR-WPAN technology is designed for applications where WLAN solutions are too expensive or extremely low-power operation is needed, and/or the performance of a technology such as Bluetooth is not required.

Annex C identifies additional specifications regarding WPAN, WLAN and WMAN wireless networks.

5.1.4.2 IEEE 802.15.1 and Bluetooth WPANs

Bluetooth is a short-range, low bandwidth WPAN technology that was originally published as IEEE 802.15.1 but has since gone on to be maintained outside the IEEE by the Bluetooth Special Interest Group (SIG). Similar to IEEE 802.11 and IEEE 802.15.4, it operates in the 2.4 GHz ISM band as designated by the FCC and similar governing bodies in Europe and Asia. Bluetooth employs Frequency Hopping Spread Spectrum (FHSS) modulation to divide this frequency range into 79 1-MHz subchannels and hops from channel to channel 1600 times a second as depicted in figure 5-4.

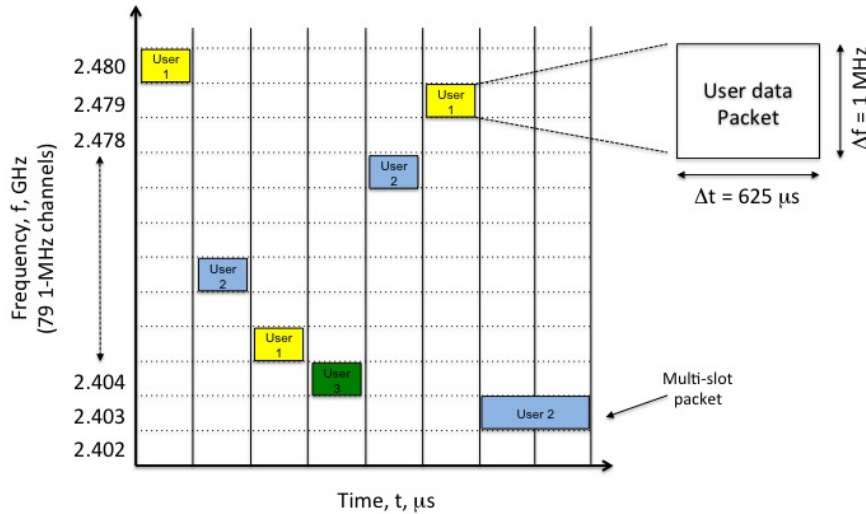


Figure 5-4: IEEE 802.15.1 Bluetooth Frequency Hopping Spread Spectrum³

Transmitting and receiving devices must synchronize on the same hop sequence to communicate. Bluetooth wireless networks and devices are designed to be relatively low-powered to maximize battery life. Most Bluetooth devices transmit at a power level of 1 mW (0 dBm). A Bluetooth network can support both data and voice links, but is limited to an eight-member piconet with one master and up to seven slaves. Several piconets can be combined to form a scatternet, which enables a hierarchical network topology (see figure 5-5). Because of the Bluetooth networking architecture, its range and data throughput are constrained; it is best suited as a cable-replacement technology, rather than as a replacement for the Wi-Fi WLAN networks.

Bluetooth v. 1.1 was originally released as IEEE 802.15.1-2002, but Bluetooth v. 1.2 and subsequent iterations have been released under the Bluetooth SIG. The current Bluetooth v. 4.0 release supports three protocols: Classic Bluetooth, Bluetooth High Speed, and Bluetooth Low Energy. Classic Bluetooth covers the legacy Bluetooth protocols and provides data rates up to 2.1 Mb/s. Bluetooth Low Energy introduces a new protocol stack aimed at very low-power applications running on small capacity batteries and provides up to 260 kb/s achievable data rate. Bluetooth High Speed uses a hybrid Bluetooth/802.11 connection to achieve data rates up to 24 Mb/s (reference [21]).

³ Source: reference [58].

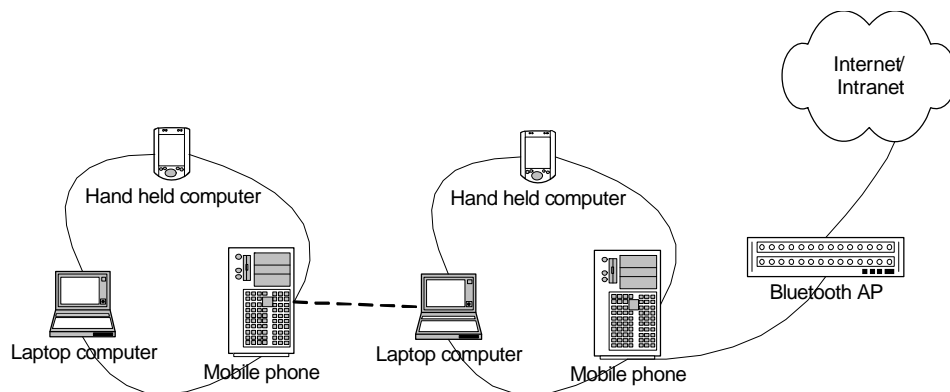


Figure 5-5: Two Bluetooth Piconets Combine to Form a Simple Scatternet⁴

5.1.4.3 IEEE 802.15.4 WPAN

IEEE 802.15.4 devices have ultra low power and low bandwidth requirements, and the standard is primarily aimed at the expected proliferation of wireless sensor networks for monitoring and control applications (see reference [22]). Questions have been raised as to whether 802.15.4 and Bluetooth are aimed at the same market. Although certainly several areas of the market overlap, the two systems have several important differences. Bluetooth is more suited for ad hoc networks, where users come and go at will, whereas 802.15.4 operates better with nodes that are reasonably static. A Bluetooth piconet (figure 5-5) is usually somewhat short-lived, is limited to only eight active devices, and is able to transfer different types of data (asynchronous, isochronous, and synchronous) with reasonable efficiency. A standard 802.15.4 network will contain a PAN coordinator and up to 65535 nodes (when using 2B short addressing), but the network itself is most efficient when network size and transmit duty cycles are low and data frames are small. Thus an 802.15.4 network does not support isochronous or synchronous data link types. A final important operational difference is that battery-powered Bluetooth devices are expected to be periodically recharged whenever necessary, whereas 802.15.4-equipped devices are expected to run for months or years on a primary battery.

The IEEE 802.15.4 standard does not define a complete protocol stack in the Open Systems Interconnection (OSI) model. Instead, it provides the Physical (PHY) Layer and Medium Access Control (MAC) sublayer of the Data Link Layer in an OSI-type stack. IEEE 802.15.4-2011, the most current iteration as of the publication of this Green Book, defines seven different PHY modulation schemes operating across a number of different frequency bands. Of these, the Offset Quadrature Phase-Shift Keying (O-QPSK) Direct Sequence Spread Spectrum (DSSS) PHY operating at 2.4 GHz has seen the widest deployment and forms the basis of most other protocols building on IEEE 802.15.4. It provides a Physical Layer data rate of 250 kb/s. The IEEE 802.15.4 MAC sublayer defines the mechanism for building a PAN using two classes of devices: Full-Function Devices (FFDs) and reduced-function devices (RFDs). Each PAN must have an FFD designated as the PAN coordinator, which is responsible for advertising the PAN to other devices and mediating the process of

⁴ Source: reference [55].

joining the PAN. Both RFDs and FFDs can associate with the PAN coordinator, and once joined to the PAN, other FFDs can act as secondary coordinators, advertising the PAN to candidate ‘child’ devices and associating with them as synchronization ‘parents’.

Once joined to the PAN, devices can operate in one of two modes: beaconing and non-beaconing. In beaconing mode, a coordinator divides time into a series of equal-sized intervals called frames and groups these into a repeating set called a superframe. The superframe begins with a beacon frame transmission from the coordinator. If the coordinator has pending messages for any of its child devices, their addresses will be contained in the beacon. The beacon frame also provides a basic time synchronization service for the child nodes. Once a child node has executed all its tasks in a superframe (e.g., requesting/receiving any pending message from the coordinator and sending any message to the coordinator), it can set a timer and deactivate its receiver to go into a lower-power ‘sleep’ state, to ‘awaken’ and re-activate its receiver in time for the process to repeat with the beginning of the next superframe. Following transmission of the beacon frame, the coordinator enters a phase where a number of subsequent frames are designated as Contention-Access Periods (CAPs) and Contention-Free Periods (CFPs). In the CAP frames, nodes compete for channel access, using a Carrier Sense Multiple Access with Collision Avoidance (CSMA-CA) scheme. A device that wishes to send a message to its coordinator checks its receiver, and if it detects the transmission of another device it backs off for a random amount of time before trying again. If no competing transmission is detected, then the device is free to transmit its message. In CFP frames, the device will have previously negotiated a guaranteed time slot with its coordinator, and it is free to use the medium without interference from other neighboring devices associated with that coordinator (i.e., CSMA-CA is not required prior to transmitting). In beaconing mode, an RFD can choose to either track its coordinator’s beacon, waking up at the beginning of each superframe, or it can choose to sleep its receiver for a longer period and re-acquire the beacon at a later time chosen by an upper networking stack layer. Once a coordinator has sequenced through the frames for its CAP and CFP, it then enters an inactive period where it is free to deactivate its own receiver.

The other operational mode allowed by IEEE 802.15.4, non-beaconing mode, does not provide a synchronization service for FFDs or RFDs. Under this paradigm, after waking up at a time defined by an upper networking stack layer, a sleeping RFD must poll its coordinator for outstanding messages. This mode allows for greater power savings at RFDs by allowing them to potentially sleep for longer periods, but the benefits are asymmetrical, as coordinators must effectively always remain awake to service awakening RFDs or risk missing the RFD polling messages. Moreover, timely delivery of messages from a coordinator to an RFD cannot be guaranteed, since the IEEE 802.15.4 Physical Layer and MAC sublayer do not define a wake-up schedule for RFDs under non-beaconing mode (a risk shared with beaconing mode when RFDs choose not to track coordinator beacons from superframe to superframe).

In addition to communicating with its coordinator and any child devices, an FFD can also communicate with a peer FFD, provided it knows the peer’s network addresses and is awake at the same time. An RFD, on the other hand, can only communicate with its parent FFD.

Using these associations, a number of potential network topologies can be built. On one end of the spectrum of complexity, this can be a very simple star topology where devices directly communicate only with the central PAN controller. On the other, nodes can form a mesh topology, where nodes can communicate with any peers within radio range, and data can travel across the network, hopping from node to node as necessary to traverse the distance from source to destination. The peer-to-peer mesh topology is typically more complex but provides a much more robust networking environment, where messages can follow multiple possible routes and deal with links that may fail at times. These two different topologies are illustrated graphically in figure 5-6.

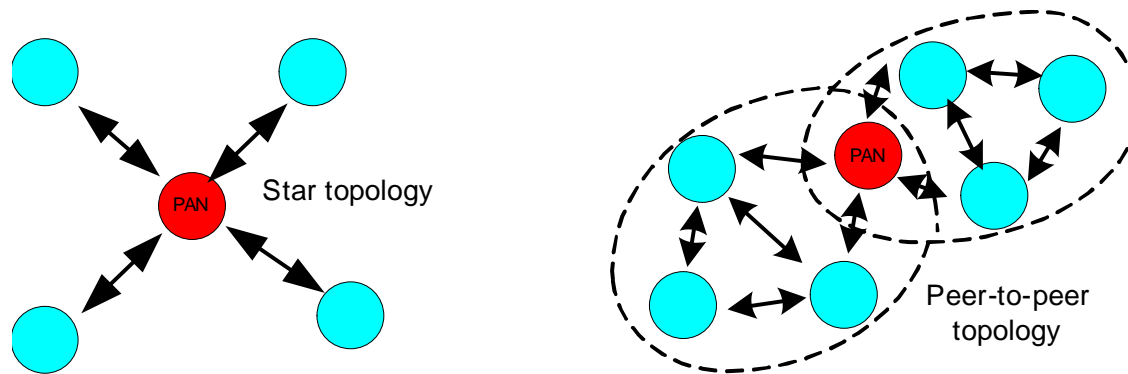


Figure 5-6: IEEE 802.15.4 Network Topologies⁵

While the Physical Layer and MAC sublayer of IEEE 802.15.4 provided the device-to-device interconnections to build these more complex topologies, they give no guidance on how to route messages through multiple intermediate devices. Nor do they give guidance on how to formulate a message to pass across the network. It is up to other protocols to provide these additional layers, for example, the Network (NWK) and Application (APP) Layers, which are respectively responsible for message routing and construction.

Standards supplying these higher protocol layers have made great progress in the decade since IEEE 802.15.4's introduction. The first, and perhaps most well-known, is ZigBee, which was introduced in 2004 but is most currently found as ZigBee-2007 (reference [23]). ZigBee defines ZigBee Coordinator (ZC) and ZigBee Router (ZR) devices that correspond to the IEEE 802.15.4 PAN coordinator and coordinator FFDs, respectively, and a ZigBee End Device (ZED) that corresponds to the IEEE 802.15.4 RFD. ZigBee-2007 makes use of the earlier IEEE 802.15.4-2003 standard, and adopts both the O-QPSK PHY in the 2.4 GHz band as discussed above and a Binary Phase-Shift Keying (BPSK) PHY operating in the 800 MHz and 900 MHz bands. The latter PHY option provides substantially lower data rates of 20 kb/s at 800 MHz and 40 kb/s at 900 MHz. ZigBee-2007 can support star, tree, and mesh topologies. The ZigBee Alliance defines two configurations of ZigBee-2007 for certification, called 'ZigBee' and 'ZigBee-PRO' (reference [24]). Both support mesh routing through Ad-Hoc On-Demand Distance Vector (AODV) routing, where a source device wishing to transmit discovers a route to a destination device via a broadcast request. Next-hop addressing is stored at each intermediate device along the route, enabling subsequent use

⁵ Source: reference [55].

of that path. Moreover, the ‘ZigBee’ profile supports tree-based routing, and ‘ZigBee-PRO’ supports source routing, where the route is specified directly in the packet header.

While the ZigBee-2007 standard discusses use of both the beacon-enabled and non-beaconing versions of the IEEE 802.15.4-2003 MAC, both the ‘ZigBee’ and ‘ZigBee-PRO’ stack profiles explicitly disallow use of any of the beacon-enabled features. Therefore ZigBee Alliance certified Zigbee-2007 implementations are beaconless. The beaconless version of the 802.15.4-2003 MAC is far easier to implement and requires less code space, but its use does imply certain restrictions on the network. For example, while non-routing ZEDs can (and should) sleep as much as possible, ZCs and ZRs must remain awake at all times to enable timely servicing of ZEDs, whose wakeup schedules are defined at their Application Layers and are unknown to their parent devices. Thus ZCs and ZRs are, for practical purposes, restricted to operate on mains rather than battery power in typical deployments. As another consequence of this lack of synchronization, timely delivery of messages to a ZED through its coordinator is not provided by the ZigBee stack profile and is instead dependent on the Application-Layer wakeup schedule at that particular ZED.

This dependence on ZED wakeup schedules is not the only feature that imposes potential latency penalties on message delivery in ZigBee-2007 networks. The decentralized nature of the protocol, while allowing for quick network formation with no global oversight, builds a framework that is reactionary in the face of message delivery failures. If a message fails to traverse the next hop in its routing path after several retries, a failure message is routed back to the sender, which must pause to broadcast another route discovery message, wait for a reply, and then re-attempt the original transmission along the replacement route. Latency issues may also arise when a ZigBee-2007 network operates in the presence of significant radio frequency (RF) interference. The CSMA-CA scheme used by the non-beaconing 802.15.4-2003 MAC backs off whenever a sending device detects energy in its channel. This energy can be due to transmissions from other ZigBee devices, but it can also come from non-802.15.4 wireless networking devices (e.g., Bluetooth, Wi-Fi), cordless telephones, EM noise from machinery, etc. Thus, sometime after its deployment in a clear PHY channel, a ZigBee network may find itself stuck in an environment subject to a high degree of interference and be unable to reliably transmit on its designated channel. ZigBee-2007 suggests that the ZC can collect transmission failure statistics from the ZRs in its network and, after a sufficient time and failure count, can perform a channel scan and decide to switch the operational frequency of the network to a clearer channel. Notification of this channel change then propagates outward among ZRs on the original channel, after which they switch to the new channel. Since both the failure statistics prompting the channel change and the change notifications require use of the original, troublesome channel, network-wide implementation and dissemination of the channel change may take quite some time. ZRs which fail to receive the notification will, after some number of failed attempts to communicate with their parent devices, begin scanning to find the new operational channel. ZEDs will only deduce this change when they awaken and, after some time, fail to reach their parent ZRs. Thus the recovery time for a channel change may be lengthy, and the latency penalty suffered by messages to/from ZEDs and ZRs may be quite high.

For these reasons, another family of higher-layer protocols has been developed as an alternative to ZigBee-2007, targeted primarily at the industrial monitoring and control market where latency bounds in the face of unpredictable RF interference are required. To answer this need, a pair of international standards have emerged: WirelessHART (IEC 62591) (reference [25]) and ISA100.11a-2011 (reference [26]). Both are originally based on the Time Synchronized Mesh Protocol (TSMP) (see reference [27]). At its Application Layer, WirelessHART is only intended to interface with Highway Addressable Remote Transducer (HART) industrial automation devices, and as such its application will be limited in the spaceflight context and its details will not be elaborated here. Instead, this summary will focus on ISA100.11a, which supports arbitrary Application Layers with a very similar set of lower protocol layers.

ISA100.11a-2011 adopts the 2.4 GHz O-QPSK DSSS PHY as described in IEEE 802.15.4-2006, but it replaces the MAC of that standard with an alternative, frequency-hopping MAC. Similar to the beaconing approach of the IEEE 802.15.4-2006, the ISA100.11a-2011 MAC divides time into slots at each device and collects these slots into superframes. Unlike IEEE 802.15.4-2006, however, network-wide synchronization of device clocks is implemented so that superframe start times are aligned at each device in the network (rather than just pair-wise at parent/child devices). This alignment of slot times allows adjacent devices to cycle their transmitters/receivers through the (up to) 16 channels of the 2.4 GHz PHY in each superframe, affording them ample opportunity to successfully complete transmissions even when some subset of the available channels are not usable because of unpredictable RF interference.

Communicating device pairs must be set so that their transmitters/receivers operate in the same channel in a given time slot, and the process must be repeated across the network for all pairs that need to communicate in that slot, so a centralized Network Manager (NM) is given configuration control of the entire network to properly optimize channel allocations. The allocation process begins when source devices that wish to transmit to destination devices request communication resources from the NM, which then configures transmitter/receiver frequency assignments in slots across each interim device hop from source to destination. Graph routing across the mesh network is used to determine route paths, and multiple alternative next-hop neighbors may be designated at each interim device. Once a route is configured and used to send a message, a device with interim custody of the message in a given slot will first perform clear-channel assessment using the CSMA-CA mechanism at its assigned transmit frequency for that slot, and provided it finds an open channel it will attempt the transmission to a neighbor and wait to receive an acknowledgement of receipt in the same time slot. If it does not receive such an acknowledgement, it will re-attempt the transmission in a subsequent slot at a different frequency (as configured by the NM). Multiple failures may trigger the device to direct further transmission attempts to a different neighbor within its radio range (again, as configured by the NM). The overall channel access scheme thus incorporates elements of frequency, time, and spatial diversity.

The NM constantly updates slot configurations to allocate and free resources as Application Layer requirements dictate, and it also harvests network statistics to continually optimize the channel access scheme. Channels that are repeatedly problematic can be blacklisted and avoided in future iterations, and they can optionally be whitelisted for re-insertion into the channel hopping pattern at a later date. Tight time synchronization allows all devices, including routers,

to turn off their receivers and enter a low-power sleep state when they are not utilized in a given time slot as transmitters or receivers. Thus both routes and end devices can be battery powered in an ISA100.11a-2011 network.

So, whereas ZigBee-2007 is reactionary in response to repeated channel access failures, ISA100.11a is proactive. Devices are configured such that they have multiple, pre-coordinated opportunities to deliver their messages across each hop from source to destination. In the all but catastrophic interference scenarios (e.g., all 16 channels continuously occupied by interferers), messages should be able to successfully traverse the network. And since the NM has centrally computed all channel-hopping patterns, worst-case latency bounds on the delivery of messages can be calculated for each source/destination device pair. This robustness comes with a price, however: establishment of an ISA100.11a-2011 network can take quite some time, compared to an ad-hoc network like ZigBee-2007, since the work to implement fail-over routing paths and channel access schemes must be done prior to a device joining and using the network to route its traffic. In addition, the network manager must be computationally far more sophisticated than the other devices in the network, given the optimization algorithms it must run. These features increase the overall cost and complexity of an ISA100.11a-2011 network.

It is notable that a recently approved amendment to IEEE 802.15.4-2011, IEEE 802.15.4e-2012 (reference [28]) suggests incorporating a MAC enhancement, referred to as Time-Synchronized Channel Hopping (TSCH), that is very similar to the ISA100.11a-2011 MAC. The goal of this update is to address industrial and commercial applications with “critical requirements such as low latency, robustness in the harsh industrial RF environment, and determinism that are not adequately addressed by IEEE Std 802.15.4-2011” (reference [28]). IEEE 802.15.4e-2012 is relatively new, and as of the publication of this Green Book it is not yet commercially available in hardware, so it is too soon to comment on the operational similarities of TSCH and the ISA100.11a-2011 MAC. However, it is possible that TSCH may in time become the preferred reference for a frequency-hopping IEEE 802.15.4 MAC in the spaceflight context.

Finally, one other feature which as of yet has limited commercial instantiations is worth mentioning in this Green Book. IEEE 802.15.4-2011 describes an ultra-wideband PHY operating in the sub-gigahertz, 3.1-4.8 GHz, and 6.0-10.6 GHz bands. Bit rates of up to 27.24 Mb/s are supported, substantially increasing the 250 kb/s rate of the standard O-QPSK PHY at 2.4 GHz. In addition to an increased transmission rate, the UWB PHY also supports ranging between devices, with range potentially accurate to 1 m.

5.1.4.4 IEEE 802.15.3 WPAN

IEEE 802.15.3 is a consortium of device manufacturers that aims at developing and designing a standard for WPAN communications based on the IEEE 802.15.3 MAC/PHY specification. It has worked with the European Computer Manufacturers Association (ECMA) standardization body to define technical specifications of such a solution. Their investigations have led to the definition of two ECMA standards ECMA 368 (reference [29]) and ECMA 369 (reference [30]). ECMA 368 defines MAC and Physical Layer of the system. ECMA 369 gives a detailed description of APIs between the MAC and Physical Layer. More

recently, ECMA 368 and ECMA 369 have been accepted as OSI standards: ISO/IEC 26907 and ISO/IEC 26908.

The system specified is well suited for short-range communications (range < 10 m). It can support high data rate applications (such as video, streaming, file transfer) as well as low data rate applications (sensors for instance). IEEE 802.15.3 is designed to be backward compatible with other wireless short-range technologies such as Bluetooth, wireless, and USB. The reuse of legacy technologies requires the definition of a Protocol Adaptation Layer (PAL) that resides on top of a common MAC sublayer (see figure 5-7).

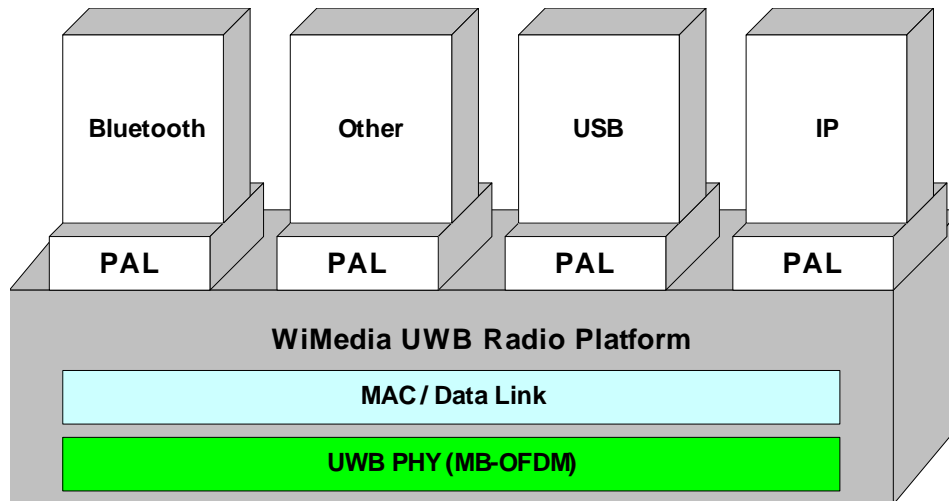


Figure 5-7: IEEE 802.15.3 Architecture with a Common MAC

IEEE 802.15.3 devices require very low power consumption, 2 mW/Mb/s according to reference [31]. Furthermore, coexistence with other wireless techniques has been also taken into account in the standards with the definition of Detect And Avoid (DAA) algorithms, typically for operation with WiMAX or WLAN (reference [31]).

IEEE 802.15.3 solutions operate in the in 3.1–10.6 GHz band. The spectrum is subdivided in 14 bands of 528 MHz each, providing a capacity of up to 480 Mb/s each (uncoded data). The transmission is based on UWB and Multi-Band Orthogonal Frequency Division Multiplexing (MB-OFDM) when building the UWB pulse. MB-OFDM uses 128 subcarriers, among which 10 are guard subcarriers, 12 are pilots, and 100 are devoted to data transmission. The modulation used is QPSK; 16-QAM can be used in some configurations. Access to medium is shared via a multi-frame handled at the MAC sublayer that allows mixing bursty traffic and isochronous traffic. The MAC sublayer works in a decentralized manner and provides two independent data transfer mechanisms adapted for bursty and periodic data:

- a) Priority Contention Access (PCA), based on Carrier Sense Multiple Access with Collision Avoidance (CSMA/CA);
- b) Distributed Reservation Protocol (DRP), reservation-based contention-free access.

MAC transmission relies on a multi-frame of 256 Medium Access Slots (figure 5-8, 256 μ s duration each). The first part of the multi-frame is used by the nodes to transmit beacons signals and reservation requests. Other slots are reserved for PCA.

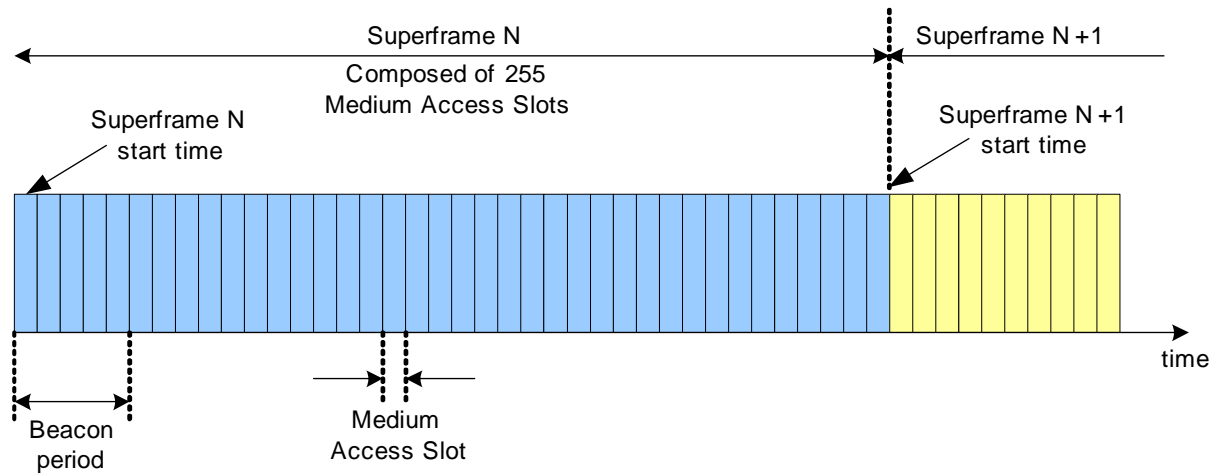


Figure 5-8: IEEE 802.15.3 MAC Superframe Structure

IEEE WiMedia is an industry alliance that is independent of standards bodies, and as such, the success and longevity of these standards may be more closely linked to the IEEE WiMedia Alliance commercial success.

NOTE – The WiMedia consortium has ceased its activities in 2009 after finalization of its UWB MAC/PHY specifications V1.5 to let the related industry groups oversee future specification developments in-house. This has coincided with the USB Implementers Forum (usb.org)’s adoption of the WiMedia UWB PHY for its WirelessUSB specification. It is worth mentioning that at the time of publishing, very few commercial computer manufacturers have integrated this technology in their platforms, generating serious doubts about the long-term survival of this high-speed technology standard.

5.1.5 WLAN TECHNOLOGY OVERVIEW AND STANDARDS

5.1.5.1 WLAN Background

WLANs were created as the wireless extension of the IEEE 802.3 LAN, which was designed for high-end data networking. Among the system requirements of a WLAN are seamless roaming, message forwarding, longest possible range, and capacity for a large population of devices distributed throughout the network. The first 802.11 WLAN standard was created in 1997; however, it only supported a maximum of two Mb/s and did not catch on. It was not until 1999 when 802.11 began to gain popularity, as the original standard was expanded creating 802.11a and 802.11b. 802.11b was the first widely accepted WLAN standard. In 2003, the 802.11g standard, which combined the best of both 802.11a and 802.11b, was

ratified and became the next widely adopted WLAN standard. In 2009, the 802.11n amendment was ratified, which again increased the maximum data rate.

Many other amendments to 802.11 have been adopted over the years that address issues beyond the Physical-Layer data rates. In 2012, the entire standard was revised and restated as IEEE 802.11-2012 (reference [32]) to include all amendments to the original standard up to that point. The summary of information regarding 802.11 presented in this document is based on reference [32] and details may be found in that document.

Two new draft standards, which will increase maximum data rates even further, have either been recently ratified or are expected to be ratified soon: 802.11ac and 802.11ad. 802.11ac operates only in the 5 GHz ISM band, and 802.11ad operates only in the unlicensed part of the 60 GHz (V-band) spectrum. An excellent discussion of the current state of the 802.11 ‘universe’ is given in reference [33]. A WLAN that uses any of the 802.11 standards is often referred to as a Wireless Fidelity (Wi-Fi) network. Finally, 802.11ad is sometimes referred to as WiGig.

5.1.5.2 WLAN Reference Model

The IEEE 802.11 standard defines only two layers of the OSI stack: the MAC sublayer of the Data Link Layer (DLL) and the PHY. The layers and sublayers described in the standard, and the manner in which they interact, are illustrated in figure 4-14 of reference [32].

To utilize an 802.11 network, the logical link layer of the network DLL, which resides outside the boundary of the 802.11 network, requests data transport services from the 802.11 MAC via the MAC subLayer Management Entity (MLME) Service Access Port (SAP). This request is actually accomplished by issuing a software function call or some similar mechanism. Conceptually, the data payload, which is called the MAC sublayer Service Data Unit (MSDU) is passed to the MAC through a separate service access port called simply the MAC SAP. In reality, this transfer is accomplished by some mechanism such as passing a pointer to the MLME SAP that identifies the location of the MSDU.

The MLME in turn requests its data transport services from the 802.11 PHY via the Physical Layer Management Entity (PLME) SAP. In this case, the MAC adds necessary header information to the MSDU to form the MAC Layer Protocol Data Unit (MPDU), which is conceptually passed through the PHY SAP to the upper sublayer of the PHY (the Physical Layer Convergence Procedure or PLCP sublayer) for transport. Within the PLCP sublayer, the MPDU becomes the Physical Layer Service Unit (PSDU), which is augmented with header information to form the Physical Layer Protocol Data Unit (PPDU) and passed to the lower layer of the PHY (the Physical Medium Dependent or PMD sublayer) for actual physical transmission.

5.1.5.3 WLAN Architecture

The basic organizational unit of any 802.11 WLAN is the *basic service set* (BSS). Although the BSS has a specific technical definition, it is probably most useful as a concept representing a set of independent wireless stations (STAs) that are logically organized to communicate with each other. There are actually three distinct BSS types: an *Independent BSS* (IBSS), an *infrastructure BSS*, and a *Mesh BSS* (MBSS). Every BSS is identified with a particular *Service Set Identifier* (SSID), but it is possible for multiple infrastructure BSSes to share the same SSID if they can communicate via a *distribution system* (DS) such as an Ethernet backbone and are organized into an *Extended Service Set* (ESS).

The simplest BSS is the IBSS, which is really just a set of STAs that have agreed to communicate with each other without recourse to either the *message forwarding* or *relaying* service provided in an MBSS or the *Access Point* (AP) available in an infrastructure BSS to provide access to a DS. IBSSes have limited utility and are excellent examples of *ad hoc networks*, which require little or no infrastructure and generally exist for only a short period of time. One of the problems with an IBSS is that even though STAs may communicate directly with each other in what is usually called *mesh mode*, they do not relay packets for each other, so that not all STAs within the IBSS may be able to communicate with each other. Forming an MBSS instead solves this problem.

Formally, an MBSS consists of autonomous stations (*mesh STAs*) that are not affiliated with any other BSS. Inside the MBSS, all STAs establish wireless links with neighbor STAs to mutually exchange messages. Further, using *multi-hop relaying* (or *forwarding*), messages can be transferred between STAs that are not in direct communication with each other. Only mesh stations participate in mesh functionalities such as formation of the MBSS, path selection, and forwarding. Accordingly, a mesh STA is not a member of an IBSS or an infrastructure BSS, and mesh STAs do not communicate directly with non-mesh STAs. However, instead of existing independently, an MBSS may also access a DS through one or more *mesh gates* and connect with other BSSes through the DS. In this way, mesh STAs can communicate with non-mesh STAs.

The infrastructure BSS is far more common than either the IBSS or the MBSS and, practically speaking, far more useful. In an infrastructure BSS, STAs communicate with each other through a unique station called an AP. That is, STAs send all of their packets to the AP and the AP relays them to other STAs within the BSS. In addition, the AP will typically serve as a *bridge* or *portal* to a (usually wired) DS. Stations within the BSS can communicate with devices outside of the BSS, or outside of the WLAN altogether, by utilizing the AP to route packets across the DS rather than to STAs within the BSS. Furthermore, STAs in one infrastructure BSS can be joined with STAs in other BSS's to form a single logical ESS, which is identified by a single SSID.

Interestingly, one use for an MBSS is to serve as the DS that connects the AP in one BSS with the AP in another or to an AP that is connected to a wired network. This is possible because one device can perform any combination of the functions of an AP, a portal, and a mesh gate. The configuration of a mesh gate that is collocated with an access point allows the utilization of the mesh BSS as a distribution system medium. In this case, two different

entities (mesh STA and access point) exist in the collocated device and the mesh BSS is hidden to STAs that associate to the access point. These ideas and architectural components are illustrated in figure 4-8 of reference [32].

The final piece of the WLAN architectural puzzle is the *Station-To-Station Link* (STSL) established using the *Direct-Link Setup* (DLS) procedure. The term STSL refers to a generic mechanism that allows direct station-to-station communication while remaining in the infrastructure mode. Establishment of this type of link includes an initialization step. The STSL is terminated by specific teardown procedures prescribed in the standard. The only example of this procedure currently specified is direct link established by the DLS.

The DLS applies only in an infrastructure BSS, and the initialization step involves a handshaking procedure between the two stations linked by the STSL. The handshake is accomplished by the two stations exchanging packets routed through the AP in the usual infrastructure manner. After the DLS has been established between the two stations, packets are transmitted directly between the two and not routed through the AP. An extension of DLS called Tunneled DLS (TDLS) can be used to establish a STSL between two stations even if the AP in the BSS is not direct-link aware. This is accomplished by encapsulating the DLS signaling (handshake) frames in standard data frames that are transmitted through the AP transparently.

5.1.5.4 WLAN Services

All delivery of data by or within an 802.11 network is accomplished by requesting *services* from the network. The delivery may or may not involve the use of a DS. A DS may be created from many different technologies including current IEEE 802 wired LANs. The 802.11 standard does not constrain the DS to be either Data Link or Network Layer based. Nor does 802.11 constrain a DS to be either centralized or distributed in nature. The standard explicitly does not specify the details of DS implementations. Instead, 802.11 specifies the services offered by the network. The services are associated with different components of the architecture. There are two categories of IEEE 802.11 service: the station service (SS), which is implemented at every 802.11 STA, and the Distribution System Service (DSS), which is implemented within the DS but accessed and supported by STAs that function as an AP, a mesh gate, or a portal. Both categories of service are used by the 802.11 MAC sublayer. The complete set of IEEE 802.11 architectural services are as follows:

- a) *Authentication*—an SS service. Used by all STAs to establish their identity to STAs with which they communicate.
- b) *Association*—a DS service. Used to support the Distribution service by associating a STA with an AP.
- c) *Deauthentication*—a SS service. Used to terminate an Authentication.
- d) *Disassociation*—a DS service. Used to terminate an Association.

- e) *Distribution*—a DS service. Used to support distribution of messages within a DS via an AP.
- f) *Integration*—a DS service. Used to support distribution of messages within a DS via a portal.
- g) *Data confidentiality*—an SS service. Used to provide data encryption and decryption to secure individual wireless links.
- h) *Reassociation*—a DS service. Used to move a current Association from one AP to another.
- i) *MAC Layer Service Data Unit (MSDU) delivery*—an SS service. Used to transfer an MSDU from the MAC on one STA to the MAC on another station over a wireless link.

NOTE – This is not a very good name for a specific service since many other services directly support MSDU delivery.

- j) *Dynamic Frequency Selection (DFS)*—an SS service. Used to satisfy regulatory requirements to implement mechanisms to avoid co-channel operation with radar systems and to provide uniform utilization of available channels.
- k) *Transmitter Power Control (TPC)*—an SS service. Used to satisfy regulatory requirements to limit transmitted power in order to mitigate interference on other systems.
- l) *Higher layer timer synchronization*—an SS service. Used to facilitate Quality of Service (QoS) provisions for applications that require time synchronization between stations.
- m) *QoS traffic scheduling*—an SS service. Used to facilitate QoS provisions that provide prioritizing access to the medium based on type of traffic.
- n) *Radio measurement*—an SS service. Used to provide the ability to request, perform, and report radio measurements in supported channels and to provide information regarding neighboring APs.
- o) *Dynamic Station Enablement (DSE)*—an SS service. Used to provide the ability to automate the channel provisioning and regulatory controls needed for unregistered IEEE 802.11 STAs to operate as dependent STAs in licensed spectrum.

5.1.5.5 WLAN Channel Plan

802.11 networks operate in several different frequency bands. As of the most recent revision of the standard (802.11-2012) there are 802.11 Physical Layer definitions for the 2.4 GHz Industrial, Scientific and Medical (ISM) band (originally defined in amendments 802.11b,g,n and the 5 GHz Unlicensed National Information Infrastructure (UNII) band (originally defined in amendments 802.11b,g,n). In addition, the 802.11ac and 802.11ad amendments define new Physical Layers in the 5 GHz UNII band and in the unlicensed 60 GHz band, respectively.

The 802.11 standard divides each of the operating frequency bands into channels in different ways. For example, the 2.4 GHz band is divided into 14 channels each 20 MHz wide using a -10 dBr bandwidth definition or 22 MHz wide using a -20 dBr bandwidth definition. The first 13 channels are spaced 5 MHz apart starting with channel 1 at 2412 MHz and channel 13 at 2472 MHz. An additional 14th channel is centered at 2484 MHz. Most of the world uses only the first 13 channels; however, North America only uses channels 1–11. Each country applies its own regulations to the allowable channels, users, and maximum power levels within each frequency band. Figure 5-9 shows the channel plan for the 14 possible 802.11 2.4 GHz channels. Up to three 802.11 networks can be concurrently deployed and co-located in space and time without interference. An example of this includes using non-overlapping channels 1, 6, and 11 for each of the networks.

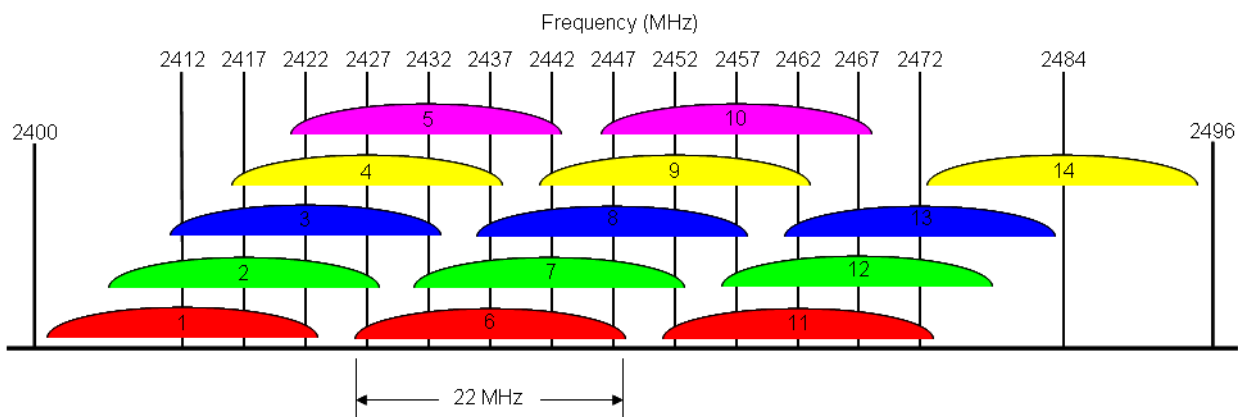


Figure 5-9: IEEE 802.11 2.4 GHz Channel Allocations

5.1.5.6 IEEE 802.11 MAC Layer

5.1.5.6.1 General

The architecture of the MAC sublayer includes the *Distributed Coordination Function* (DCF), the *Point Coordination Function* (PCF), the *Hybrid Coordination Function* (HCF), and the *Mesh Coordination Function* (MCF). A representation of the MAC architecture is shown in figure 9-1 of reference [32], in which the PCF and HCF services are provided using the services of the DCF. In a station that does not support QoS services (a *non-QoS STA*), HCF is not present. In a station that supports QoS services (a *QoS STA*), both DCF and HCF are present. PCF is optional in all STAs. Because of the distributed nature of the MBSS, only the MCF is present in a mesh STA.

5.1.5.6.2 DCF

The fundamental access method of an 802.11 MAC is a contention-based distributed algorithm implemented by the DCF. This access method is known as *carrier sense multiple access with collision avoidance* (CSMA/CA). The DCF is implemented in all 802.11 STAs.

For a STA to transmit using the DCF, it must first sense the medium to determine if another STA is transmitting. This channel sensing operation, which is called *Clear Channel Assessment* (CCA), is actually performed by the PHY as a service requested by the MAC. If CCA indicates that the medium is not busy, the transmission may proceed. The CSMA/CA distributed algorithm mandates that a gap of a minimum specified duration exists between contiguous frame sequences. A transmitting STA must verify that the medium is idle for this required duration before attempting to transmit. If the medium is determined to be busy, the STA must defer transmission until the end of the current transmission. After deferral, or prior to attempting to transmit again immediately after a successful transmission, the STA must select a random *backoff interval* and decrement the *backoff interval counter* while the medium is idle. When the backoff interval counter reaches zero, the station may transmit.

A transmission is successful either when an acknowledgement (ACK) frame is received from the MAC of the STA to which the frame (MPDU) was transmitted or when a frame addressed to multiple stations is transmitted completely. (That is, a *unicast* frame is acknowledged but a *multicast* or *broadcast* frame is not.) A refinement of this method may be used under various circumstances to further minimize collisions by having the transmitting and receiving STAs exchange short control frames prior to data transmission (called *request-to-send* or RTS and *clear-to-send* or CTS frames) after determining that the medium is idle and after any deferrals or backoffs.

5.1.5.6.3 PCF

The 802.11 MAC may also incorporate an optional access method called a PCF, which is only usable within infrastructure network configurations. This access method uses a *point coordinator* (PC) operating at the AP to determine which STA currently has the right to transmit. The operation is essentially that of polling, with the PC performing the role of the polling master. The operation of the PCF may require additional coordination, not specified in the standard, to permit efficient operation in cases where multiple point-coordinated BSSes are operating on the same channel, in overlapping physical space.

The PCF uses a virtual *carrier sense* (CS) mechanism aided by an access priority mechanism to accomplish its function. The PCF distributes information within so-called beacon management frames to gain control of the medium by setting the Network Allocation Vector (NAV) in all STAs within range of the transmission. In addition, all frame transmissions under the PCF may use an *InterFrame Space* (IFS) that is smaller than the IFS for frames transmitted via the DCF. The use of a smaller IFS implies that point-coordinated traffic has priority access to the medium over STAs in overlapping BSSes operating under the DCF access method.

The access priority provided by a PCF may be utilized to create a *contention-free* (CF) access method. The PC controls the frame transmissions of the STAs so as to eliminate contention for a limited period of time.

5.1.5.6.4 HCF

5.1.5.6.4.1 General

The QoS facility in 802.11, which is the set of enhanced functions, channel access rules, frame formats, frame exchange sequences, and managed objects used to provide parameterized and prioritized QoS, is provided by the HCF. The HCF is only usable in QoS network configurations and is implemented in all QoS STAs except mesh STAs. Instead, mesh STAs implement the MCF. The HCF combines functions from the DCF and PCF with some enhanced, QoS-specific mechanisms and frame subtypes to allow a uniform set of frame exchange sequences to be used for QoS data transfers during both the *Contention Period* (CP) and the *Contention-Free Period* (CFP). The HCF uses both a contention-based channel access method, called the *Enhanced Distributed Channel Access* (EDCA) mechanism, for contention-based transfer and a controlled channel access, referred to as the *HCF Controlled Channel Access* (HCCA) mechanism, for contention-free transfer.

STAs may obtain *transmission opportunities* (TXOPs) using both a contention-based mechanism and a controlled channel access mechanism. If a TXOP is obtained using the contention-based mechanism, it is defined as an *EDCA TXOP*. If a TXOP is obtained using the controlled channel access mechanism, it is defined as an *HCCA TXOP*.

5.1.5.6.4.2 HCF Contention-based Channel Access (EDCA)

The EDCA mechanism provides differentiated, distributed access to the channel for STAs using eight different *User Priorities* (UPs). In addition, the EDCA mechanism defines four *Access Categories* (ACs) that provide support for the delivery of traffic with UPs at the STAs. The ACs are derived from the UPs as shown in table 9-1 of reference [32]. For each AC, an enhanced variant of the DCF, called an *Enhanced Distributed Channel Access Function* (EDCAF), contends for TXOPs using a set of EDCA parameters.

5.1.5.6.4.3 HCF Controlled Channel Access (HCCA)

The HCCA mechanism uses a QoS-aware centralized coordinator, called a *hybrid coordinator* (HC), and operates under rules that are different from the PC of the PCF. The HC is collocated with the AP of the BSS and uses the HC's higher priority of access to the channel to initiate frame exchange sequences and to allocate TXOPs to itself and other STAs in order to provide limited-duration *contention access periods* (CAPs) for contention-free transfer of QoS data.

5.1.5.6.5 Mesh Coordination Function (MCF)

The mesh facility includes an additional coordination function called MCF that is usable only in an MBSS. Mesh STAs only implement the MCF. MCF has both a contention-based channel access mechanism and a contention-free channel access mechanism. The contention-based mechanism is EDCA and the contention free mechanism is called *MCF Controlled Channel Access* (MCCA).

5.1.5.6.6 Combined use of DCF, PCF, and HCF

The DCF and one of the two centralized coordination functions (either PCF or HCF) are defined so they can operate within the same BSS. When a PC is operating in a BSS, the PCF and DCF access methods alternate, with a CFP followed by a CP. When an HC is operating in a BSS, it may generate an alternation of CFP and CP in the same way as a PC, using the DCF access method only during the CP. The HCF access methods (controlled and contention-based) operate sequentially when the channel is in CP. Sequential operation allows the polled and contention-based access methods to alternate, within intervals as short as the time to transmit a frame exchange sequence.

5.1.5.7 IEEE 802.11 Physical Layers

5.1.5.7.1 General

There are actually seven different Physical Layers defined in 802.11-2012, but only two are still in common use: The Orthogonal Frequency Division Modulation (OFDM) PHY, which encompasses essentially what used to be 802.11a and 802.11g, and the High Throughput (HT) PHY, which encompasses 802.11n. The HT PHY is itself just a MIMO version of the OFDM PHY that supports up to four separate data streams using spatial multiplexing and MIMO beamforming. In addition, both the 802.11ac and 802.11ad amendments incorporate Physical Layers based on the OFDM PHY.

5.1.5.7.2 OFDM PHY

The OFDM PHY provides Physical-Layer data rates of 6, 9, 12, 18, 24, 36, 48, and 54 Mb/s on 20 MHz channels. The system uses 64 subcarriers, of which 48 are modulated with data. The remaining subcarriers are either modulated with pilot signals or set to zero to serve as guard bands. The data are modulated using binary or quadrature phase shift keying (BPSK or QPSK) or using 16- or 64-quadrature amplitude modulation (16-QAM or 64-QAM). Forward error correction convolutional coding is used with coding rates of 1/2, 2/3, or 3/4, and an 800 ns *guard interval* (GI) in the form of a cyclic prefix is appended to all OFDM symbols. The OFDM PHY also provides ‘half-clocked’ data rates of 3, 4.5, 6, 9, 12, 18, 24, and 27 Mb/s on 10 MHz channels and ‘quarter-clocked’ data rates of 1.4, 2.25, 3, 4.5, 6, 9, 12, and 13.5 Mb/s on 5 MHz channels. Half-clocked operation on 10 MHz channels doubles the OFDM symbol times and CCA times and quarter-clocked operation on 5 MHz channels quadruples the OFDM symbol times and CCA times. The OFDM PHY is defined for use in both the 2.4 GHz ISM band and the 5 GHz UNII band.

The modulation and coding schemes associated with the 8 possible (fully clocked) data rates available with the OFDM PHY are presented in table 18-4 of reference [32]. All OFDM data modulation is defined at baseband in the frequency domain as modulation of subcarriers. To transmit the data, an OFDM symbol is first transformed into the time domain using an inverse fast Fourier transform (IFFT) of length 64, padded with a cyclic prefix to create a GI, and then modulated onto the channel carrier signal for transmission in the correct frequency band. Using a cyclic prefix as a GI allows the linear convolutions performed by the physical channel to be captured and manipulated at the receiver as cyclic convolutions using the FFT and the IFFT.

5.1.5.7.3 High Throughput PHY

The High Throughput (HT) PHY is based on the OFDM PHY extended to MIMO systems with at most 4 transmit and receive antennas. Transmission on up to four *spatial streams* is defined for operation on channels with 20 MHz bandwidth. In addition, transmission on one to four spatial streams is defined for channels with 40 MHz bandwidth (sometimes called *channel bonding*). These features are capable of supporting data rates up to 600 Mb/s (four spatial streams, 40 MHz bandwidth). The HT PHY data subcarriers are modulated using BPSK, QPSK, 16-QAM, or 64-QAM. Forward Error Correction (FEC) coding is used with coding rates of 1/2, 2/3, 3/4, or 5/6 based on either Binary Convolutional Coding (BCC) or Low-Density-Parity-Check (LDPC) coding. Support for LDPC codes is an optional feature. Other optional features at both transmit and receive sides are a 400 ns short GI, transmit beamforming, HT-greenfield format (which shortens frames by eliminating backward compatibility with the basic OFDM PHY), and Space-Time-Block-Coding (STBC). The HT PHY is defined for use in both the 2.4 GHz ISM band and the 5 GHz UNII band.

For the HT PHY, there are a large number of possible data rates, corresponding to the various combinations of modulation, coding scheme, GI, and number of spatial streams (N_{ss}). Each of the possible combinations is designated by a modulation and coding scheme (MCS) index between 0 and 76. As an example, the details regarding MCS indices 0–7 are given in the tables 20–29 and 20–30 of reference [32].

5.1.5.7.4 802.11ac Very High Throughput PHY

The Very High Throughput (VHT) PHY is based on the HT PHY, which in turn is based on the OFDM PHY; however, the VHT PHY is defined for use only in the 5 GHz UNII band. The VHT PHY extends the maximum number of space-time streams supported to eight and provides support for multi-user (MU) transmissions. An MU transmission supports up to four users with up to four space-time streams per user with the total number of space-time streams not exceeding eight.

The VHT PHY provides support for 20 MHz, 40 MHz, 80 MHz and 160 MHz contiguous channel bandwidths and support for 80+80 MHz non-contiguous channel bandwidth. The VHT PHY data subcarriers are modulated using BPSK, QPSK, 16-QAM, 64-QAM and 256-QAM. FEC convolutional or LDPC coding is used with a coding rate of 1/2, 2/3, 3/4, or 5/6.

A VHT STA must support:

- 20 MHz, 40 MHz and 80 MHz channel bandwidths;
- Single spatial stream with Modulation and Coding Schemes (MCSes) 0 to 7 (transmit and receive) in all supported channel bandwidths;
- Binary convolutional coding.

A VHT STA may optionally support:

- From 2–8 spatial streams (transmit and receive);
- 400 ns short guard interval (transmit and receive);
- MIMO Beamforming (both Single-User (SU) and Downlink Multi-User (DL-MU));
- STBC (transmit and receive);
- LDPC (transmit and receive);
- MU transmit and receive;
- Support for 160 MHz channel width;
- Support for 80+80 MHz channel width;
- MCSes 8 and 9 (transmit and receive).

For a 20 MHz VHT transmission, the 20 MHz is divided into 64 subcarriers. The signal is transmitted on subcarriers -28 to -1 and 1 to 28 , with 0 being the center (DC) subcarrier. The maximum data rate on a 20 MHz channel is 693.3 Mb/s. For a 40 MHz VHT transmission, the 40 MHz is divided into 128 subcarriers. The signal is transmitted on subcarriers -58 to -2 and 2 to 58 , and the maximum data rate is 1.6 Gb/s. For an 80 MHz VHT transmission, the 80 MHz is divided into 256 subcarriers. The signal is transmitted on subcarriers -122 to -2 and 2 to 122 , and the maximum data rate is 3.4667 Gb/s. For a 160 MHz VHT transmission, the 160 MHz is divided into 512 subcarriers. The signal is transmitted on subcarriers -250 to -130 , -126 to -6 , 6 to 126 , and 130 to 250 , and the maximum data rate is 6.9333 Gb/s. For a non-contiguous 80+80 MHz VHT transmission, each 80 MHz frequency segment is divided into 256 subcarriers. In each frequency segment, the signal is transmitted on subcarriers -122 to -2 and 2 to 122 . The maximum data rate on an 80+80 MHz channel is also 6.9333 Gb/s.

5.1.5.7.5 802.11ad Directional Multi-Gigabit PHY

5.1.5.7.5.1 General

The Directional Multi-Gigabit (DMG) PHY operates in the unlicensed 60 GHz band on four 1.88 GHz channels with center frequencies of 58.32 GHz, 60.48 GHz, 62.64 GHz, and 64.80 GHz, respectively. The PHY supports three modulation methods:

- A control modulation using MCS 0;
- A single carrier (SC) modulation using MCS 1 to MCS 12 and MCS 25 to MCS 31;
- An OFDM modulation using MCS 13 to MCS 24.

The SC PHY is further sub-divided by defining a low-power option referred to separately as the low-power SC PHY. All three PHYs share a common preamble in the transmitted packet, consisting of a Short Training field (STF) and a Common Estimation field (CE). The preamble

is followed by a header, a data field, and optional training (TRN) fields to complete the packet (technically the Physical-Layer Protocol Data Unit (PPDU)). The DMG PHY may be used with no beamforming for short-range communication, but is designed to be used with optional directional beamforming (not MIMO beamforming) for longer-range communication.

The Control PHY and the SC PHY both modulate a single carrier with a data stream spread using a chip rate of 1.76 GHz. The OFDM PHY modulates data onto 336 subcarriers of bandwidth 5.15625 MHz using an IFFT of length 512 (including pilot and null subcarriers) to transform the data into the time domain for transmission. Further details for each PHY are given below.

5.1.5.7.5.2 DMG Control PHY

Support for the DMG Control PHY is mandatory. The control PHY uses MCS 0, which comprises Differential Binary Phase-Shift Keying (DBPSK) modulation, LDPC coding with rate $\frac{1}{2}$, and a data rate of 27.5 Mb/s. The data bits are scrambled, encoded, and modulated, and the DBPSK symbols are then transmitted after being spread using a complex-valued spreading sequence of length 32.

5.1.5.7.5.3 DMG SC PHY

Support for the DMG SC PHY is mandatory for some MCS indices. The various modulation and coding schemes used in the SC PPDU are described in tables 21–18 of reference [34]. MCS indices below 4 are mandatory, and the rest are optional. The data bits are scrambled, encoded, modulated, padded with a GI of 64 $\pi/2$ -BPSK modulated bits, and then transmitted in blocks of 512 complex-valued symbols. The maximum data rate for the DMG SC PHY is 4.62 Gb/s.

5.1.5.7.5.4 DMG Low-Power SC PHY

Support for the Low-Power SC PHY is optional. The modulation and coding schemes used in this PHY are listed in the tables 21–22 of reference [34]. The data bits are scrambled, encoded, modulated, padded with GI bits and transmitted in blocks of 512 complex-valued symbols. The addition of GI bits and the blocking procedure for the Low-Power SC PHY are somewhat more involved than the blocking procedure used in the SC PHY. The maximum data rate for the DMB Low-Power SC PHY is 2.503 Gb/s.

5.1.5.7.5.5 DMG OFDM PHY

Support for the DMG OFDM PHY is optional. The modulation and coding schemes used in this PHY are listed in tables 21–14 of reference [34]. The data bits are scramble, encoded, and modulated using OFDM modulation. The maximum data rate for the DMG OFDM PHY is 6.75675 Gb/s.

5.1.5.8 IEEE 802.11 Coexistence with IEEE 802.15.1 and 802.15.4

Typical RF power for 802.11 devices is between 30 mW and 100 mW. Interference between 802.15.1 and 802.11 will occur when there is an overlap of both time and frequency between transmissions associated with each technology. 802.15.1 is considered less susceptible to interference because of its frequency hopping capability. 802.11 is considered more susceptible to interference because it inhabits a fixed 22 MHz frequency band. Because of the 802.11 CSMA/CA MAC, if an 802.11 transmission is interfered with by another transmission, 802.11 will retransmit, leading to successful transmission but reduced throughput. In the case of 802.11a, which transmits in the 5 GHz UNII band, no interference potential from 802.15.1 devices exists.

Several mechanisms to reduce potential interference between the 802.15.1, 802.11 and 802.15.4 devices have been identified so that the three different wireless technologies can co-exist, including:⁶

- a) Adequate spacing between 802.11 APs and 802.15.1 APs.
- b) Strategic placement of 802.11 APs to optimize the distance between the wireless clients and the APs.
- c) Synchronization of device transmission in the time domain such that there is a low probability of more than one device transmitting at any single time. In practice, this is the more typical scenario, especially with sensors and end devices that are power-aware. These devices power up their radio transmitter only periodically and transmit their buffered information to a base station.
- d) Implementation of a collaborative mechanism, where base stations and devices exchange information between each other in an effort to intelligently optimize bandwidth between the different technologies.
- e) Engineered clear channel assignment techniques that specifically limit the hopping frequencies available to 802.15.1 devices to exist outside the 22 MHz channel band for an 802.11 implementation.

For IEEE 802.15.4 devices, where the focus is on enabling wireless sensor network communications, analyses have shown that assuming automated or manual frequency management is employed, it is reasonable to expect that the 802.15.4 network will typically have little to impact on 802.11 performance.

5.1.5.9 Additional References

Additional information regarding the 802.11 WLAN standard is provided in annex C.

⁶ Source: reference [59].

5.1.6 WMAN TECHNOLOGY OVERVIEW AND STANDARDS

5.1.6.1 General

WMANs are typically targeted for external use, as described below, extension to indoor environments is of significant interest since it could obviate the need for additional networks. This same motivation applies to spacecraft, habitats, and rovers. Even if these vehicles do support independent WLANs, an overlapping WMAN network warrants consideration for several reasons. First, interoperability between the networks must be addressed. Second, the WMAN could serve as a redundant network within the vehicle and in the vehicle proximity, providing this capability has been properly designed at the outset. Third, in certain contingency scenarios, such as that of a depressurized vehicle, the crew could be required to enter the vehicle in a pressured suit. In that case, there could be a dependency on a WMAN network established for suit communications.

5.1.6.2 WMAN Background

WMANs are intended to support Broadband Wireless Access (BWA). BWA guarantees support for user connections to core networks at data rates greater than 1.544 Mb/s, according to the ITU definition. The latest deployed forms of WMAN solutions are designed to comply with ITU International Mobile Telecommunications-2000 (IMT-2000) requirements (reference [35]), making them officially 3rd Generation (3G) RF standards. The emerging next generation of WMAN solutions are designed to meet ITU IMT-Advanced requirements (reference [36]), making them 4th Generation (4G) RF standards. IMT-Advanced requires 1 Gb/s performance for static systems, and 100 Mb/s for fast-moving mobile systems. The IEEE 802.16 and 3rd Generation Partnership Project (3GPP) Long Term Evolution (LTE) standards are comprised of 3G standards and new products to meet 4G standards. In addition, the 3GPP LTE process also defines interoperability capabilities between IEEE 802.16 and LTE from LTE Release 9 onwards (reference [37]).

5.1.6.3 WMAN and Next-Generation Networking for Disruption Tolerance

WMAN standards are generally designed for highly disruption-tolerant networking with a large number of users in highly mobile environments. This requires that the standards have a large number of defined user and management network standards that move beyond the older LAN-style IP networking seen in most IEEE standards. This Next Generation Networking (NGN) has an impact on the structure of networks that will use these standards, and represents a shift to new forms of networking designed to handle large numbers of extremely mobile nodes, for which conventional IPv4 and IPv6 concepts, and even the OSI model, start to break down. In addition, the structure of LTE, and its support for WiMAX, moves NGN to a next stage of evolution, known as a converged NGN, in which complex multimedia networking requirements are integrated into mobility. This transition is expected to be very important to the future of spaceflight, given complex mobility, multimedia and disruption-tolerant networking requirements, especially in the Human SpaceFlight (HSF) and Human Space Exploration (HSE) domains.

5.1.6.4 WiMAX Background

The central aim of the IEEE 802.16 family of standards is to address BWA, particularly for the ‘last mile’ segment. A WMAN that uses any of the 802.16 standards is often referred to as a WiMAX network. The original 802.16 standard, published in December 2001, was developed for fixed LOS deployments in the 10–66 GHz range (see reference [38]). This standard specified a single carrier modulation and offered either Time Division Duplex (TDD) or Frequency Division Duplex (FDD) variants (see reference [39]).

Soon thereafter, base station rooftop deployments were envisioned for ease of service provider and/or customer installation. The concept of rooftop deployments introduced possible Non-Line-Of-Sight (NLOS) conditions (i.e., other buildings, foliage, etc.). Therefore the 802.16a amendment was approved in January 2003. This amendment specified NLOS extensions in the 3–11 GHz range. The maximum data rate specified for this amendment was 70 Mb/s. The maximum range, however, reached out to approximately 31 miles (49.9 Km) at lesser data rates. The modulation options were extended to include single carrier, OFDM, and OFDMA (which allows users to transmit simultaneously in the uplink). Again, both TDD and FDD variants were specified. In September 2003 a revision project, called 802.16d, was initiated with the goal of aligning the 802.16 standard with the European Telecommunications Standards Institute (ETSI) HiperMAN standard as well as defining conformance and test specifications. The 802.16d project resulted in the release of 802.16-2004, which is often referred to as fixed WiMAX, and superseded all previous amendments. Mobility was not supported by 802.16-2004.

As the working group continued to address the problems associated with NLOS deployments, wireless access by smart, mobile, data hungry devices began to grab market share. The working group began to address the problem of mobility support with the development of the 802.16e-2005 amendment, which is often called mobile WiMAX, but is now absorbed into a combined fixed-mobile standard (see reference [38]). This amendment, among other things, allows for the focusing of energy by mobile units into narrower swaths of spectrum in order to combat problems associated with fading. This amendment also allows for MIMO operation with multiple antennas at both Base Station (BS) and Subscriber Station (SS). Mobile speeds of up to 120 km/h or approximately 75 mph are claimed by this amendment. The 802.16-2012 standard is presently the most-deployed version of WiMAX. To meet IMT-Advanced requirements, the IEEE 802.16m-2011 standard, now incorporated in the IEEE 802.16.1 standard (see reference [40]), has been developed, which updates the WiMAX air interface to meet the 1 Gb/s fixed and 100 Mb/s mobile requirements. IEEE 802.16.1-2012 is the core standard for the WiMAX Rel 2.0 candidate for IMT-Advanced.

5.1.6.5 WiMAX Architecture

The two main components of the WMAN architecture are BSes and SSes. The 802.16 standard was developed for Point-to-MultiPoint (PMP) networks.

The downlink is defined as the wireless link(s) that carry information from the BS to the SSes. The uplink is defined as the wireless link(s) that carry information from the SSes to

the BSes. In this architecture, shown in figure 5-10, the BS serves as the coordinator for all system resources, including timing and power. The mesh capabilities defined by the standard are also discussed in terms of this architecture.

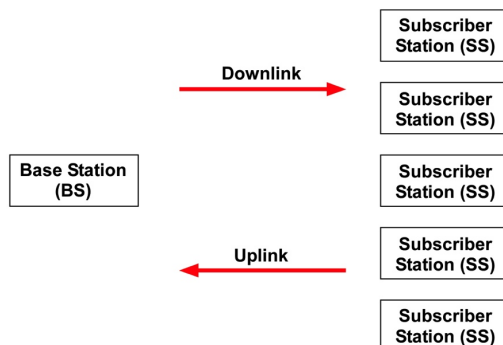


Figure 5-10: WMAN Architecture

The modern WiMAX-forum architecture (reference [41]) provides user mobility behind a series of BS/ABS units via a core network divided into two components, the Access Service Network (ASN) and the Connectivity Service Network (CSN). The ASN provides management of the radio network and provides core mobility control and a mobility anchor, allowing for the IP address of a SS/AMS node to remain fixed as the node hands-off from one BS/ABS to the next one. The CSN adds further mobility anchor functions, including IPv4/IPv6 address provision, in addition to authentication and other access control functions, while also providing connectivity to the external network. The Network Access Provider (NAP) is the business unit that provides the ASN. The Network Service Provider (NSP) is the business unit that provides the CSN. There can be more than one CSN in a network, as indicated in reference [41].

5.1.6.6 WiMAX Channel Plan

Internationally, there is not yet a uniform channel plan for WMAN systems. The 802.16 standards specify carrier frequencies up to 66 GHz and channel bandwidths up to 20 MHz; however, these have not as of yet been reflected in the available system profiles. The WiMAX forum, established to ensure the compatibility of equipment produced by various vendors, has published system profiles for 2.3 GHz, 2.5 GHz, and 3.5 GHz land mobile applications as licensed users. Additionally, a system profile is also available for unlicensed deployments in the 5.8 GHz upper UNII band. The current fixed WiMAX profiles have available channel bandwidths of 3.5 MHz, 5 MHz, 7 MHz, and 10 MHz. The mobile WiMAX profiles have available channel bandwidths of 5 MHz, 8.75 MHz and 10 MHz. Future systems are designed for use of up to a 20 MHz bandwidth, to be compliant with IMT-Advanced requirements. Much will depend on individual service providers' licensed spectrum.

Although the 802.16e-2005 amendment was intended for deployments in the 3–6 GHz range, there has been some discussion within the IEEE Working Group of deployments in the sub-1 GHz range, specifically around 700 MHz as all broadcast television moves to a digital standard. No system profiles have yet been identified for these lower frequencies.

5.1.6.7 IEEE 802.16-2012 and IEEE 802.16.1-2012 Physical Layers

The 802.16-2004 and 802.16e-2005 standards, now comprising 802.16-2012, support several different Physical Layers, including a single carrier version, OFDM, OFDMA, and what is termed as scalable OFDMA (sOFDMA). The OFDM, OFDMA and sOFDMA variants utilize different Fast Fourier Transform (FFT) sizes, equating to a varying number of subcarriers. In the TDD OFDM scheme, all subcarriers are assigned on either the uplink or downlink to an individual SS during any individual time slot. In the OFDMA and sOFDMA schemes the carrier space is broken up into a number of groups, of which there are a number of subcarriers in each group. Each subcarrier belongs to a particular subchannel, and each subchannel has one carrier in each group. The subchannels may be assigned individually to SSeS on the uplink and downlink.

802.16-2004 supports both OFDM with a FFT size of 256 and OFDMA with a FFT size of 2048. 802.16e-2005 made enhancements to the Physical Layer by employing sOFDMA, which allows for bandwidth scalability. There is a fixed relationship between the channel bandwidth and the sample rate. The sOFDMA Physical Layer in 802.16e-2005 supports FFT sizes of 128, 512, 1024 and 2048, while fixing the subcarrier frequency spacing at 10.94 kHz. This is advantageous to mobile nodes, especially when dealing with frequency shifts of the arriving signal due to Doppler effects. For instance, if constant subcarrier spacing is maintained across the entire bandwidth, Doppler shifts on the subcarriers are similar and easier to track in implementations. IEEE 802.16m-2011, now integrated into the IEEE 802.16.1-2012 standard, has improvements in the air interface, now known as WirelessMAN-Advanced, allowing for 17 b/s/Hz downlink spectral efficiency, and 9.3 b/s/Hz uplink spectral efficiency. Thus a 1 Gb/s downlink capability requires 60 MHz of spectrum, which IEEE 802.16.1-2012 would achieve by aggregation of multiple 20 MHz channels.

Even though WiMAX can provide very high spectral efficiency, OFDMA suffers from implementation issues because of the power properties of the modulation. This comes from considering the Peak to Average Power Ratio (PAPR) of the transmitted signal; OFDMA has very high PAPR, forcing the use of very linear amplifiers running at higher electrical power draw than simple single-carrier schemes. The amplifiers are expensive and, a larger issue for spaceflight, become a large power sink in mobile communication systems.

5.1.6.8 IEEE 802.16-2012 and IEEE 802.16.1-2012 MAC Layer

The 802.16-2012 standards were developed around the notions of guaranteed data flows and differentiated services. Therefore a deterministic access scheme was chosen rather than a carrier sense, contingency-based scheme as in the 802.11 WLAN standards. The MAC sublayers for 802.16-2012 are centralized and connection oriented, with each connection having a unique ID assigned by the BS. The SS only needs to compete for initial network entry, after which the SS is allocated an access slot by the BS. The access slot can expand or contract, but remains assigned to the SS. Each connection is capable of carrying various levels of data traffic. This allows the 802.16 standards to provide strong support for QoS, based on the Data Over Cable Service Interface Specifications (DOCSIS) standard (reference [39]). The MAC sublayers also utilize Automatic Repeat Request (ARQ) capabilities to perform retransmissions at the Data Link Layer if data is lost.

5.1.6.9 IEEE 802.16 Mesh Operation

The IEEE 802.16 standard describes both a PMP mode and a Mesh mode of operation. The Mesh capabilities in the standard appear to have come from some service providers' desires to have a simple path to deploy additional BSeS and repeating structures to extend their coverage or networks. Therefore the mesh capability applies most appropriately to the backhaul or BS mesh.

Although the 802.16-2012 standard makes provisions for Mesh mode, this capability is an optional portion of the standard. Current WiMAX-certified equipment is entirely provided as a cellular system replacement or overlay. The Mesh capability allows a system of BSeS to provide coverage to a service area of need. Mesh capability between SSeS is not defined in 802.16-2012. However, there is a standard amendment, IEEE 802.16j-2009 which provides for the requirements for repeaters within this architecture. This multi-hop relay capability provides extended coverage and increased throughput.

5.1.6.10 3GPP LTE Background

Unlike WiMAX standards, which are IEEE standards that grew from an extension of LAN concepts to the WMAN environment, LTE standards have grown from mobile cell-phone communications, with a growth towards IP-based networking. Release levels for LTE prior to Release 10 are designed to be a pathway to IMT Advanced implementations called LTE Advanced, and designated by Release 10 or above. LTE standards are all based around an evolution of high-mobility communications and have a sophisticated core network and protocol design to support complex services delivered with no disruption when passing from one area of LTE coverage to the next. LTE, although similar to WiMAX in modulation concepts, has some trades made in which uplink performance is reduced to decrease cost and complexity of user equipment while also reducing electrical power requirements. However, the LTE Roadmap includes downlink rates of 2998.6 Mbit/s and uplink rates of 1497.8 Mbit/s over a 100 MHz channel by Rel. 10, primarily through an ever-increasing use of MIMO.

3GPP LTE standards are many, and are referenced by their Technical Standard (TS) designation. These standards range from air interface, hand-off between wireless cells, management processes, interoperation between differing implementations and technology, user data flow and even to processes for transport of various media, including 3D stereoscopic high-rate/definition video.

LTE and the Next-Generation Mobile Network are formed via three organizations:

- The Next-Generation Mobile Network (NGMN) Alliance develops the requirements for LTE;
- 3GPP develops LTE Specifications;⁷
- LTE/SAE Trial Initiative (LTSI) performs interoperability testing.

⁷ © 2012. 3GPP™ TSs and TRs are the property of ARIB, ATIS, CCSA, ETSI, TTA and TTC who jointly own the copyright in them. They are subject to further modifications and are therefore provided 'as is' for information purposes only. Further use is strictly prohibited.

5.1.6.11 3GPP LTE and Evolved Packet System Architecture

LTE is often known as the Evolved Packet System (EPS), which refers to a division between the air interface, the Evolved Universal Terrestrial Radio Access Network (E-UTRAN), and the all-IP core network connecting to services and user network, called the Evolved Packet Core (EPC).

The EPS also has two divisions in terms of protocols and data paths: the Control Plane deals with mobility control, user authorization and QoS-level allocation, whereas the User Plane deals with actual user traffic, including providing for tunneled disruption- and mobility-tolerant transport of IPv4/IPv6 user traffic. The split of the network, from the evolved Node B (eNB) to the core network, is as shown in figure 5-11, with the most important functions being described in the following subsections. Here, the eNB is the main node found on cell towers, and provides the RF communication to and from User Equipment (UE). There can be multiple eNBs in an EPS, communicating directly with each other, or on separate subnets communicating with the EPC.

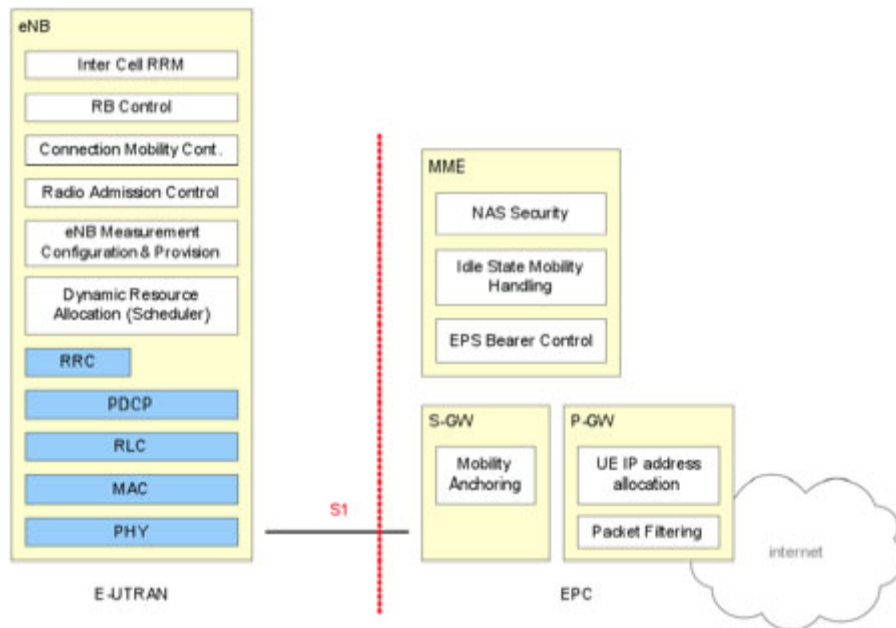


Figure 5-11: Functional Split in LTE/EPS⁸

The EPC is designed to interface to a wide set of air interface standards, ranging across other 3GPP air interfaces, but also allowing for WiMAX and Wi-Fi. It provides a wide range of services that can be considered important for mission-critical mobile networking.

⁸ Source: reference [43].

5.1.6.12 3GPP LTE Channel Plan

Unfortunately, the Channel Plan for LTE is very complex because of the requirement to ‘fit in’ the emerging LTE system in available channel blocks for mobile communication in different nations; there is no consistent international single block of LTE available. However, this has resulted in LTE hardware being available for almost any RF band from 703 MHz up to 3800 MHz. Operating bands are often fragmented and can operate in 1.25, 2.5, 5, 10, 15 and 20 MHz channels for LTE (pre-Rel. 10) with aggregation to 100 MHz total bandwidth planned for LTE Advanced (Rel. 10 and above). The presently allocated 44 bands are described in 3GPP TS 36.101, of which the present version is V11.1.0 (reference [42]).

5.1.6.13 3GPP LTE E-UTRAN Physical Layer

The E-UTRAN Physical Layer is closely related to the WiMAX Physical Layer, and is described by the 3GPP TS 36 series of standards. There are two different Physical Layer options, one based on frequency-division duplexing (FDD), and one on time-division duplexing (TDD). The version of E-UTRAN used in North America and Europe is based on FDD, and details of that Physical Layer will be described here.

In FDD E-UTRAN the downlink layer is based on fixed-spacing OFDMA with 15 kHz bandwidth subcarriers. Not all subcarriers are occupied. For 1.25 MHz channel bandwidth, a 128-point FFT is used, but only 76 carriers are occupied (for 1.14 MHz occupied bandwidth). For 20 MHz channel bandwidth, a 2048-point FFT is used, but 1201 carriers are occupied (for 18.105 MHz occupied bandwidth). The frame lasts 10 ms and is comprised of 0.5 ms slots.

The downlink layer uses many different transmission modes. A UE has 2 antennas, and an eNB may use 4 x 2 or 2 x 2 MIMO, or antenna diversity, or single antenna transmission to the UE.

The uplink in LTE is based on Single Carrier Frequency Division Multiple Access (SC-FDMA). In this process, an extra Discrete Fourier Transform (DFT) is inserted before the usual IFFT stage in OFDMA. Indeed, apart from this stage, the OFDMA parameters for LTE uplink are identical to the downlink. The uplink uses an M -point DFT that is then zero-filled before an N -point IFFT ($N < M$). The result is a signal that resembles a single-carrier spectral profile, resulting in less power consumption and low PAPR, but with a reduced data rate compared to direct OFDMA. The UE LTE uplink only uses a single transmitter and thus no SU-MIMO. However, LTE-Advanced allows for both SU-MIMO and MU-MIMO to be used by the UE for uplink, with the eNB receiving more than one data stream from different UEs in the same spectrum.

With these approaches, the LTE downlink can support 16.3 b/s/Hz for 4x4 MIMO, and the uplink can support 4.32 b/s/Hz (in 64-QAM SISO mode). The reduction in uplink spectral efficiency from WiMAX is a trade to allow for reduced amplifier cost, and increase battery lifetime, via the far lower PAPR of SC-FDMA.

5.1.6.14 3GPP LTE E-UTRAN MAC Layer

The E-UTRAN standards, just like in WiMAX standards, were developed around the notions of guaranteed data flows and differentiated services. The MAC sublayer is described in 3GPP TS 36.321, but involves complex communication with many other components of the EPS. Deterministic access is used, under the management of the eNBs, using a unique ID assigned by the services in the EPC. The UE requests network entry and proceeds through a series of stages to establish a frequency-division access slot. The MAC sublayer provides assembly and disassembly of information to and from the eNB, which is divided into multiple bearers, representing combinations of source, destination, and required QoS. The MAC sublayer implements Hybrid Automatic Repeat Request (HARQ) capabilities to perform retransmissions at the Data Link Layer if data is lost. The MAC sublayer on the eNB provides scheduling, power, modulation scheme, and channel allocation.

5.1.6.15 3GPP Evolved Packet Core

5.1.6.15.1 General

The EPC is an all-IP core network that has evolved to not only provide high-quality multimedia transport and network access, but to deal with a wide range of security and mobility issues that will be increasingly critical in a converged NGN. It is the first step to the Future Internet, in which spatial and air interface mobility is frequent and must happen with no disruption to network sockets or packet loss.

A modern full LTE deployment can use one or many mobility management protocols to avoid IP socket disruption during HO or roaming. UEs can undergo direct HO from eNB to eNB within the same network, via the X2 interface (3GPP TS 36.420), with latencies less than 5 ms. For HO when the eNBs cannot directly communicate, they can handle the transfer of custody via the S1 interface (3GPP TS 23.401 and TS 29.274), communicating into the EPC.

The IETF Proxy Mobile IPv6 (PMIP) protocol, defined by RFC 5213, can be used to provide an IP address for the UE that is fixed without the client node needing to know it is moving between physical networks (unlike Mobile IP), even as the UE undergoes HO and roaming. PMIP technology can be used by a wide range of wireless and wired communication standards (and is also standardized in WiMAX / IEEE 802.16m-2011). The older GPRS Tunneling Protocol (GTP), the basis of pre-LTE 2G and 3G technologies, can also be used, and must generally be available for legacy support. General focus is on the PMIP implementation of an EPC, given that it provides the greatest flexibility and interoperability for missions, and is an IETF standard, but the components of an EPC will also allow the GTP implementations, and will even provide interoperability between different mobility protocols, across a wide range of air interfaces and network architectures.

A basic non-roaming, PMIP-based, EPS architecture, but supporting legacy 2G/3G standards, is shown in figure 5-12. In a GTP-based EPS, the S5 reference point interface is replaced by GTP-U traffic.

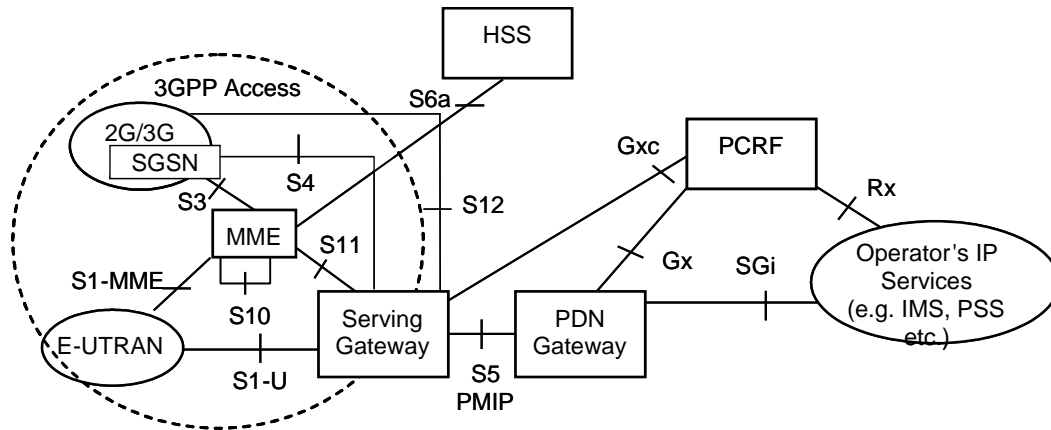


Figure 5-12: Basic Non-Roaming PMIP-Based EPS with Legacy Support⁹

Here the E-UTRAN air interface, which can include many eNB nodes, with handoff between nodes, interfaces to a 3GPP Access set of server components via the two S1 interfaces; S1-U provides interface for User Plane protocols, whereas S1-MME provides interface for Control Plane protocols. Various other servers then provide authentication, QoS allocation, and translation and interface of User Plane protocols to IP-based services, including external networks, as indicated in figure 5-12.

There are the following functions:

5.1.6.15.2 Mobility Management Entity

Mobility Management Entity (MME) functionality is defined in 3GPP TS 23.401 and TS 23.402, and provides support for UE management and access to the EPC. The MME provides for HO between eNBs that are not on the same local network (whereas eNBs on the same local network can support HO directly).

5.1.6.15.3 Home Subscriber Server

The Home Subscriber Server (HSS) provides information to the EPC on access rights and priority for UEs, and provides all authentication services. The MME queries the HSS via the Diameter (RFC 6733) protocol.

5.1.6.15.4 Serving Gateway

The Serving Gateway (SGW) provides a mobility anchor for User Plane traffic from a UE, and is described in 3GPP TS 23.401 and TS 23.402. The SGW passes traffic between the Packet Data Network (PDN) Gateway and the eNB. A UE is only represented by one SGW but it is possible to move from one SGW to another during HO.

⁹ Source: reference [44].

5.1.6.15.5 PDN Gateway

PDN Gateway (PDN GW or PDG) functionality is described in 3GPP TS 23.401 and TS 23.402. It is the path for User Plane traffic from or destined to a UE, to or from a network external to the EPS, respectively. The PDN GW processes accesses connected to the EPC via a GTP-based and/or PMIP-based S5/S8 interface. The PDN GW supports functionality specified in TS 23.401 that is common to both PMIP-based and GTP-based S5/S8 interfaces. It also provides for access to EPC via non-3GPP accesses via a range of mechanisms incorporating other servers. The EPC can support more than one external network by the use of multiple PDN GW servers. The general end-to-end protocol stack is as shown in figure 5-13. Here PDCP is the 3GPP Packet Data Convergence Protocol (3GPP TS 25.323) and RLC is the Radio Link Control protocol (3GPP TS 25.322), and provides for passage of PDCP, including automatic retransmission, from the UE to the eNB. The eNB converts PDCP User Plane data into tunneled traffic via GTP-U that is carried over UDP to the PDN GW, at which point tunnel information is stripped and the IP payload is delivered to the external network. This process is reversed for external network to UE-based Application Layer traffic.

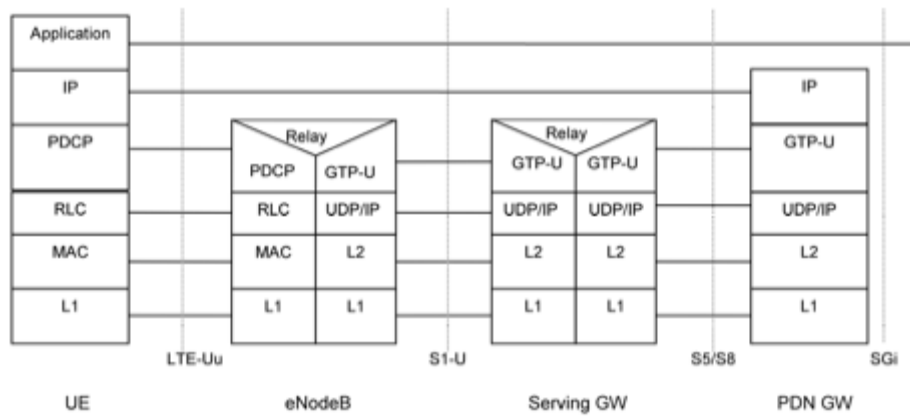


Figure 5-13: User Plane Traffic from UE to PDN GW¹⁰

Additionally, the PDN GW is the user plane anchor for mobility between 3GPP access and non-3GPP access.

5.1.6.15.6 Policy and Charging Rules Function

The functionality of Policy and Charging Rules Function (PCRF) is described in TS 23.203. The PCRF communicates with the three components of the network that handle routing of User Plane traffic, and establishes a through-network set of QoS values for each network traffic flow.

¹⁰ Source: reference [45].

5.1.6.15.7 Serving GPRS Support Node

The Serving GPRS Support Node (SGSN) is an optional node that allows older 2G/3G standards to interoperate with the EPC, providing a gateway for those technologies into the EPC, and allowing use of the MME and HSS to provide authentication and access control.

5.1.6.16 Use of other Air Interfaces in EPC

EPC, via 3GPP standards as of Release 10, provides mechanisms for non-LTE (E-UTRAN) air interface untrusted systems to connect to the network core. The key servers in this process are the enhanced Packet Data Gateway (ePDG), which provides a link from external systems into the EPC, via the PDN GW, and the Access Network Discovery and Selection Function (ANDSF), which allows a UE to discover and select a non-LTE air interface for off-loaded connection to the EPC. The ePDG is described under 3GPP TS 23.234, and the ANDSF under 3GPP TS 23.402, 24.302, and 24.312.

Trusted connections are also possible, and these take place via the PDN GW / PDG. This takes place via processes specified in the 3GPP TS 23.234 and TS 33.234 standards, using interfaces W_a , W_n , and W_g . A Wireless LAN Gateway (WAG) provides the routing and filtering functions between the alternate wireless network. The TS 33.234 specifications allow this access to be trusted, and the same methodology can be applied for multiple air interfaces. One such interface is WiMAX, and the connection from the WiMAX ASN and CSN is specified by reference [41]. 3GPP AAA services, supporting the HSS, are made available to the WiMAX system, thus providing the inter-system trust.

The 3GPP TS 23.234 and TS 33.234 specifications are core to the EPC becoming a general networking approach to hand-off and roaming disruption-tolerant communications. Presently, many wireless standards are converging on this standard approach.

This allows an EPC to be built to handle many different wireless networks inside a spacecraft, providing for failover and mobility between those networks.

5.1.6.17 Relaying in EPC

In addition to UE to eNB communications, EPS also provides for RF-relayed communication via the use of a Donor eNB (DeNB), acting as a RF gateway for the DeNB into the EPC, a Relay Node (RN) under control of the DeNB and an RN MME controlling RN mobility. The RN provides an RF-to-RF link from the UE to the EPC via the U_u and U_n interfaces. The process is specified in 3GPP TS 23.401 and is shown in figure 5-14.

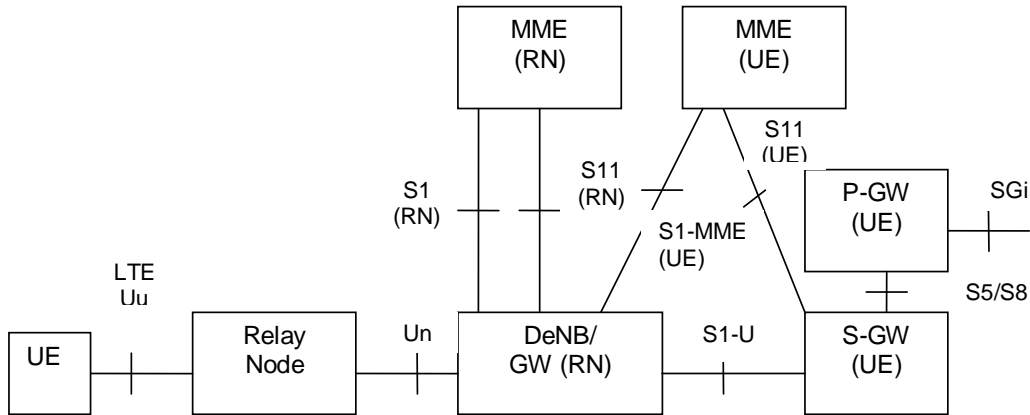


Figure 5-14: The Relay Function in LTE Advanced¹¹

5.1.7 OPTICAL COMMUNICATIONS OVERVIEW AND STANDARDS

Table 5-3: IEEE 11073 and IrDA Optical Standards

Standard	IrDA	IEEE 11073
Data rate	From 115 kb/s to 16 Mb/s	115 kb/s
Frequency band	Baseband	Baseband
Network size (# nodes)	Up to 127 (supported by high level protocols)	Up to 127 (supported by high level protocols)
Tx peak power	100 mW	100 mW
Omni range	Designed for LOS transmission	Designed for LOS transmission
Network topologies	Only master-slave configuration	Only master-slave configuration
Complexity	Low	Very low
Power requirements	Assuming a 1-percent emission time, consumption below 10 nA on standby	Assuming a 1-percent emission time, consumption below 10 nA on standby
System resources	Integrated emitter-receiver device + software controller	Integrated emitter-receiver device + software controller
Battery life (days)		
Modulation techniques	OOK, PPM	PPM
Energy / txd bit	≈0.2 nJ	≈0.2 nJ

NOTES

- For about 100 mW, IrDA is supposed to have a range about 1.5 m. This range can be increased by means of optical lenses to 3–4 meters.
- Pulse Position Modulation (PPM) is less bandwidth efficient but shows an increased robustness against multipath penalty on diffuse or quasi-diffuse channels. On the other hand, On-Off Keying (OOK) modulation is simpler to implement and easier to receive on a day-to-day basis. Also possible is a ‘direct translation’ of an OOK system on a direct-sequence spread-spectrum one.

¹¹ Source: reference [45].

5.1.7.1 The Infrared Physical Layer

Infrared and visible light are of near wavelengths and thus behave similarly. Infrared light is absorbed by dark objects, reflected by light objects, and cannot penetrate walls. Today's WLAN products that use IR transmission operate at wavelengths near 850 nm. This is because transmitter and receiver hardware implementation for these bands is cheaper and also because the air offers the least signal attenuation at that point of the IR spectrum. The IR signal is produced either by semiconductor laser diodes or LEDs, with the former being preferable because their electrical-to-optical conversion behavior is more linear. However, the LED approach is cheaper and the IEEE 802.11 IR Physical Layer specification can easily be met by using LEDs for IR transmission.

Three different techniques are commonly used to operate an IR product: diffused transmission that occurs from an omnidirectional transmitter, reflection of the transmitted signal on a ceiling, and focused transmission. In the latter, the transmission range depends on the emitted beam's power, and its degree of focusing can be several kilometers. It is obvious that such ranges are not needed for most WLAN implementations. However, focused IR transmission is often used to connect LANs located in the same or different buildings where a clear LOS exists between the wireless IR bridges or routers.

In omnidirectional transmission, the mobile node's transmitter utilizes a set of lenses that converts the narrow optical laser beam to a wider one. The optical signal produced is then radiated in all directions, thus providing coverage to other WLAN nodes. In ceiling-bounced transmission, the signal is aimed at a point on a diffusely reflective ceiling and is received in an omnidirectional way by the WLAN nodes. In cases where BSes are deployed, they are placed on the ceiling, and the transmitted signal is aimed at the BS, which acts as a repeater by radiating the received focused signal over a wider range. Ranges that rarely exceed 20 meters characterize both this and the omnidirectional technique.

IR radiation offers significant advantages over other Physical Layer implementations. The infrared spectrum offers the ability to achieve very high data rates. Basic principles of information theory have shown that nondirected optical channels have very large Shannon capacities, and thus transfer rates in the order of 1 Gb/s are theoretically achievable. The IR spectrum is not regulated in any country, a fact that helps keep costs down.

Another strength of IR is the fact that in most cases transmitted IR signals are demodulated by detecting their amplitude, not their frequency or phase. This fact reduces the receiver complexity, since it does not need to include precision frequency conversion circuits, and thus lowers overall system cost. IR radiation is immune to electromagnetic noise and cannot penetrate walls and opaque objects. The latter is of significant help in achieving WLAN security, since IR transmissions do not escape the geographical area of a building or closed office. Furthermore, co-channel interference can potentially be eliminated if IR-impenetrable objects, such as walls, separate adjacent cells.

IR transmission also exhibits drawbacks. IR systems share a part of the EM spectrum that is also used by the Sun, thus making use of IR-based WLANs practical only for indoor application. Fluorescent lights also emit radiation in the IR spectrum causing Signal-to-

Interference Ratio (SIR) degradation at the IR receivers. A solution to this problem could be the use of high-power transmitters; however, power consumption and eye safety issues limit the use of this approach. Limits in IR transmitted power levels and the presence of IR opaque objects lead to reduced transmission ranges, which means more base stations need to be installed in an infrastructure WLAN. Since the base stations are connected with wire, the amount of wiring might not be significantly less than that of a wired LAN. Another disadvantage of IR transmission, especially in the diffused approach, is the increased occurrence of multipath propagation, which leads to ISI, effectively reducing transmission rates. Another drawback of IR WLANs is the fact that producers seem to be reluctant to implement IEEE 802.11-compliant products using IR technology.

NOTE – Optical narrow-band filter can address these issues.

5.1.7.2 IrDA

The Infrared Data Association (IrDA) defines physical specifications and communications protocol standards for the short-range exchange of data over infrared light, for uses such as PANs.

The IrDA™ Standard presents different speeds:

- Standard IrDA (SIR): Up to 115 kb/s;
- Medium Speed IrDA (MIR): 1 Mb/s;
- Fast IrDA (FIR): 4 Mb/s;
- Very Fast IrDA (VFIR): 16 Mb/s.

Additionally, an Ultra-Fast IrDA (UFIR) mode that will support 100 Mb/s is under development.

The IrDA physical specifications require that a minimum irradiance be maintained so that a signal is visible up to a meter away. Similarly, the specifications require that a maximum irradiance not be exceeded so that a receiver is not overwhelmed with brightness when a device comes close. In practice, there are some devices on the market that do not reach one meter, while other devices may reach up to several meters. There are also devices that do not tolerate extreme closeness. The typical sweet spot for IrDA communications is from 5 cm to 60 cm away from a transceiver, in the center of the cone.

IrDA data communications operate in half-duplex mode because while transmitting, a device's receiver is blinded by the light of its own transmitter, and thus full-duplex communication is not feasible. The two devices that communicate simulate full duplex communication by quickly turning the link around.

5.1.7.3 IrSimple™

IrSimple™ protocol, recently proposed by the IrDA, promises a simple infrared protocol for fast wireless communication between mobile devices and digital home appliances.

IrSimple™ achieves at least 4 to 10 times faster data transmission speeds by improving the efficiency of the infrared IrDA protocol. However, the existing flow control scheme adopted by IrSimple™ protocol consumes a considerable amount of energy and resources by retransmitting large-sized information frames in case the receiving secondary station remains busy because of the handling of other tasks and therefore cannot send the acknowledgement of received frames. Some studies are being developed in order to reduce this consumption.

5.1.7.4 IEEE 11073

The IEEE 11073 standard establishes a connection-oriented transport profile and Physical Layer suitable for medical device communications that use short-range infrared wireless. It defines communications services and protocols that are consistent with specifications of the IrDA and are optimized for Point-Of-Care (POC) applications at or near the patient. This standard also supports use cases consistent with industry practice for handheld PDAs and network APs that support IrDA-infrared communication.

5.2 SUPPORTING TECHNOLOGIES

5.2.1 QUALITY OF SERVICE

Transmission of potentially multiplexed streams of voice, video and data over a communications channel can be controlled from a data prioritization management scheme as employed in QoS mechanisms. With the ability to transit digital voice and video over a digital packet switched network, QoS guarantees for space and ground communication networks are operational requirements. Similar to security-related concerns, mechanisms to provide the provision of QoS to an application reside at multiple layers of the OSI network stack including the Application Layer, the Transport and Network Layers, and ultimately via the Data Link or MAC sublayer. The IEEE 802.11, 802.15 and 802.16 wireless protocols and the RFID protocols that are ISO compliant provide QoS and Security provisions. To pragmatically design and access both QoS and security it is necessary to perform the analysis across the communication network stack spanning the Application Layer to the Physical Layer. This analysis is performed in the Wireless Working Group Magenta Books that are companion documents to this Green Book. A fundamental observation regarding QoS in networks is that often a network architect can provide QoS by engineering the network data rate capacities to provide ample margin, thereby ensuring QoS provisioning in practice for all network data flows, as is often done in telecommunications networks. This strategy is implementable pragmatically when the network is under complete control ('we own the network') of a single service provider. The counter argument to this philosophy is the practical realization that, given a network instantiation, usage of the network can nominally be expected to increase over time, thus necessitating QoS provision at some point to ensure

Application Layer requirements are met. Figure 5-15 depicts the reference Spacecraft Onboard Interface Services (SOIS) architecture: QoS and security provisioning can potentially take place within the User Applications, and/or at the SOIS Application Support, Transfer or Subnetwork Layer. Table 5-4 summarizes representative QoS provision mechanisms at different layers of the OSI protocol stack.

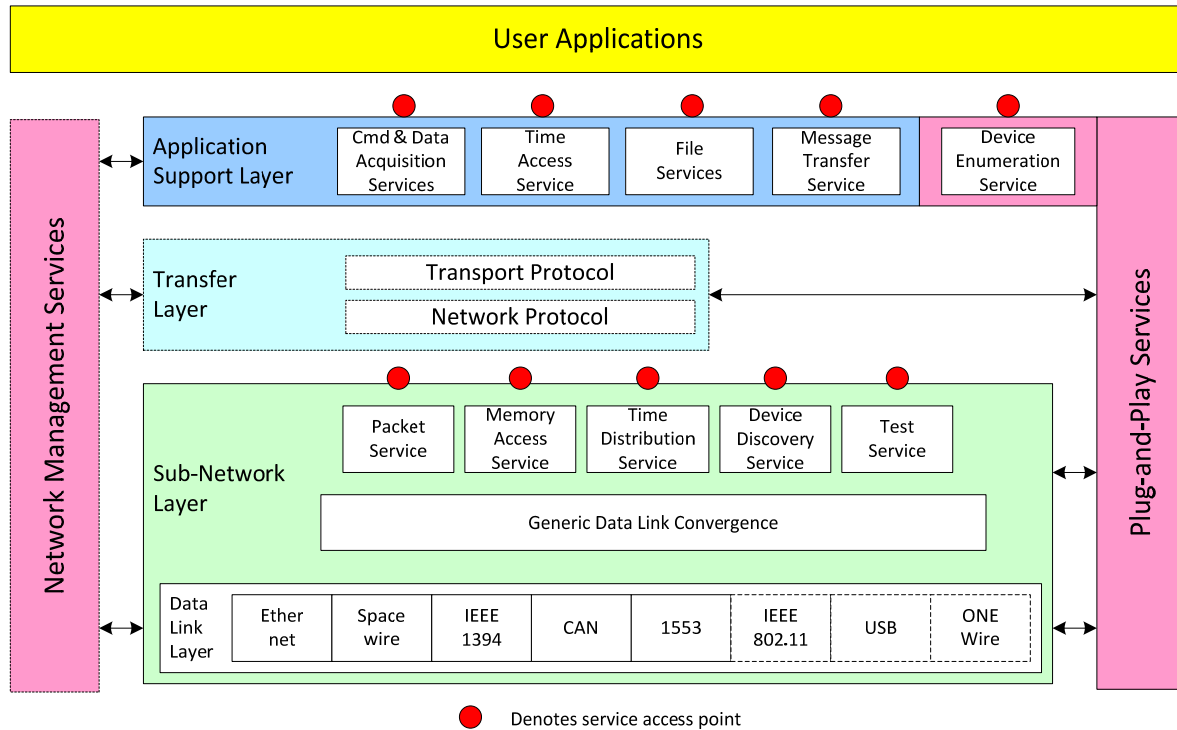


Figure 5-15: The Spacecraft Onboard Interface Services (SOIS) Architecture

5.2.2 SECURITY

Security of wireless data communications is important for space communications systems designers to address. The Wireless Working Group Magenta Books contain several threat analyses associated with usage in canonical operational scenarios. These threat analyses follow the prescribed assessment model and methodology as specified in CCSDS 350.1-G-1, *Security Threats against Space Missions* (reference [46]). Similar to QoS provisioning, security provision can span multiple layers of the OSI protocol stack, although an important difference to note is that security provision needs to be provided by just one layer of the OSI stack (e.g., IPsec for IP networks or BSP for DTN networks). Table 5-4 summarizes representative security provision mechanisms at different layers of the OSI protocol stack.

Table 5-4: Wireless LAN Security and Quality of Service Provisions

OSI Layer	Function	Protocols	Security Provision	QoS Provision
Application	Application data protection and consumption	Application	Application	Application
Presentation	Data representation	Middleware	Middleware-specific security provision to Application Layer	Middleware-specific QoS provision to Application Layer
Session	Interhost communications			
Transport	End-to-end transmission reliability	Transport UDP, TCP	TLS, SSL	RTP, DCCP, SCTP
Network	Addressing and routing	Network IP, DTN	IPSec, BSP	IntServ, DiffServ
MAC	Media access, frame transmission	IEEE 802.11 IEEE 802.15 IEEE 802.16	IEEE 802.11 IEEE 802.15 IEEE 802.16	IEEE 802.11 IEEE 802.15.1 IEEE 802.15.4e IEEE 802.16
PHY	Signaling, bit transmission	PHY-encoding & modulation	FHSS, DSSS, OFDMA	OFDMA

6 EMI/EMC CONCERNS FOR WIRELESS SPACE NETWORKS

6.1 OVERVIEW

This section relates EMI issues and the possible mitigation techniques to reduce their impacts onboard a spacecraft. This area needs to be thoroughly investigated; the integration of wireless networks within a spacecraft may cause disturbances with other instruments if interference source identification is not appropriately covered during the design phase.

This section presents a preliminary general assessment of frequency management issues to be reconsidered for each specific real mission application or scenario that utilizes wireless communications.

6.2 INTRODUCTION

EMI is the degradation in the performance of equipment due to the operation of another system and hence is the opposite of EMC.

Spacecraft commonly contain a number of transmitters and sensitive receivers and have to be electrically clean; that is to say onboard systems must not impair the operation of other onboard systems.

The introduction of wireless link radiation into any system requires foresight and preparation to ensure that sensitive circuitry is not affected. Suppression of potential conducted and induced noise at the wireless radiated frequencies (and harmonics) is important for onboard equipment and should be part of the specification of that equipment. If particularly sensitive equipment is susceptible to such frequencies, then choices will have to be made about how to mitigate such effects, whether by suppression, mutually exclusive operations, or acceptance of loss of performance, should that be possible. In some cases the selection of an alternative wireless frequency may be necessary.

In systems where there are multiple mission elements, such as may be found in spacecraft swarms or collaborating planetary surface components (e.g., rovers, landers), care must be taken to ensure that cross-element interference does not result in poorer performance of any of the elements unless this can be tolerated.

When discussing EMC or EMI, it is common to refer to an interfering transmitter as a *culprit* and a receiver that is interfered with as a *victim*.

An example of the band occupancy by a satellite is shown below in figure 6-1:

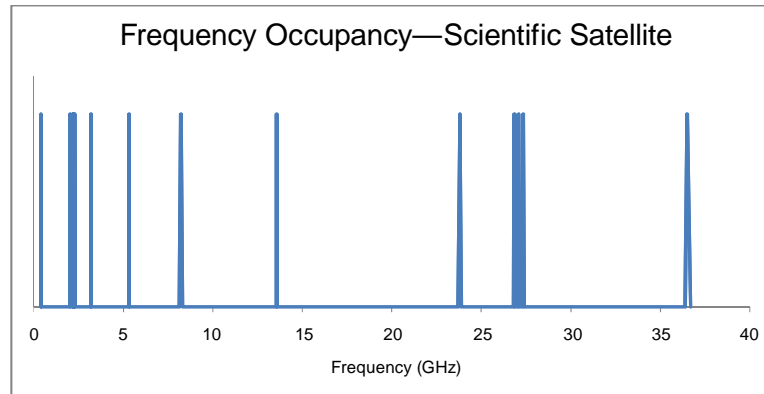


Figure 6-1: Typical Occupancy Band for a Satellite

Close to the wireless bands are found the Spacecraft and Launcher TM/TC bands, of which an example is shown below in figure 6-2:

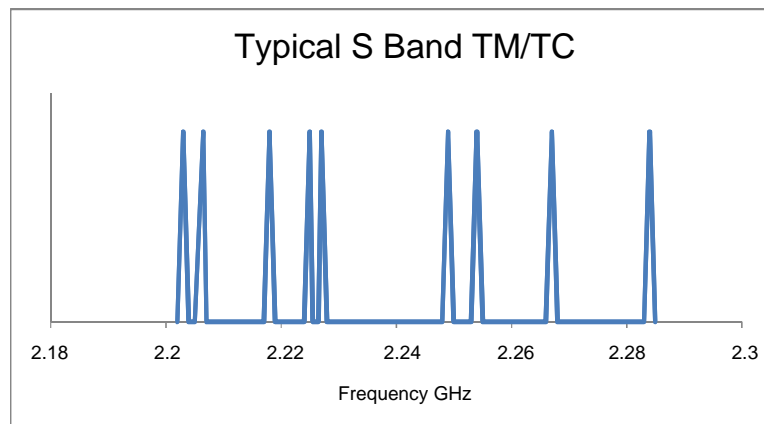


Figure 6-2: Spacecraft and Launcher TM/TC Bands

Interoperability could be achieved by all systems radiating and receiving only within their designated bands. Alternatively, many modern wireless systems are designed to interoperate within the same band. With either approach, there remain several mechanisms that can cause issues within a space-borne system, such as the following:

- a) **Out of Band Emissions.** All radiating systems will have some radiation out of band, such as harmonics of the radiating band, and leakage of intermediate frequencies or local oscillators. This can be true even of a receive-only system; as a terrestrial example, television detectors work by detecting the radiation of the local oscillator by the antenna. Careful filtering is required to reduce these out-of-band emissions to an acceptable level in the onboard environment.
- b) **Out of Band Sensitivity.** Although receivers have input protection, receivers have some sensitivity outside their operating band and sensitive receivers could have unexpected requirements. This was the cause in the Sheffield case.

- c) **Inter-Modulation Products.** Inter-modulation products give the worst problems in spacecraft EMC testing and have many methods of production. Common causes include the pickup of radiated components by poorly screened components, such as Printed Circuit Board (PCB) track or RF stubs being conductively coupled into mixers elsewhere and generating other frequencies. To avoid this it is necessary to thoroughly screen all parts carrying RF, and the use of stubs should be avoided where possible.

Certain precautions are standard in all RF packaging. The spacing of fixings that close boxes should be chosen to attenuate not only unwanted frequencies escaping, but also to attenuate incoming interfering frequencies.

It is important to ensure that any harmonics are filtered out to the noise level. There must be no intentional out of band emission. This may require the implementation of output filtering that is more stringent than that implemented in COTS systems.

It must be remembered that spacecraft receivers are generally more sensitive than terrestrial ones because of the propagation distances involved in radar or communications, or the sensitivity needed to measure microwave spectrometry with a radiometer. As an example, a Synthetic Aperture Radar (SAR) or radiometer receiver damage level below -40dBm (60dB down on the allowed 2.4 GHz output level) is not uncommon.

6.3 POTENTIAL RF COMPATIBILITY ISSUES

The main issues with 2.4 GHz systems revolve around interference with S-band systems. Previous tests of Bluetooth and 802.11b systems have shown no generated products in any S-band frequency range specified to be associated with launcher or spacecraft telemetry or telecommand. Any interference with such systems would be a result of intermodulation with signals of about 200 MHz, which of course could be associated with an intermediate frequency elsewhere on the spacecraft.

Another example concerns the Doppler Orbitography and Radio-positioning Integrated by Satellite (DORIS) system used on ENVISAT, TOPEX/POSEIDON, and others. The Doppler measurement frequency is 2.03625 GHz, and the ionospheric correction frequency is 401.25 MHz. Putting these together produces 2.43750 GHz, overlaid by band 6 of the 802.11g series (2.437 GHz center frequency), so a band 6 interferer mixed with the DORIS ionospheric correction frequency would come in directly on the Doppler measurement frequency, desensitizing or damaging the instrument. Similarly, intermodulation between the DORIS measurement frequency and 802.11g band 6 would produce 401.25 MHz, which not only is the DORIS ionospheric correction frequency but is also used by Search and Rescue and ARGOS systems.

Other possible victims of 2.4 GHz interference could be S-band SAR, though this is little used, or S-band altimetry (generally used as part of a dual frequency system). Again, this would be an intermodulation issue as these radars operate higher in S band, typically around 3.2 GHz.

Another issue that has to be considered is interaction or interference between wireless standards operating in the same area. Multiple Bluetooth systems will slow each other down, but the number of Bluetooth networks that can coexist is not determinable in such a simple fashion as the 802.11 cases, which have one network per non-overlapping channel for maximum throughput. Bluetooth systems all operate on the same frequencies and change in sequence so the effect of multiple networks is determined by settling time and channel occupancy Probability Density Function, modified by the presence or absence of Adjacent Channel Interference (ACI).

Bluetooth and 802.11b have been tested together and coexist, but the throughput of 802.11g products can depend on whether there are 802.11b products nearby. Performance is best in environments where an 802.11g AP is communicating only with 802.11g clients in a homogeneous WLAN. In these environments, the data rate within 20 meters is 54 Mb/s, and the throughput is 22–24 Mb/s when using TCP.

In addition to interference between different 802.11b and 802.11g systems, one must also consider interference between 2.4 GHz 802.11 systems and 802.15.4 low-power sensor networks operating in the same vicinity. For example, a number of studies have shown that 802.11 can seriously degrade 802.15.4 performance (see references [47], [48], [49], and [50]).

When considering 802.11a systems the main spacecraft concerns revolve around the 5.3–5.4 GHz space-borne SAR band and harmonic interference with the X-band SAR and direct to ground systems. This is a matter for careful filtering.

In Europe, the 802.11a system is allowed to operate providing Dynamic Frequency Selection (DFS) and Transmit Power Control (TPC) are implemented as specified in EN 301 893, UK Interface Requirement 2006, and IEEE 802.11h (Amendment 5: Spectrum and Transmit Power Management Extensions in the 5 GHz band in Europe). This is because of interference with radar systems such as C-band weather radars (land and air based), and ancillary resources, such as the Microwave Landing System, resulting in a need to listen before transmitting and moderate the output power.

DFS and TPC should not affect a system operating indoors in a well-screened environment, as the system should not be able to detect and respond to outdoor emissions. It does mean that integration halls would need to be carefully screened as the operation of DFS and TPC will slow down the 802.11a link by increasing the transfer overhead and reducing the link budget.

Approved European frequencies for the low-band system are from 5.180 GHz to 5.320 GHz, only allowed to operate indoors (not a problem for spacecraft integration!) with a maximum Equivalent Isotropically Radiated Power (EIRP) of 200 mW. The upper three bands (5.280 GHz, 5.300 GHz, 5.320 GHz) overlap legacy radar systems of ESA and ESA members (Radarsat-1 5.285 GHz to 5.315 GHz and ENVISAT 5.319 GHz to 5.339 GHz), though newer systems have moved fractionally higher: Radarsat-2 and Sentinel-1 are to occupy 5355 to 5455 MHz. It may be difficult to use this system with a C-band radar satellite, as the receivers are very sensitive and could be incapacitated by out-of-band emissions or

intermodulation products. Damage level for the unattenuated Sentinel-1 receiver is specified at -43 dBm in band, 66 dB down on the in-band power level of this system.

The upper band is license exempt, but still requires the implementation of DFS and TPC, and occupies the band 5.500 GHz to 5.700 GHz with a maximum EIRP of 1 W (30 dBm) at a maximum mean EIRP density of 50 mW/MHz in any 1 MHz band. This band is license exempt indoors or out, but all these frequencies are below the U.S. upper-band frequencies, though the lower-band frequencies are the same, so for a joint ESA-NASA project it would be logical to operate on lower band only.

6.4 GUIDANCE IN EMC/EMI DESIGN AND TEST

It is clear from the foregoing that spectral management of spacecraft could dictate not only which wireless systems to use, but which bands they operate on. In this area the 802.11 systems are probably better for spacecraft use because their frequency occupancy is stable and hence more predictable than the Bluetooth Frequency Hopping Spread Spectrum (FHSS) system. Therefore in the 802.11 systems the prediction, measurement, and containment of direct products and intermodulation products is more deterministic than that for Bluetooth, which switches frequency with time and thus might not show up an issue with a transient modulator in test until the wrong moment.

It is difficult to generalize to a larger extent, as electromagnetic compatibility has often been the subject of specific books. Two useful documents for further guidance in design and test are:

- a) Marshall Space Flight Center Electromagnetic Compatibility Design and Interference Control (MEDIC) Handbook (reference [51]) available from the NASA Technical Reports Server, <http://ntrs.nasa.gov/search.jsp>;
- b) Space Engineering—Electromagnetic Compatibility, ECSS-E-ST-20-07C published in July 2008 (reference [52]) available from the European Cooperation for Space Standardization website www.ecss.nl;
- c) Space Engineering—Electromagnetic Compatibility Handbook, ECSS-E-HB-20-07A published in September 2013 (reference [53]) available from the European Cooperation for Space Standardization website www.ecss.nl.

Both these documents refer to individual project documents as the ultimate control for a spacecraft. For any project, the spacecraft prime will always be ultimately responsible for ensuring EMC and thus dictating spectrum management, as only the prime or the controlling agency will have visibility of full spectrum occupancy for a spacecraft. A useful tool for calculating intermodulation products is the *RF Cascade Workbook*, an Excel spreadsheet available from www.rfcafe.com.

7 CONCLUSIONS AND RECOMMENDATIONS

Table 7-1: Key Application Areas for Intravehicle Space Communication Domains

Functional Domain	Application Areas	Number of nodes	Data Rate	Range	Applicable Standards
Intra-vehicle	Inventory monitoring	100s	Very Low	< 10 m	ISO 18000-6C EPCglobal
	Environmental monitoring (e.g., temperature, pressure, humidity, radiation, water quality)	10s to 100s	Low to Medium	< 100 m	802.15.4 802.15.4e ISA100.11a
	Physiological monitoring (includes EVA suit biomedical monitoring)	1 to 10	Low to Medium	< 100 m	802.15.1 802.15.4 802.15.4e ISA100.11a
	Crew member location tracking	1 to 10	Medium to High	< 300 m	802.11 802.15.3 802.15.4 802.16 LTE
	Structural monitoring	10s	Medium to High	< 300 m	802.11 802.15.3
	Intra-spacecraft communications (voice and video)	10s	Medium to High	< 300 m	802.15.1 802.11 802.16 LTE
	Process monitoring and automated control and Scientific monitoring and control	10s to 100s	Low to High	< 300 m	802.15.3 802.15.4 802.15.4e ISA100.11a 802.11 802.16 LTE
	Retro-fit of existing vehicle with new capabilities	10s to 100s	Low to High	10 m – 100 km	802.15.3 802.15.4 802.11 802.16 LTE
AIT activities	Spacecraft assembly, integration and test	10s to 100s	Medium	< 100 m	802.15.3 802.15.4 802.15.4e ISA100.11a 802.11
Inter-vehicle	Inter-spacecraft communications (voice, video and data)	10	High to extremely high	1 m – 100 km	802.16 LTE Prox-1 AOS
Planetary Surface	IVA-EVA, EVA-EVA, Habitat-to-LRV, LRV-crew communications (voice, video and data)	10	Medium to High	1 m – 50 km	802.11 802.16 LTE
	Robotic Operations	10s	Low to High	1 m – 50 km	802.15.3 802.15.4 802.11 802.16 LTE
Orbiter relay to Surface*	Surface-to-orbit communications (voice, video and data)	10	High to extremely high	> 200 km	LTE Prox-1 AOS
* Application areas not addressed in this Green Book					

This report has overviewed RF and optical wireless technologies and networks, which have the potential for utilization for space mission operations. Table 7-1, below, summarizes wireless technologies and corresponding areas of utilization within the intra-vehicle application domain. All of the standards-based technologies summarized in table 7-1 merit inclusion in an engineering trade analysis regarding potential wireless communications solutions. Any solution will be dependent upon mission requirements and constraints.

General Recommendation: Utilization of products that employ standards-based communications protocols is a key strategy to support internal and external mobile communications for space exploration. IEEE communication protocols are very mature, provide a defined upgrade path, directly support the IP protocol, and facilitate interoperability. Interoperability is necessary to improve reliability, reduce complexity, increase software and hardware reusability, and enable multi-developer or multi-agency support. Commercial products employing standards-based communications protocols provide increased reliability resulting from market competition and a deployment base that numbers in the millions. With the advance of commercial wireless technologies, wireless communications technologies are mature enough that COTS and IEEE products will *spin-in* to support wireless communications for space applications instead of the traditional technology *spin-out* from space agencies to the commercial market sector.

Specific recommended practices, relating to the above intra-vehicle wireless technologies, are given in two follow-on CCSDS Magenta Books:

- *Spacecraft Onboard Interface Services—RFID-Based Inventory Management Systems*. Issue 1. Recommendation for Space Data System Practices (Magenta Book), CCSDS 881.0-M-1. Washington, D.C.: CCSDS, May 2012.
- *Spacecraft Onboard Interface Systems—Low Data-Rate Wireless Communications for Spacecraft Monitoring and Control*. Issue 1. Recommendation for Space Data System Practices (Magenta Book), CCSDS 882.0-M-1. Washington, D.C.: CCSDS, May 2013.

ANNEX A

ACRONYMS

ACI	Adjacent Channel Interference
ACK	Acknowledgement
AIT	Assembly, Integration, and Test
AM	Amplitude Modulation
ANSI	American National Standards Institute
AP	Access Point
APD	Avalanche Photodiode
API	Application Programming Interface
APP	Application Layer
ARQ	Automatic Repeat Request
ASIC	Application Specific Integrated Chip
ASK	Amplitude-Shift Keying
AWGN	Additive White Gaussian Noise
BS	Base Station
BSS	Basic Service Set
BSP	Bundle Security Protocol
BWA	Broadband Wireless Access
CAN	Controller Area Network
CCSDS	Consultative Committee for Space Data Systems
CDM	Code Division Multiplexing
CDMA	Code Division Multiple Access
COTS	Commercial-off-the-shelf
CSMA-CA	Carrier-sense, Multiple Access-Collision Avoidance

CSMA-CD	Carrier-sense, Multiple Access-Collision Detection
DCCP	Datagram Congestion Control Protocol
DFS	Dynamic Frequency Selection
DOCSIS	Data Over Cable Service Interface Specifications
DRP	Distributed Reservation Protocol
DSSS	Direct Sequence Spread Spectrum
DTN	Delay Tolerant Networking
ECCS	European Cooperation for Space Standardization
ECG	Electrocardiogram
ECMA	European Computer Manufacturers Association
EEG	Electroencephalogram
EIRP	Equivalent Isotropically Radiated Power
EM	Electromagnetic
EMC	Electromagnetic Compatibility
EMI	Electromagnetic Interference
EPC	Electronic Product Code
ETSI	European Telecommunications Standards Institute
EVA	Extra-vehicular Activity
FCC	Federal Communications Commission
FDD	Frequency Division Duplex
FDM	Frequency Division Multiplexing
FDMA	Frequency Division Multiple Access
FHSS	Frequency Hopping Spread Spectrum
FM	Frequency Modulation
FMC	Fixed-Mobile Convergence

CCSDS HISTORICAL DOCUMENT
CCSDS REPORT CONCERNING INTEROPERABLE WIRELESS NETWORK COMMUNICATIONS

FOV	Field of View
FSK	Frequency-Shift Keying
FSO	Free Space Optics
HO	Hand-off
IDT	Interdigital Transducer
IEC	International Electrotechnical Commission
IEEE	Institute of Electrical and Electronics Engineers
IMS	Inventory Management System
IP	Internet Protocol
IPSec	Internet Protocol Security
IR	Infrared
IrDA	Infrared Data Association
ISI	Intersymbol Interference
ISM	Industrial, Scientific and Medical
ISO	International Organization for Standardization
ITU-R	International Telecommunication Union-Radiocommunications
IVA	Internal-vehicle Activity
LAN	Local Area Network
LD	Laser Diode
LED	Light Emitting Diode
LOS	Line of Sight
LR-WPAN	Low-Rate Wireless Personal Area Network
LRV	Lunar Rover Vehicle
LTE	Long-Term Evolution
MAC	Media Access Control

MB-OFDM	Multi-Band Orthogonal Frequency Division Multiplexing
MIMO	Multiple-input, multiple-output
MISO	Multiple-input, single-output
MS	Mobile Station
NFV	Network Function Virtualization
NIB	Non-interference Basis
NLOS	Non-Line-of-Sight
NWK	Network Layer
OBDH	Onboard Data Handling
OFDM	Orthogonal Frequency Division Multiplexing
OOK	On-Off Keying
PAL	Protocol Adaptation Layer
PAN	Personal Area Network
PCA	Priority Contention Access
PCB	Printed Circuit Board
PCM	Pulse Code Modulation
PDA	Personal Digital Assistant
PHY	Physical
PM	Phase Modulation
PMP	Point-to-Multipoint
PN	Pseudonoise
PPM	Pulse Position Modulation
PSK	Phase-Shift Keying
QAM	Quadrature Amplitude Modulation
QoS	Quality of Service

CCSDS HISTORICAL DOCUMENT
CCSDS REPORT CONCERNING INTEROPERABLE WIRELESS NETWORK COMMUNICATIONS

QPSK	Quadrature Phase-Shift Keying
RF	Radio Frequency
RFID	Radio Frequency Identification
RSVP	Resource Reservation Protocol
RTP	Real-time Transport Protocol
RV	Rover Vehicle
SAR	Synthetic Aperture Radar
SAW	Surface Acoustic Wave
SCTP	Stream Control Transmission Protocol
SDM	Space Division Multiplexing
SDMA	Space Division Multiple Access
SDN	Software-Defined Networking
SIS	Space Internetworking Services
SIMO	Single-input, multiple-output
SIR	Signal-to-Interference ratio
SISO	Single-input, single-output
SLS	Space Link Services
SNR, S/N	Signal-to-Noise ratio
SOIS	Spacecraft Onboard Interface Services
SS	Subscriber Station
TBD	to be determined
TCD	Temperature Coefficient of Delay
TCP	Transmission Control Protocol
TDD	Time Division Duplex
TDM	Time Division Multiplexing

TDMA	Time Division Multiple Access
TPC	Transmit Power Control
TSMP	Time Synchronized Mesh Protocol
UDP	User Datagram Protocol
UNII	Unlicensed National Information Infrastructure
UPC	Universal Product Code
UWB	Ultra Wide Band
WLAN	Wireless Local Area Network
WMAN	Wireless Metropolitan Area Network
WORM	Write-Once, Read-Many
WPAN	Wireless Personal Area Network
WSN	Wireless Sensor Network
WWG	Wireless Working Group
ZED	ZigBee End Device
ZR	ZigBee Router

ANNEX B

GLOSSARY

active tag. A type of RFID tag that contains an internal power source, and in some cases also a radio transceiver. These additional component(s) are used to enhance the effective read/write range and rate of data transfer characteristics of the RFID tag. This type of integrated tag circuit is usually of a complex design with many components. Active tags can transmit over the greatest distances (30+ meters).

ADC. Automated Data Collection.

ad hoc. A network typically created in a spontaneous manner. An ad hoc network requires no formal infrastructure and is limited in temporal and spatial extent.

agile reader. A reader that can read different types of RFID tags, either made by different manufacturers or operating on different frequencies.

antenna. A device for sending or receiving electromagnetic waves.

anti-collision. A feature of RFID systems that enables a batch of tags to be read in one reader field by preventing the radio waves from interfering with one another. It also prevents individual tags from being read more than once.

attenuation. The reduction in amplitude or strength of a signal as a function of distance.

automatic data capture, ADC. Methods of collecting data and entering it directly into a computer system without human intervention. Automatic Identification (Auto-ID) Refers to any technologies for capturing and processing data into a computer system without using a keyboard and includes bar coding, RFID, and voice recognition.

Auto-Id Center. A group of potential RFID end users, technology companies, and academia. The Auto-ID Center began at the Massachusetts Institute of Technology (MIT) and is now a global entity. It is focused on driving the commercialization of ultra-low cost RFID solutions that use Internet-like infrastructure for tracking goods throughout the global supply chain. The Auto-ID Center organization is now called EPCglobal.

backscatter. A method of RF propagation onboard an RFID tag.

bandwidth. The difference in Hertz between the upper and lower limiting frequencies of a spectrum.

BiStatix. A type of RFID tag design, where the enclosed circuit is manufactured using printable conductive inks and silicon layering.

bit. The smallest unit of digital information; in binary code, a single '0' or '1'. A 96-bit EPC is a string of 96 zeroes and ones.

byte. Eight bits. One byte of memory is needed to generate an alpha character or digit. So bytes can be thought of in terms of characters.

carrier wave. A continuous frequency capable of being modulated with a second (baseband or information-carrying) signal.

chip based RFID. RFID tags that contain a silicon computer chip and therefore can store information and transmit it to a reader.

collision. Radio Signals interfering with one another. Signals from tags and readers can collide.

die. A tiny square of silicon with an integrated circuit etched on it, more commonly known as a silicon chip.

Differentiated Services, DiffServ. A computer networking architecture that specifies a simple, scalable, and coarse-grained mechanism for classifying and managing network traffic and for providing Quality of Service (QoS) guarantees on modern IP networks.

electronic article surveillance tags, EAS tags. Single bit (either 'on' or 'off') electronic tags used to detect items for anti-theft purposes. EAS technology is similar to RFID in that it uses similar frequency bands.

electromagnetic compatibility, EMC. The ability of a technology or product to coexist in an environment with other electro-magnetic devices.

Electronic Product Code, EPC. A standard format for a 96-bit code that was developed by the Auto-ID Center. It is designed to enable identification of products down to the unique item level. EPCs have memory allocated for the product manufacturer, product category, and the individual item. The benefit of EPCs over traditional bar codes is their ability to be read without line of sight and their ability to track down to the individual item versus at the SKU level.

EPCglobal. The association of companies that are working together to set standards for RFID in the retail supply chain. EPCglobal is a joint venture between EAN International and the Uniform Code Council, Inc.

far field. An operating specification for an RFID tag to have a read / write range of greater than one meter.

frequency. A band of operation for radio-based technologies. Frequencies allocated for RFID use exist in the low, high, ultra-high, and microwave frequency bands. Each frequency has its own advantages and disadvantages, such as read distance, tag size, and resistance to electronic noise.

gen 2. The second generation global protocol operating in the UHF range. The current choice for many retail supply chain carton and pallet compliance applications, starting in 2006.

Global Trade Item Number, GTIN. A superset of bar code standards that is used internationally. In addition to manufacturer and product category, GTIN also includes shipping, weight, and other information. The EPC is designed to provide continuity with GTIN.

group selection. A mode of operation whereby an interrogator can search for and identify unique tags within an RF portal or RF field of view.

Global Tag, GTAG. A standardization initiative of the Uniform Code Council (UCC) and the European Article Numbering Association (EAN) for supply-chain tracking applications using UHF RFID frequencies.

high-frequency RFID (13.56 MHz). RFID that uses the high-end 13.56 MHz radio frequency band and features medium sized tags with relatively good reading distances. In the U.S., 13.56 MHz tags can be typically read at approximately 3–4 inches with a handheld reader and 1.5 to 2 meters with a portal reader.

integrated circuit, IC. Another name for a chip or microchip.

interrogator. A device that is used to read and or write data to RFID tags.

Integrated Services, IntServ. An architecture that specifies the elements to guarantee Quality of Service (QoS) on networks.

line-of-sight. Technology that requires an item to be ‘seen’ to be automatically identified by a machine. Unlike bar codes and OCR technologies, RFID tags can be read ‘through’ merchandise and most packaging with no line of sight required.

low-cost RFID. RFID tags that cost less than \$0.50 with typically one meter of read range.

low-frequency RFID (125 & 134 kHz). Low frequency radio band allocated for RFID use. The main disadvantage of low frequency RFID is its cost and relatively slow data transfer as well as its inability to read many tags at the same time.

multiple tag read/write. Reading and writing of multiple RFID tags at the same time. Reading and writing of multiple tags is achieved through the anti-collision feature of RFID.

microwave RFID frequency (2,450 MHz or 2.45 GHz). A microwave frequency band allocated for RFID use, used for item-level tracking, including retail merchandise. Typically microwave RFID technologies feature the smallest label footprint and read distances up to 18 inches with a handheld reader and perhaps up to 1–1.5 meters with a portal reader. This frequency also offers fast data transmission but is somewhat more bothered by shielding of liquid products and reflections from metal structures, etc.

near field. An operating specification for an RFID tag to be near or in close proximity to an interrogator’s antenna. Near field capable interrogators and corresponding RFID tags typically have a read / write range of 4–6 inches.

passive RFID tag. An RFID tag that does not use a battery. Passive tags draw their power from the reader. The reader transmits a low power radio signal through its antenna. The tag in turn receives it through its own antenna to power the integrated circuit (chip). Using the energy it gets from the signal, the tag will briefly converse with the reader for verification and the exchange of data. As a result, passive tags can transmit information over shorter distances (typically three meters or less) than active tags.

perpetual inventory. The ability to know one's inventory position at any given time. RFID offers the promise of being able to perform automatic inventory counts.

radio frequency identification, RFID. A method of identifying items uniquely using radio waves. Radio waves do not require line of site and can pass through materials like cardboard and plastic but not metals and some liquids.

read range. The distance from which a reader can communicate with a tag. Several factors including frequency used, orientation of the tag, power of the reader, and design of the antenna affect range.

reader. An interrogator. The RFID reader communicates via radio waves with the RFID tag and passes information in digital form to the computer system. Readers can be configured with antennas in many formats including handheld devices, portals or conveyor mounted.

read-only tags. Tags that contain data that cannot be changed. Read-only chips are less expensive than read-write chips.

read-write tags. RFID chips that can be read and written multiple times. Read-write tags can accept data at various points along the distribution cycle. This may include transaction data at the retail point of sale. They are typically more expensive than read-only tags but offer more flexibility.

RF absorption. A radio phenomenon that occurs when transmitted RF signal energy is consumed or rapidly dispersed by some material in the pathway of the RF transmission.

RF cancellation. A radio phenomenon that occurs where a transmitted RF signal is neutralized by competing RF interference.

RF frequency. A defined radio protocol to transmit and receive data. RFID frequency types include 2.45 GHz, 915 MHz, 13.56 GHz, and 125 kHz.

RF reflection. A radio phenomenon that occurs when a transmitted RF signal is echoed off of another RF radiator placed within the pathway of the RF transmission.

radio frequency data collection, RFDC. An implementation of automated data collection whereby portable ADC reader devices are connected to a host computer via RF so that interactive data transfers can occur.

RFID. A means of storing and retrieving data via electromagnetic transmission to a radio frequency-compatible integrated circuit.

RFID site survey. A comprehensive analysis to determine or confirm that a proposed RFID solution meets the intended application requirements and technology specifications of use. It also defines the equipment needed to implement a proposed RFID system and outlines the responsibilities of each party involved with the system implementation.

RFID transponder. Another name for an RFID tag. Typically refers to a microchip that is attached to an antenna, which communicates with a reader via radio waves. RFID tags contain serial numbers that are permanently encoded, allowing them to be uniquely identified. RFID tags vary widely in design. They may operate at one of several frequency bands, may be active or passive, and may be read-only or read-write.

RF portal. A defined physical area of RF signal saturation, also known as an RF depth of field and/or physical RF field of view.

smart label. A label that contains an RFID chip and antenna. These labels can store information, such as a unique serial number, and communicate with a reader.

spread spectrum. A technique in which the information in a signal is spread over a wider bandwidth using a spreading code.

tag. The generic term for a radio frequency identification device. Also sometimes referred to as smart labels.

tag collision. Interference caused when more than one RFID tag sends back signals to the reader at the same time.

transponder. A type of integrated circuit designed to store data and respond to RF transmissions of a given frequency. A transponder is another name for an RFID tag.

ultra-high frequency RFID (850 to 950 MHz). UHF radio band allocated for RFID use. UHF RFID can send information faster and farther than high- and low-frequency tags. UHF RFID is gaining industry support as the choice bandwidth for inventory tracking applications including pallets and cases. UHF RFID features larger tags and readers with the longest read distances (1 meter with handheld readers and more than 3 meters with portal readers).

write broadcast capability. An RFID technology characteristic that allows data to be written to multiple tags while those tags are within an RF portal.

write once read many chip, WORM chip. Chip that can be written once and then becomes read-only afterwards.

ANNEX C

WIRELESS STANDARDS AND RF QUICK REFERENCE

The following quick-reference tables are a concise summary of the following topics:

- IEEE WPAN, WLAN, and WMAN standards activities;
- Detailed IEEE WPAN and WLAN specifications summary;
- ITU Industrial, Scientific, and Medical RF band designations; and
- Commonly used RF Band designations.

The tables are presented in a single annex for ease of future reference.

Table C-1: IEEE 802.11 Standards and Working Group Activities

IEEE 802.11 Standard	Description	Status (as of June, 2013) ¹²
IEEE 802.11-2012	MAC and PHY specifications; up to 600 Mb/s; 2.4 and 5 GHz; MIMO support for up to 4 spatial streams.	Approved 2012; incorporates all previous amendments including 802.11a,b,g,i,n,s among others (see NOTE below).
IEEE 802.11ac	Enhancements for Very High Throughput for Operation in Bands below 6 GHz; up to 6.9333 Gb/s; 5 GHz only; MIMO support of up to 8 spatial streams and multi-user MIMO.	Unapproved draft (D5.00)
IEEE 802.11ad-2012	Enhancements for Very High Throughput in the 60 GHz Band; up to 7.75675 Gb/s; support for directional beamforming.	Approved 2012
IEEE 802.11aa-2012	MAC Enhancements for Robust Audio Video Streaming	Approved 2012
IEEE 802.11ae-2012	Prioritization of Management Streams	Approved 2012
IEEE 802.11ak	General Link Study Group	No document
IEEE 802.11aq	Pre-Association Study Group	No document
IEEE 802.11aj	China Millimeter Wave Study Group	No document
IEEE 802.11ai	Fast Initial Link Set-Up	Unapproved draft (D0.30)
IEEE 802.11ah	Sub 1 GHz Study Group	No document
IEEE 802.11af	TV White Spaces Operation	Unapproved draft (D2.20)

¹² See reference [32].

NOTE – IEEE Std 802.11-2012, incorporates the following amendments into the 2007 revision:

- IEEE Std 802.11k™-2008: Radio Resource Measurement of Wireless LANs (Amendment 1)
- IEEE Std 802.11r™-2008: Fast Basic Service Set (BSS) Transition (Amendment 2)
- IEEE Std 802.11y™-2008: 3650–3700 MHz Operation in USA (Amendment 3)
- IEEE Std 802.11w™-2009: Protected Management Frames (Amendment 4)
- IEEE Std 802.11n™-2009: Enhancements for Higher Throughput (Amendment 5)
- IEEE Std 802.11p™-2010: Wireless Access in Vehicular Environments (Amendment 6)
- IEEE Std 802.11z™-2010: Extensions to Direct-Link Setup (DLS) (Amendment 7)
- IEEE Std 802.11v™-2011: IEEE 802.11 Wireless Network Management (Amendment 8)
- IEEE Std 802.11u™-2011: Interworking with External Networks (Amendment 9)
- IEEE Std 802.11s™-2011: Mesh Networking (Amendment 10)

Table C-2: IEEE 802.15 Standards and Working Group Activities

IEEE 802.15 Standard	Description	Status (as of May, 2012)
IEEE 802.15.1	WPAN; up to 1 Mb/s; 2.4 GHz	Approved 2002 as IEEE Std. 802.15.1TM-2002; development transitioned to Bluetooth Special Interest Group
IEEE 802.15.2	WPAN and WLAN coexistence; 2.4 GHz	Approved 2003; group in hibernation
IEEE 802.15.3	HR-WPAN; 11–55 Mb/s; 2.4 GHz	P802.15.3TM Draft Standard complete
IEEE 802.15.3a	110 Mb/s UWB PHY Layer; considered OFDM-UWB and DS-UWB	PAR withdrawn
IEEE 802.15.3b	MAC implementation and interoperability enhancements	Little progress since 2004
IEEE 802.15.3c	mmWave WPAN; 2 Gb/s; 57–64 GHz	Approved 2009 as IEEE St. 802.15.3c TM -2009; group placed into hibernation
IEEE 802.15.4	LR-WPAN; 20–250 kb/s, 850 kb/s, 6.81 Mb/s, 27.24 Mb/s; 868, 915, 2400 MHz; long battery life; precision-ranging and higher data-rates with UWB PHY.	Approved 2003; updated by IEEE 802.15.4-2006 and IEEE 802.15.4-2011. Incorporates IEEE 802.15.4a amendment
IEEE 802.15.4a	Precision ranging LR-WPAN; UWB precision ranging @ 2.4 GHz	Approved as IEEE Std. 802.15.4a-2007; superseded by IEEE Std. 802.15.4-2011; slow commercial pick-up
IEEE 802.15.4b	Enhancements to 802.15.4	Status uncertain
IEEE 802.15.4c	Alternative PHY for China	Approved as IEEE Std. 802.15.4c-2009; superseded by IEEE Std. 802.15.4-2011
IEEE 802.15.4d	Alternative PHY for Japan	Approved as IEEE Std. 802.15.4d-2009; superseded by IEEE Std. 802.15.4-2011
IEEE 802.15.4e	Add functionality to IEEE Std. 802.15.4-2011 MAC to better support industrial markets	Approved as IEEE Std. 802.15.4e-2012
IEEE 802.15.4f	Active RFID—define new PHY and modifications to MAC to support RFID	Approved as IEEE Std. 802.15.4f-2012
IEEE 802.15.4g	Smart utility networks	Approved as IEEE Std. 802.15.4g-2012
IEEE 802.15.4j	Add 2.360 MHz–2.400 MHz PHY to IEEE Std. 802.15.4-2011	Approved as IEEE Std. 802.15.4j-2013
IEEE 802.15.4k	Add PHY in support of point to multi-thousands of points to IEEE Std. 802.15.4-2011	Draft standard
IEEE 802.15.4m	Add PHY to IEEE Std. 802.15.4 in support of TV white space operation	Pre-draft stage
IEEE 802.15.4n	Add PHY/MAC to IEEE Std. 802.15.4 in support of 174–216 MHz, 407–425 MHz, and 608–630 MHz medical bands in China.	Pre-draft stage
IEEE 802.15.4p	Amend IEEE Std. 802.15.4 in support of rail transit applications	Pre-draft stage

IEEE 802.15 Standard	Description	Status (as of May, 2012)
IEEE 802.15.4q	Amend IEEE Std. 802.15.4 to include an ultra low power PHY	Pre-draft stage
IEEE 802.15.5	WPAN Mesh networking	Approved as IEEE Std. 802.15.5-2009
IEEE 802.15.6	Body Area Networks (BANs)	Approved as IEEE Std. 802.15.6-2012
IEEE 802.15.7	PHY and MAC standard for Visible Light Communications (VLC)	Approved as IEEE Std. 802.15.7-2011
IEEE 802.15.8	PHY and MAC standard for WPAN peer-aware communications	Pre-draft stage.
IEEE 802.15.9	Recommended practice for key management protocol datagram exchange framework.	Pre-draft stage.

Table C-3: IEEE 802.16 Standards and Working Group Activities

IEEE 802.16 Standard	Description	Status (as of September, 2013)
IEEE 802.16	WMAN; OFDM; 96–134 Mb/s; 2–11 and 10–66 GHz; QoS & security in standard	Approved 2004; Latest version 2012
IEEE 802.16h	Improved coexistence mechanisms	Approved 2010
IEEE 802.16j	Multihop relay specification	Approved 2009
IEEE 802.16k	MAC-sublayer Bridging	Approved 2007
IEEE 802.16n	Higher Reliability Networks	Approved 2013
IEEE 802.16p	Enhancements to Support Machine-to-Machine Applications	Approved 2012
IEEE 802.16.1	Advanced Air Interface; 100 Mb/s for mobile and 1 Gb/s for fixed	Approved 2012
IEEE 802.16.1a	Advanced Air Interface—Higher Reliability Networks	Approved 2013
IEEE 802.16.1b	Advanced Air Interface—Enhancements to Support Machine-to-Machine Applications	Approved 2012
IEEE 802.16.2	Coexistence of Fixed Broadband Wireless Systems	Approved 2003; Published as IEEE Std 802.16.2-2004

Table C-4: 3GPP Specifications Series

3GPP Specification Series	Subject Area of Series
21 series	General Requirements
22 series	Service Aspects (3GPP Stage 1)
23 series	Technical realization (3GPP Stage 2)
24 series	Signaling protocols (3GPP Stage 3) for user equipment (UE) to network
25 series	Radio system aspects
26 series	Audio/video encoding
27 series	Data
28 series	3GPP Stage 3 protocols for Core Network, plus some operations and management, and charging specifications
29 series	3GPP Stage 3 protocols for communication between fixed networks
31 series	Subscriber Identity Module (SIM) specifications
32 series	Primary operations and management, and charging, specifications
33 series	Security
34 series	UE and SIM testing
35 series	Security algorithms
36 series	LTE and LTE-Advanced radio technology
37 series	Aspects of multiple radio access network technologies

Table C-5: Unlicensed RF Bands

Frequency Range*	Center Frequency
6.765–6.795 MHz	6.780 MHz
13.553–13.567 MHz	13.560 MHz
26.957–27.283 MHz	27.120 MHz
40.66–40.70 MHz	40.68 MHz
433.05–434.79 MHz	433.92 MHz
902–928 MHz	915 MHz
2.400–2.500 GHz	2.450 GHz
5.15–5.35 GHz	5.25 GHz
5.47–5.825	5.6475 GHz
5.725–5.875	5.8 GHz
24–24.25 GHz	24.125 GHz
57.38–65.74 GHz	61.56 GHz
122–123 GHz	122.5 GHz
244–246 GHz	245 GHz
* Wireless networking communications equipment use of unlicensed bands is on a non-interference basis (NIB).	

NOTE – The ITU ISM bands designation is, from a correctness perspective, only strictly applicable to terrestrial wireless communications deployments. It *may* be that these designations will hold also for space-based wireless systems, but that is yet to be determined.

Table C-6: NATO or Electronic Warfare RF Band Designations

Radar Designation	ITU Designation	IEEE Designation	Wireless Bands
HF 3–30 MHz	HF 3–30 MHz	A 0–250 MHz	
Not designated	VHF 30–300 MHz	B 250–500 MHz	
P 216–450 MHz	UHF 300–3000 MHz	C 500–1000 MHz	802.15.4
Not designated		D 1–2 GHz	
L 1–2 GHz		E 2–3 GHz	802.11 802.15.1/Bluetooth, 802.15.4
S 3–4 GHz		F 3–4 GHz	
	SHF 3–30 GHz	G 3–6 GHz	802.11
C 3–8 GHz		H 6–8 GHz	
		I 8–10 GHz	
X 8–12.4 GHz		J 10–20 GHz	
J / Ku 12.4–18 GHz		K 20–40 GHz	
K 18–26.5 GHz			
Q / Ka 26.5–40 GHz		V 50–75 GHz	802.11ad
	EHF 30–300 GHz		

ANNEX D

AUTOMATED LOGISTICS MANAGEMENT USE CASES

D1 INTRODUCTION

Identified wireless communications use cases for CCSDS agency members are summarized, typically one per page, in the following subsections.

D2 INTRA-HABITAT EQUIPMENT/LRU

Objective: Localize equipment and LRUs:

- portals or zone interrogators track equipment ingress/egress from habitat sections and rooms;
- scanned zone interrogator can provide real time tracking within coverage area.

D3 INTRA-HABITAT CONSUMABLES

Objective: Augmentation for inventory management and situational awareness:

- packaging on consumables contains RFID tag;
- refuse container interrogators read package tag and update item inventory and kills tag;
- RFID database application provides warning if product expires before item appears in trash;
- range < 1/3 meter.

D4 INTRA-HABITAT MEDICAL SUPPLIES

Objective: Inventory management, localization, and situational awareness:

- inventory management for medical instruments, supplies, and pharmaceuticals;
- provide expiration warnings, particularly for pharmaceuticals;
- provide verification or warning relating to missed administration, or dosage, of medications;
- range < 1/3 meter.

D5 SMART CONTAINERS

Description: ‘Smart containers’ can provide enhanced RFID functionality, and definitions vary. One capability attributed to ‘smart containers’ is the local storage of data about the contents. Other ‘smart containers’ interrogate local tags that are typically confined to the container, and then report that data to an exterior interrogator or network.

D6 DEEP FREEZER SAMPLES

Description: RFID could be used to manage the samples stored in the deep freezer device on the ISS. Barcodes are inappropriate because of the frosting and readability problems.



Figure D-1: MELFI Cooling System Onboard the ISS

D7 BATTERY MANAGEMENT

Description: Storing life data on batteries can simplify and ease battery management. The usage of partly loaded or over-aged batteries for experiments and tools can be avoided, e.g., on a space station.

D8 TECHNICAL CHECKS

Description: Using RFID tags fixed on checkpoints can enhance the accomplishment of technical checks. The check is automatically logged, identification of checkpoints is eased and additional data can be supplied to the personnel. RFID-tags with analogue or digital inputs can supply further information, e.g., on pressure, crack propagation, etc.

D9 PART IDENTIFICATION

Objective: immediate recognition of multitude of parts and association to database.

Description: tags on element parts (e.g., wires) provide immediate identification and association with database description, connectivity, calibration information, known location, part history, wire time domain signatures, etc. A portable, handheld interrogator would typically access this tag.

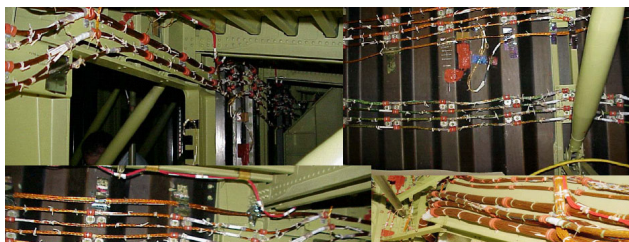


Figure D-2: Cable Runs Interior to the Shuttle

Range	Near-field, < 1/3 meter
Reader type	Portable (handheld)
Readability:	100 percent

D10 PASSIVE SENSOR TAG TELEMETRY

Description: Some RFID tags, including passive, active, and ‘semi-active’ can also provide sensor telemetry to the interrogator. Figure D-3 shows passive RFID/sensor tags attached to a liquid oxygen fuel tank. The tags are interrogated remotely at the launch site and return temperature and identification, which indirectly convey information regarding fuel levels during the tanking process. Obvious advantages include the absence of wire connections, tethers, and batteries.



Figure D-3: Passive Temperature RFID/Sensor Tags on Rocket Fuel Tank



Figure D-4: RFID Telemetry Could Increase EVA Safety via Simple Retrofit

A particular example of the important of future RFID telemetry use is the ability to retrofit and extend present mission- and life-critical telemetry. A case in hand was the July 16th 2013 EVA emergency on ISS, in which EVA crewmember Luca Parmitano's life was placed in danger because of a water leak inside his EVA suit (figure D-4). If the suit had been provided with mass RFID telemetry, at very low mass/volume and zero power impact, it may have been possible to diagnose the severity of the problem during the EVA. Indeed, such RFID telemetry would also provide for pre-EVA and post-EVA checkout and diagnosis. Additionally, ISS and future vehicles such as Orion could be retro-fitted / outfitted with large-scale RFID telemetry systems to allow crew and robotics to maintain detailed ISS systems, including vehicle skin integrity and leak localization, operational and status verification checks, again with low mass, volume and power impact. Such retrofitting could be accomplished by simply affixing the sensors to the correct location via adhesive double-sided tape, and by using small handheld reader systems to interrogate and power the sensors, with no added power and data wiring.

D11 RFID ENHANCED CONNECTORS

Description: RFID can be used to ensure that a connector is connected to the correct slot. The connector has an RFID tag, the technician queries the tag with a pen-like, millimeter range reader and the configuration gets verified. This can also be applied to the connection of non-electronic elements, e.g., fluid- or gas-carrying pipes in biological experiments.

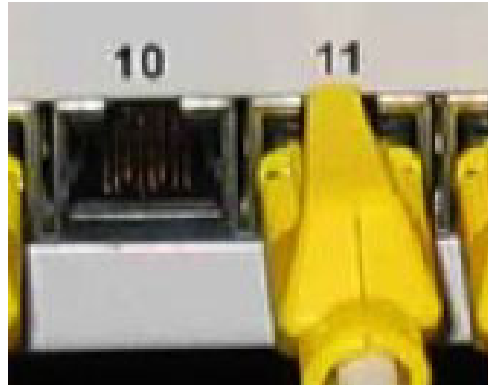


Figure D-5: RFID Enhanced Connectors

D12 RFID ENHANCED BOLT IDENTIFICATION

Description: During fastening of a bolt, an ultrasonic wave technology is used to measure its elongation. To be achievable, the bolt must be identifiable and the calibration data must be acquirable. Current procedures use barcode for bolt identification and a database for the related data. RFID would permit to locally store the ID and the required calibration data directly on the bolt.

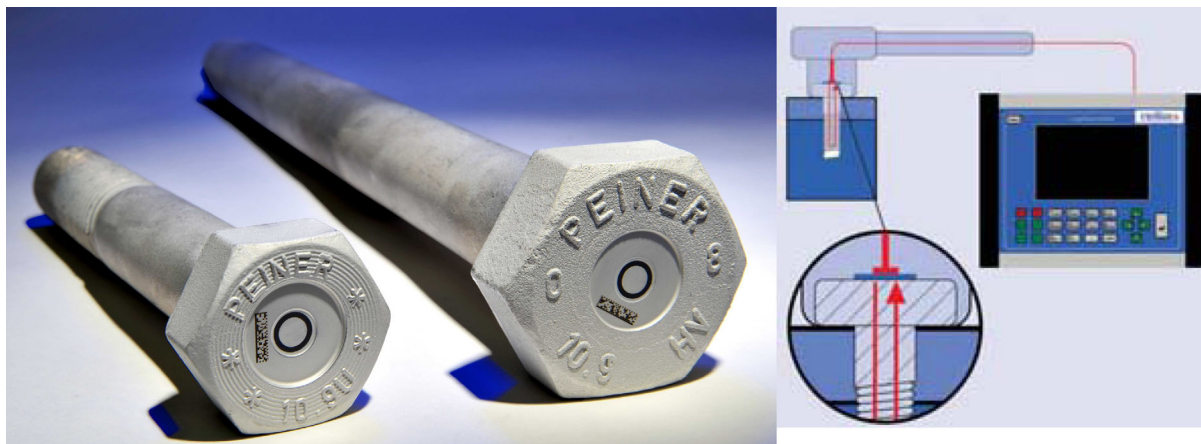


Figure D-6: RFID Bolt Identification

D13 RFID ENHANCED TORQUE SPANNER

Description: A bolt contains the recorded data (e.g., angle, date, torque) of a screwed joint. With an electronic torque wrench equipped with an RFID reader, the wrench could discover the required settings and could adjust itself automatically.



Figure D-7: RFID Torque Spanner

D14 HABITAT PROXIMITY ASSET

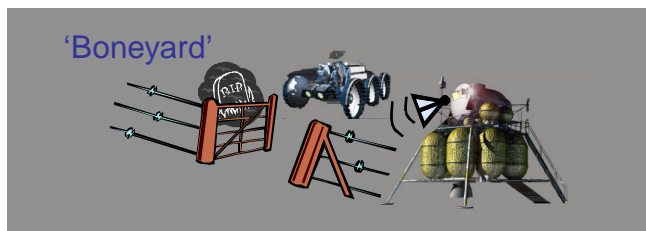


Figure D-8: Habitat Proximity Asset Localization Concept

Objective: Inventory management, localization, and situational awareness:

- provides rapid localization of external assets, equipment, and tools between habitats, tool crib;
- SMUs, rovers, bone yard, etc.;
- larger ranges, up to and possibly exceeding 200 ft.;
- reader type: portal, vehicle mounted, scanned, and/or fixed beam;
- gatekeeper: zone or portal interrogator monitors bone yard;
- spent elements serve as repository for parts;
- gatekeeper is powered by, and possibly located on or near, spent lander.

D15 SCIENCE SAMPLE INVENTORY MANAGEMENT

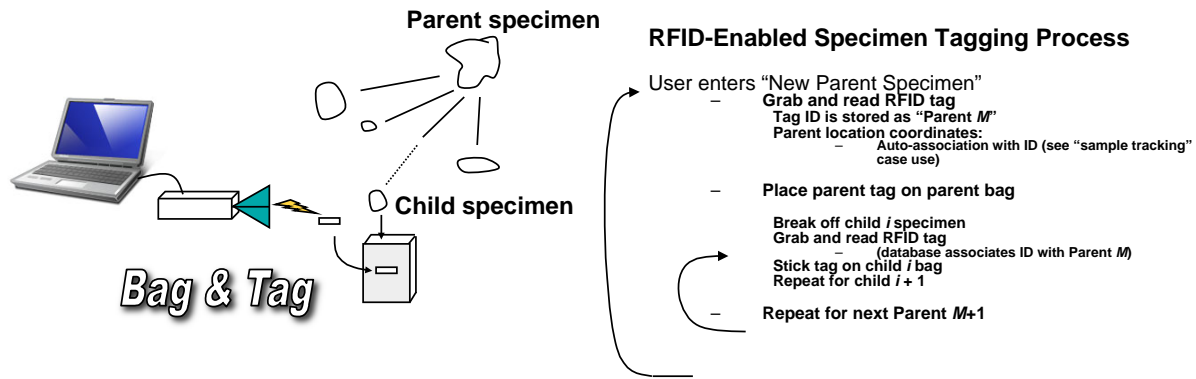


Figure D-9: Science Sample Inventory Management Concept

Objective: Track heritage (parent specimens):

- Inventory management of collected samples;
- special: requires on-site tagging (preprinted tags or portable printer).

Range	1–2 meters
Reader type	Portable (handheld)
Readability:	100 percent

D16 SCIENCE SAMPLE POSITION DETERMINATION

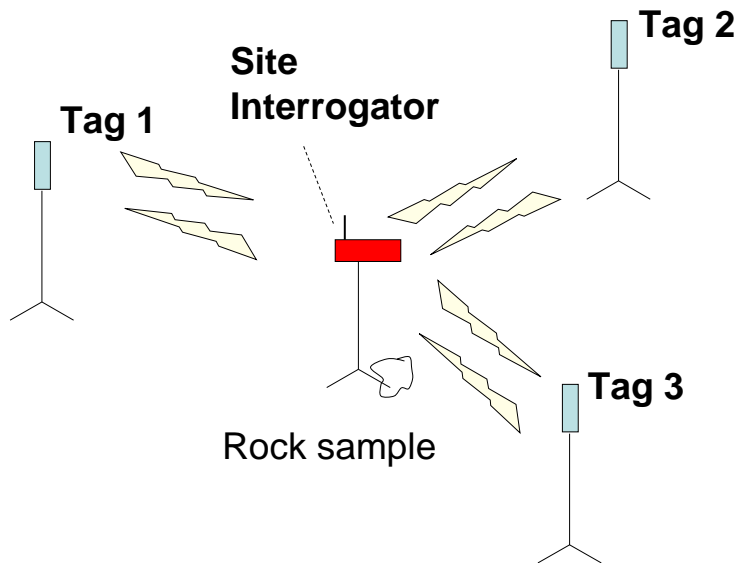


Figure D-10: Science Sample Position Determination Concept

Objective: Provide absolute location of samples within 1 m:

- dependent upon other means to accurately survey boundary tag positions;
- special: requires interrogator (at sample site) + local survey of three tags for triangulation;
- survey tags require extended range RFID.

Range	50 meters
Reader type	TBD
Readability:	100 percent

D17 SCIENCE SAMPLE TRACKING VIA UWB RFID

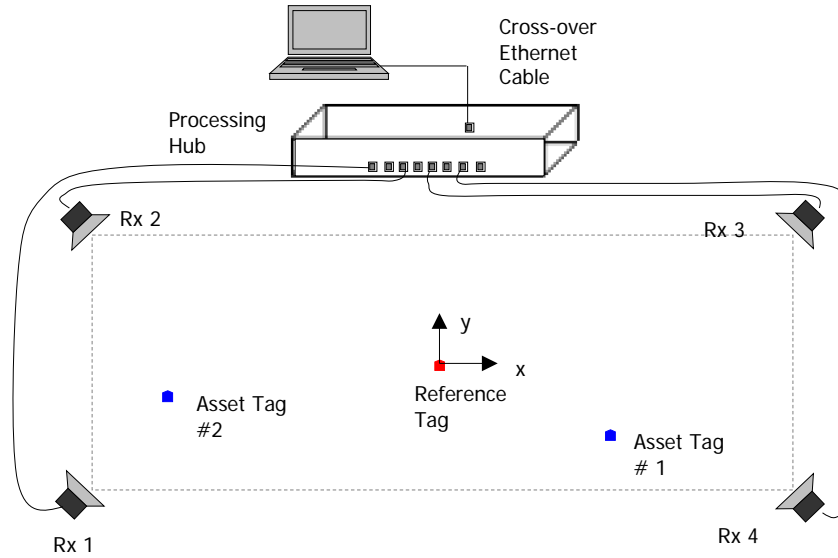


Figure D-11: Science Sample Tracking via UWB Concept

Objective: Provide absolute location of samples within 1 m:

- demonstrated accuracy +/- 10 cm;
- special: requires interrogator (at sample site) with four antennas + local survey of four interrogator antennas for triangulation.

Range	130–150 meters
Reader type	Custom COTS
Readability:	100 percent

D18 LANDING AID

Objective: Provide cooperative radar for Lander with RFID:

- lander is equipped with RFID interrogator and antenna of low- to moderate-directivity, e.g., 8 dBi;
- enhanced passive RFID tags are positioned as panels at the landing site;
- interrogator beam-steering is not required;
- requires extended range RFID tags;
- low TRL: has not been tested.

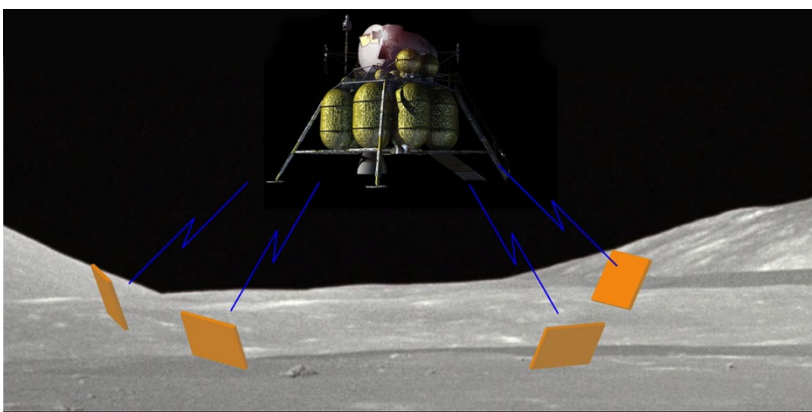


Figure D-12: RFID Landing Aid Concept

D19 LUNAR ROAD SIGN

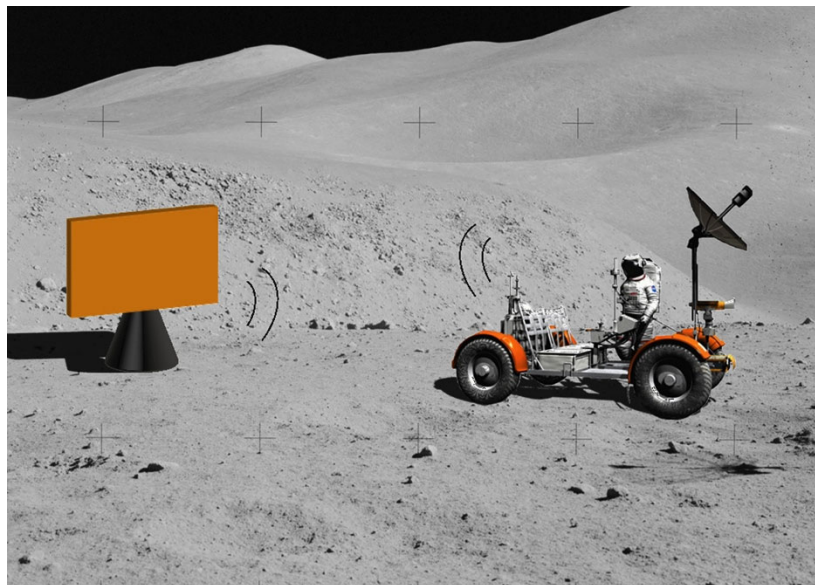


Figure D-13: RFID Lunar Road Sign Concept

Objective: Provide rover with road sign ID and range:

- range is greater than permitted by human vision;
- rover is equipped with RFID interrogator and antenna of moderate to high directivity;
- enhanced passive RFID tags are positioned as road signs upon initial excursions.

D20 VEHICLE SUPPLY TRANSFERS

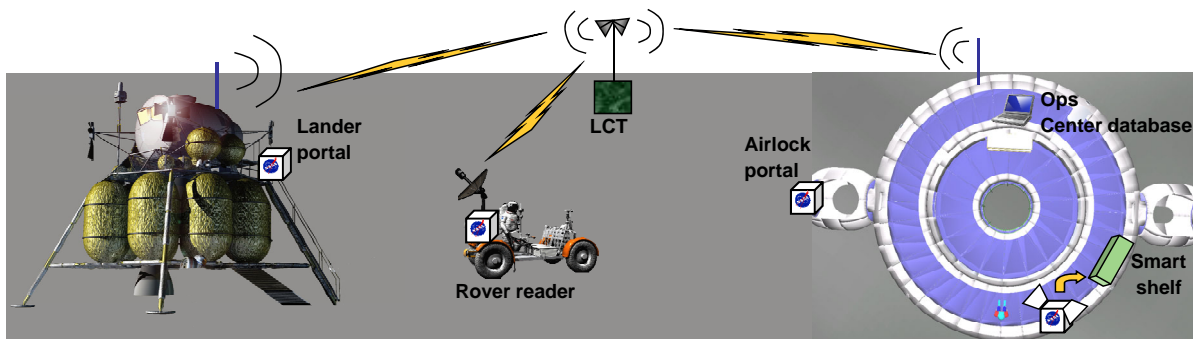


Figure D-14: RFID Vehicle Supply Transfer Concept

Objective: Accurate verification of supply transfers from any supply element to any vehicle.

Description: Ingress and egress of supplies are tracked into and out of any vehicle. RFID interrogation is portal-based. Although RFID technology can be used to determine ingress or egress of assets, auxiliary portal sensors can augment this function. Items are transferred in various forms (e.g., equipment, spares, LRUs, Cargo or Crew Transfer Bags [CTB], etc.) Early application opportunity exists for supply of the CEV Orion. Return On Investment (ROI) for RFID-based inventory management on CEV is questionable since the vehicle will not be resupplied. However, RFID application in tracking supplies to and from the vehicle is considered of significant benefit. Interrogated items will present a variety of material parameters to the interrogator. Cost for high-performance tag antennas, to assure near 100-percent read rates, if required, is likely to be offset by labor savings from reduced ground support and crew time. The technology currently permits high reliability (>90-percent read accuracy) in reading CTB level tags, i.e., tags attached to the exterior of the CTBs. Current read accuracy estimates of item-level tags within CTBs range from 70 to 95 percent, depending on the number of items within the bag and the material parameters of those items. At the intermediate level, sometimes referred to as the ‘Ziploc bag level’, portal read accuracies are typically greater than 90 percent. Vehicle transfers include: Ground-CEV; CEV-ISS; CEV-Lander; Lander-LSAM; Lander-Habitat; Lander-Rover.

Items tagged	Material
Crew Transfer Bag, CTB	Non-conductive
Equipment	Conductive
Clothing	Conductive
Food	Conductive, non-conductive, liquid
Range:	5 meters
Reader type:	Portal
Readability:	≈ 100 percent

ANNEX E

SPACECRAFT USE CASES

E1 INTRODUCTION

Identified intra-spacecraft and assembly, integration and test (AIT) wireless communications use cases for CCSDS agency members are summarized, typically one per page, in the following subsections.

E2 CONTROL OF ROBOTIC AGENTS AROUND A SPACECRAFT

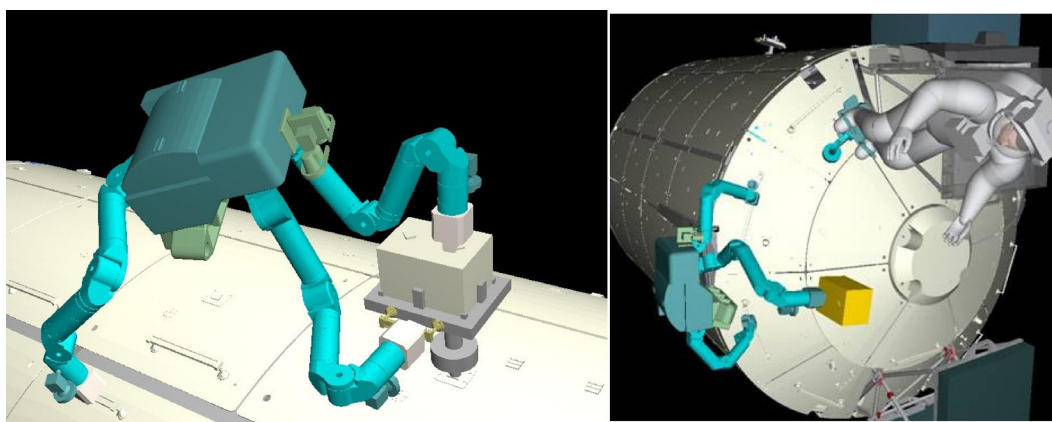


Figure E-1: Control of Robotic Agents

Objective: Give robotic agents the appropriate freedom to move around the ISS while being controlled and transfer data wirelessly.

Description: Robots are designed to execute tasks outside the international space station. They are self-powered, mobile entities required to transmit Real-time video data while being controlled by astronauts within the station or ground personnel. Normally, they shall not have any umbilical cable connections to the Home-Base. Wireless data connection is therefore necessary and the chosen technology must offer enough flexible to insure the communication while the robotic agent moves around the ISS. The complex architecture of the ISS requires that several wireless access points be used in a complementary scheme to offer a global coverage around its structure.

Range:	20m
Data rate:	High
Availability:	High
Criticality:	Medium

E3 WIRELESS SUN SENSORS

Objective: To free self-powered sun sensors from complex and unnecessary harness.

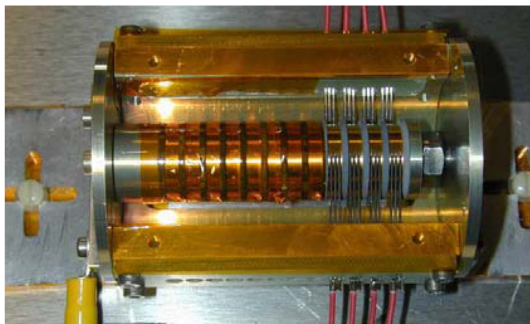
Description: Sun sensors obtain enough energy from the sun to be self-powered. The only remaining cabling is the data link. Integrating a wireless interface to a self-powered sun sensor increases the system flexibility and decreases the design and integration effort. Autonomous wireless sun sensors have been flown in the past with great success (e.g., Delft University of Technology). The use of such a sensor requires the spacecraft to have a wireless interface to communicate with it in a star-like topology.

Range:	2m
Data rate:	Low
Availability:	High
Criticality:	High

E4 ROTARY MECHANISMS



Cable wrap for limited angle



Slip ring for infinite rotation

Figure E-2: Wireless Mechanical Components

Objective: To reduce the complexity of rotating and foldable mechanisms and to offer unrestricted rotation capability.

Description: Any transmission between two parts in movement will generate problems with wires. This problem increases when the number of cycles is high or when the rotating angle is large, which force the designers to have a margin factor as high as 1.5 to 3. Wireless links will have no wear out, infinite rotation capability, no lifetime qualification tests and lower costs. Another example of application would be the energy storage in kinetic momentum.

Range:	20cm
Data rate:	Low to high
Availability:	High
Criticality:	High

E5 ACCESS POINT ON LAUNCHERS

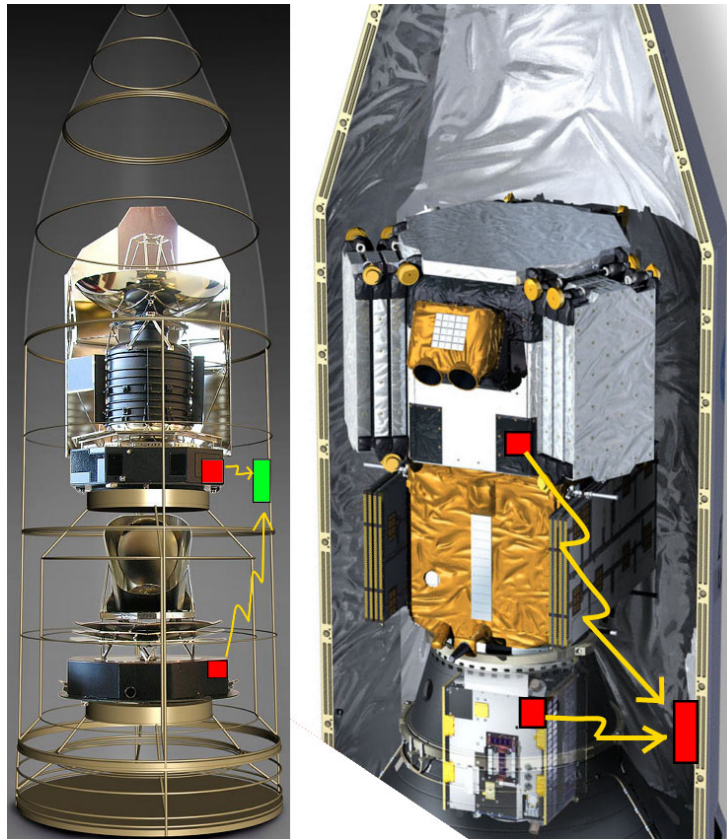


Figure E-3: Wireless Access for Launcher Payloads

Objective: Provide an untethered data link between the launcher payload (satellite) and the launcher data handling system and provide a monitoring facility to the satellites during the launch (thermal, mechanical, vibration, etc.).

Description: A wireless access point on a launcher offers the satellite the possibility to transmit internal monitoring data to the ground without the physical wired bound to the launcher. The launcher shares its data handling system through this interface and simplifies the integration of the payload within the fairing while reducing the risks of failure at separation. This scenario requires that the satellite have a wireless interface to its data handling system as well as a compatible communication protocol that can forward the satellite health data to the ground station.

Range:	2m
Data rate:	Medium
Availability:	Medium
Criticality:	Low

E6 NETWORK OF SENSORS ON LAUNCHER

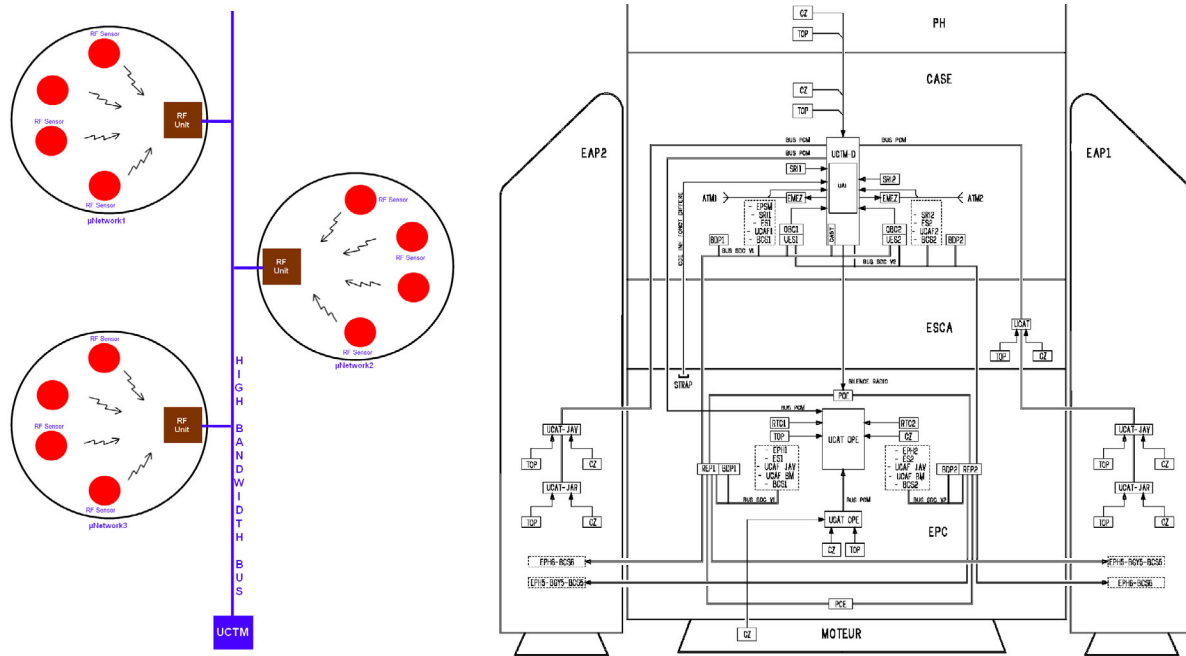


Figure E-4: Launcher and Harness Mass Reduction

Objective: Harness and launcher mass reduction.

Description: There are several dozens of sensors onboard launchers that are wired to the launcher data handling bus. For some types of sensor networks used by launchers, the reliability is not stringent (10^{-4}) but the availability is very important for the telemetry system. Launchers are between 30 and 60 meters tall, which result in long data cables. In the current wired architecture, precautions in the form of bonding and shielding have to be taken in order to protect the relatively small electrical signals against EMI. The extra harness weight on upper stages caused by the shielding itself reduces the deliverable payload capacity. The short mission time of launcher makes the wireless alternative advantageous in regard to the low-capacity, low-weight batteries that can be used to power the wireless interfaces and sensors.

Range:	3m
Data rate:	Medium
Availability:	High
Criticality:	Low

E7 FOLDABLE STRUCTURES

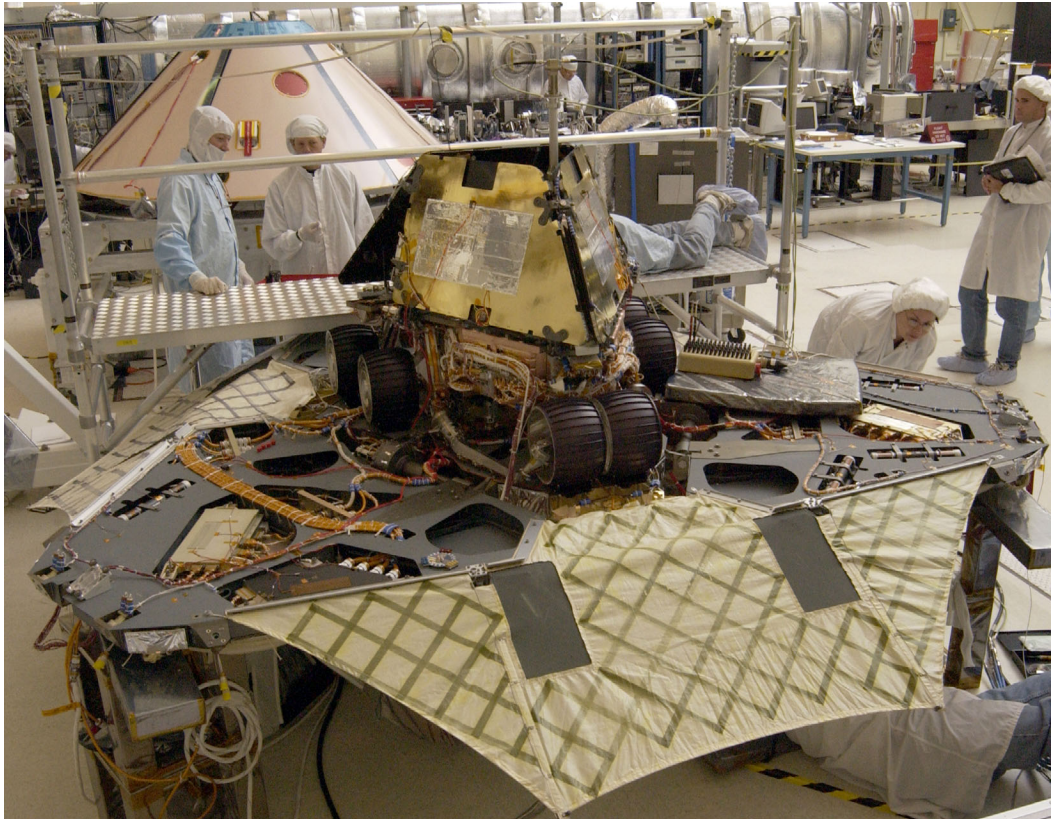


Figure E-5: Inter-Vehicle Wireless Communications

Objective: Create a data connection link between modules that separate (e.g., rover and lander).

Description: There are several subtypes of this use-case, one of them being the interconnection between a lander and its hosted rover. Rovers have power and data lines connected to the lander, this being the only way for the rover to use the solar panels of the transfer vehicle during the space travel phase. At separation, the wires are cut through a thermal process, which induces very high disturbances (e.g., changes in impedance) in the communication bus, therefore requiring the use of higher margins and special dispositions. The connection of the two data handling systems through a wireless link would simplify the separation process and its related risks on the communication bus, while still allowing the health monitoring of the rover during the space traveling phase.

Range:	Meters
Data rate:	Low to high
Availability:	Low to high
Criticality:	Low to high

E8 SCIENTIFIC INSTRUMENTATION WITHIN HEAT SHIELDS

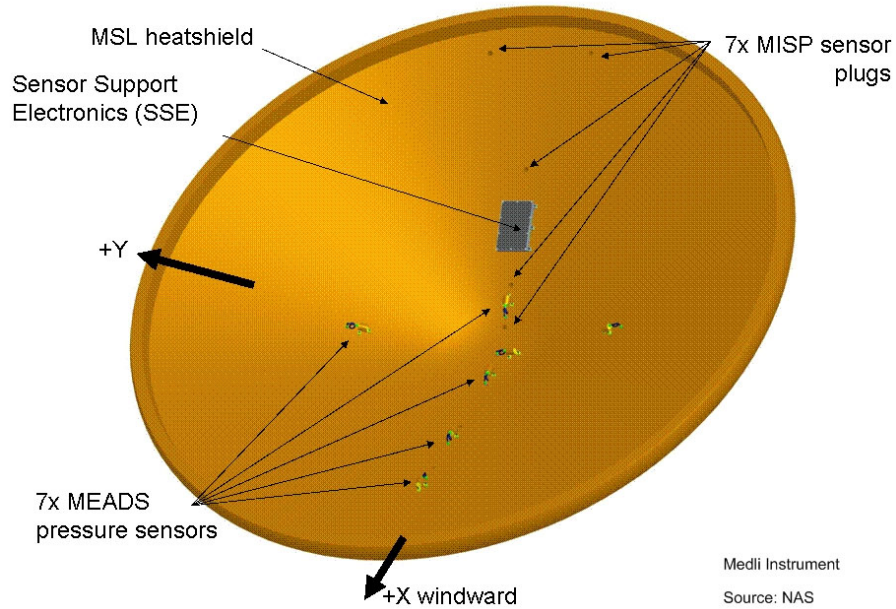


Figure E-6: Science Instrumentation Mass Reduction

Objective: Reduce the mass of the heat shield’s science instrumentation harness, the related AIT time and the risks of the shield separation process.

Description: The heat shields of atmospheric reentry vehicles has been carefully studied and modeled for several decades and permit efficient energy dissipation during the breaking phase in the atmosphere. Contrary to the general perception, there is little empirical environmental data of the heat shield locality for the descent phase. Models have been developed and validated during controlled tests on Earth, but the difficulties implied by the separation of the heat shield from the main vehicle and its corresponding safety issues have limited the deployment of sufficient instrumentation within the shield itself. Typical instrumentation being mainly made of cables connected to thermocouples, thermistors, pressure sensors and to the vehicle’s power source, these direct connections to the main vehicle induce a supplementary risk of separation failure, leading to the reluctance of integrating such instruments. This lack of sufficient and accurate empirical data pushes the spacecraft designers to increase the margins of safety, consequently increasing the heat shield mass. While wireless communication already solves the intrinsic problem of direct cable connection between the shields and the vehicle and its related safety issues, it is believed that wireless sensor nodes replacing the many instrumentation cables may have a considerable mass advantage over a cabled solution.

Range:	2m
Data rate:	Low
Availability:	Medium
Criticality:	Low

E9 CREW DOSIMETRY AND BIOLOGICAL MONITORING



Figure E-7: Crewmember Physiological Monitoring

Objective: Exploration tasks may range from simple intra-vehicular activities, to ambulation on a planetary surface, to construction of outpost habitats. On future Exploration missions, astronauts will be autonomous and required to meet a more rigorous Extra Vehicular Activity (EVA) schedule than previously during the Apollo era. Astronauts will have to respond to contingencies and medical emergencies while providing their own health care. With delayed communications, medical emergencies will need to be addressed by crewmembers trained in emergency medical procedures with minimal or no real-time support from flight surgeons in Mission Control. Wireless technologies can play a significant role in mitigating many human health and performance risks, ranging from critical communications between EVA crew, to enhanced monitoring of crew health and critical biological indicators, to monitoring and reporting of critical suit parameters, to promotion of safety and autonomy by permitting un-tethered mobility.

Description: Biomedical monitoring of physiological parameters during missions is critical to NASA for mitigating astronaut health and for minimizing risk during EVAs. Monitoring human performance and tracking suit consumables during EVA is crucial to ensure overall safety and mission success. Examples of critical parameters affecting human EVA performance are metabolic cost, heart rate (HR), heat rejection and cooling, oxygen consumption (VO₂), and suit pressure. It is vital that quantities of consumables be tracked to support EVA activities within acceptable safety margins. Other additional biomedical monitoring requirements could include methods to minimize suit-induced trauma and improve work and task efficiency during lunar surface operations.

Healthcare communication platforms can also possess the intelligence to dynamically adapt to emergency situations. Inter-suit communications could be implemented where emergency health conditions of an astronaut could be alarmed to other co-located astronauts for immediate medical attention during EVA. In situations in which an astronaut's physiological condition is degrading rapidly compared to other crewmembers, channel allocations can

adapt to permit increased telemetry from the astronaut-under-stress. Suit-to-base communications could also permit the physiological condition of an astronaut to be reported back to an IVA doctor for continuous health tracking and response advisory.

In space, astronauts experience alterations in multiple physiological systems because of exposure to microgravity. Some of these physiological changes include sensorimotor disturbances, cardiovascular deconditioning, loss of muscle mass, and strength. These changes can lead to disruption in the ability to ambulate and perform functional tasks. Health monitoring during IVA and crew exercise provides a means for evaluation and comparison to baseline muscular, neurological, and cardiovascular data collected previously in 1 g, thereby providing insight into crew health and opportunities to customize exercise prescriptions and countermeasures in space. These biological-monitoring functions, however, must not inhibit or constrain crew exercise or IVA activities. Wireless technologies can provide the necessary monitoring functionality without unnecessary tethers or restrictive devices. Other critical areas requiring environmental monitoring for crew health are lunar dust and radiation exposure.

E10 CONTAMINATION-FREE MISSIONS AIT PROCEDURES

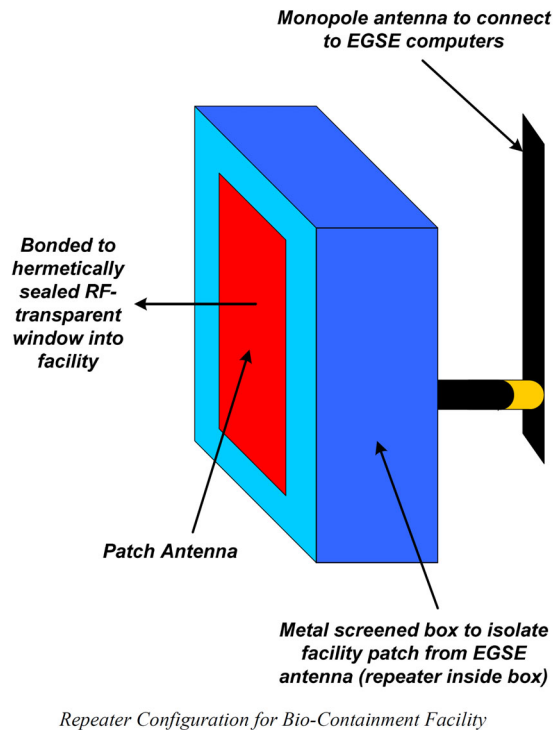


Figure E-8: Contamination-Free AIT Procedures

Objective: Reduce the risks of contamination of samples and by samples.

Description: There are several ways in which wireless systems can support AIT procedures for missions requiring low levels of contamination. The COSPAR regulations for Planetary Protection require that spacecraft intended to land on other planetary bodies are clean and free of biological contamination. The main purpose of these requirements is to maintain as well as possible, the pristine condition of such bodies for the purposes of science.

During AIT or similar procedures that occur prior to launch, the worst source of contamination is due to the presence of humans who carry and shed high levels of biological matter. By minimizing the need for hands on activities and by minimizing the time taken to integrate the spacecraft the risk of contamination can be reduced.

Removal of the need to physically connect equipments reduces human presence in the ultra-clean facilities where the spacecraft is sterilized and maintained clean. EGSE to spacecraft communications can be conducted without umbilicals that often harbor contamination. Pre-integration checks can be conducted before equipments are integrated with the spacecraft confirming correct operation and reducing the likelihood of rework should equipment be found faulty. Use of RFID for managing clean room equipments in the ultra-clean facilities helps also to contain contamination, allowing non-contact inventory management and control. The use of wireless links between clean room personnel and control room staff

removes the need to run signal cables into the clean room to run (for example) activity schedules, present AIT procedural information, and to record events as they occur. Working in ultra-clean facilities requires that the environment be constantly monitored to detect contamination that must be recorded as evidence of the cleanliness of the AIT process as well as the spacecraft. The use of wireless devices simplifies installation and also replacement in the event of failure of such a device. The absence of cables (for self powered devices in particular) also allows more flexibility of placement so the sensors can be placed for optimum effect or sensitivity.

Interplanetary spacecraft, because of the need to be compact for delivery purposes, are usually tightly packed and of complex configuration. The use of wireless technology simplifies the integration process, simplifies rework should it be necessary, reduces schedule cost and risks to the program.